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## GLOBAL JOURNAL OF COMPUTER SCIENCE AND TECHNOLOGY: F Graphics & Vision

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## Contents of the Volume

- i. Copyright Notice
- ii. Editorial Board Members
- iii. Chief Author and Dean
- iv. Table of Contents
- v. From the Chief Editor's Desk
- vi. Research and Review Papers
- 1. 3D Interfaces for Real Time Monitoring of Radwaste Storage. 1-5
- 2. 3D Wavelet Transformation for Visual Data Coding with Spatio and Temporal Scalability as Quality Artifacts: Current State of the Art. *7-21*
- 3. Performance Analysis of Intensity Averaging by Anisotropic diffusion Method for MRI Denoising Corrupted by Random Noise. *23-29*
- 4. Robust Watermarking Method for Color Images Using DCT Coefficients of Watermark. 31-34
- 5. An Impressive Method to Get Better Peak Signal Noise Ratio (PSNR), Mean Square Error (MSE) Values Using Stationary Wavelet Transform (SWT). 35-40
- 6. Performance Analysis of Modified Lifting Based DWT Architecture and FPGA Implementation for Speed and Power. *41-50*
- vii. Auxiliary Memberships
- viii. Process of Submission of Research Paper
- ix. Preferred Author Guidelines
- x. Index



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## 3D Interfaces for Real Time Monitoring of Radwaste Storage By Gianfranco Vecchio & Paolo Finocchiaro

Istituto Nazionale di Fisica Nucleare

*Abstract* - This paper describes the design and development of a 3D interface in the framework of an application for the monitoring of a radioactive waste storage. It focuses on the description of software solutions by integrating different technologies. We used only free and open-source libraries to develop the 3D environment that were subsequently integrated into the existing web application developed used Java Server Faces. In order to implement 3D graphics we used Away3D, an open source 3D graphics engine for Adobe Flash, written in ActionScript 3 and runnable in every browser that utilizes Adobe Flash Player.

Keywords : Radiation monitoring, Radioactive waste, Interactive systems, Graphics Animation, Graphical models.

GJCST-F Classification: I.2.10



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# 3D Interfaces for Real Time Monitoring of Radwaste Storage

Gianfranco Vecchio<sup>a</sup> & Paolo Finocchiaro<sup>o</sup>

*Abstract* - This paper describes the design and development of a 3D interface in the framework of an application for the monitoring of a radioactive waste storage. It focuses on the description of software solutions by integrating different technologies. We used only free and open-source libraries to develop the 3D environment that were subsequently integrated into the existing web application developed used Java Server Faces. In order to implement 3D graphics we used Away3D, an open source 3D graphics engine for Adobe Flash, written in ActionScript 3 and runnable in every browser that utilizes Adobe Flash Player.

IndexTerms : Radiation monitoring, Radioactive waste, Interactive systems, Graphics Animation, Graphical models.

#### I. INTRODUCTION

his document wants to show a new useful way to display information from data collected in a harsh environment. An electronic system for radioactive waste monitoring was developed at INFN-LNS laboratory, as part of DMNR project [1]. It consists of a network of sensors deployed around radwaste drums for the continuous detection of radiation [2]. The sensors employed so far, specially developed for this case, are made of plastic scintillating fibers whose light is readout by means of pairs of Silicon Photomultipliers (SiPM). A counting system based on FPGA boards handles the data flow coming from groups of these sensors and sends it to the front-end of the application. A relational database system is used to store the historical data, and a dynamical web application was developed to explore the database and display the corresponding data in graphical form.

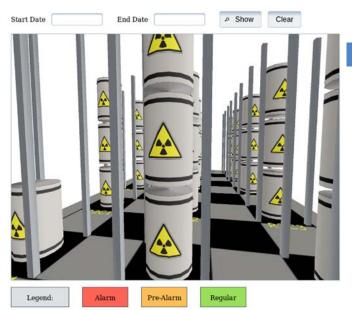
To improve the view of the storage following a database query we developed a 3D scene that shows the status of the whole plant at a glance [3], [4]. We developed two kinds of 3D scenes: a representation of the whole plant with the possibility to navigate through the drums, and the detailed view of a single structure of three drums with fiber sensors around them. In this document we are going to describe a setup with up to three drums inside tower-like structures and arranged in chessboard fashion.

The first scenario is used to investigate the status of the repository in a user-selected time range,

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and to inspect every single drum by virtual navigation through the plant using mouse and keyboard, as shown in Fig. 1.



#### Fig. 1 : 3D view of radioactive waste repository

Once the user fills the date textboxes data coming from the database will color the drums in red, orange or green according to the corresponding alarm condition (Fig. 2).

Should there be no data in the database for the chosen time range, the drum will be drawn as grey. The colored drums can be clicked on by the user, in order to have a detailed view of its status (i.e. of all the sensors surrounding it).

The second scenario represents the detail of a single box containing the drum clicked in the previous scenario. It also draws the fiber sensors around each drum (four horizontal fibers in this sample case), and the interactivity consists on the possibility to rotate the tower to gain visual access to all the fibers and their alarm status. Clicking on a fiber one can get the plot of its counting rate in the time range selected. If one clicks on a drum the collective plot for the four fibers of that drum is shown in the same chart (Fig. 3).

Moreover, clicking on a drum a polar chart of the counting rate for each sensor will be shown, along with a chart indicating maximum, minimum and average counting rate values.

E-mail α : vecchio@lns.infn.it

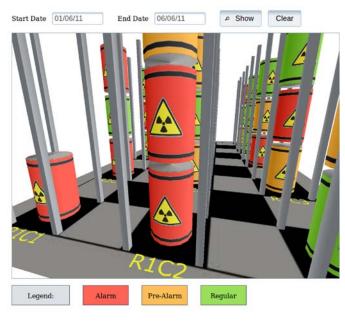


Fig. 2 : 3D view of radioactive waste plant after database query

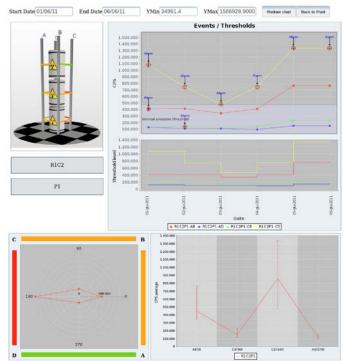
#### II. Away3D, A Library for Adobe Flash to Create the Storage Plant

In order to simulate the environment that we have to monitor as well as possible, we chose Adobe Flash for its powerfulness in the creation of complex interactive animations. Away3D [5]-[8] is a 3D graphic engine written for Adobe Flash platform in ActionScript 3 [9] originally derived from Papervision3D. It was designed to be fast and extensible and allows the easy creation of various 3D objects with the use of its graphics primitives. Working on 3D coordinates and leveraging the object oriented nature of the language, we were able to build a realistic scene of our plant. We developed our project with Flash Develop [10], an opensource Integrated Development Environment to write and to compile Action Script code and to create Shockwave Flash objects (SWF).

The navigation inside the nuclear repository is done by means of a camera, defined by Away3D, moved using mouse and keyboard. Away3D offers several different types of camera, that represent the viewer's position in space relative to the rest of the scene. For the scene representing the whole storage we used the default *Camera3D* that produces a standard front-facing perspective projection of the scene. Thanks to event handles and objects like *MouseEvent3D* and *KeyboardEvent*, we were able to add some interactivity and better user experience. Hence, like in a computer game, by using the mouse and the arrow keys (or W-A-S-D keys) one can walk and look around every drum inside the environment.

As for the scene regarding the single tower box, which can be rotated using the mouse, we employed a

different camera, the *HoverCamera3D*, that allows the user to navigate each side of the box only changing some properties like distance and rotation angles.



*Fig. 3*: 3D view of a box with detail of the four sensors. Clicking on a fiber or on a drum one can view the global counting rate chart

Each object rendered to the scene was built from geometric elements like triangles and line segments. To make this mesh more realistic we had to build images to be set in the material property of each element. To make towers, drums and fibers we used the same Away3D primitive, the *Cylinder*, and then we used different material for every type of element.

Fig. 4 shows the result of the storage plant before the application of materials in each element. We can see only the mesh that was used to build every object made by triangles and line segments, showed by default in different color by the 3D graphic engine.

Furthermore, we were able to write text inside the interactive scenes, embedding the font inside an SWF object. We used text for orientation purposes and to view the ID of each element.

Finally, the possibility to pass some parameter to external resources, was achieved thanks to flash class *ExternalInterface* that enables straightforward communication between ActionScript and SWF containers, that in our case consisted of the XHTML page with JavaScript:

ExternalInterface.call("javascript\_function\_to\_cal 1", parameters\_to\_pass\_to\_javascript);

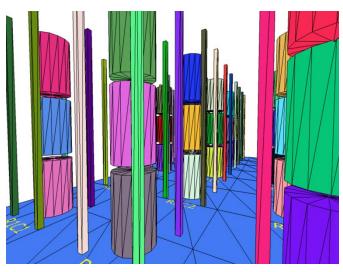


Fig. 4: Mesh of elements inside the storage plant

#### III. INTEGRATION BETWEEN DIFFERENT TECHNOLOGIES: Adobe Flash and Java Server Faces

The most complex task was represented by the encapsulation of the flash object inside the existing Java Server Faces [11], [12] project and the data exchange between the two different technologies [13]. The elements in the storage plant have to change according to the data coming from the database. Our web application contains a Java classes (the Managed Beans) that encapsulate the state of an object, and then stores all the data coming from the database after a user query. To pass data between Managed Beans and

*composite components*, introduced after 2.0 version of JSF. These allow to create reusable components and to embed external objects easily.

In our application we defined two composite components, one for each SWF object. They are simply XHTML files, implemented in JSF fashion, containing an interface and an implementation. JSF uses namespaces to include new tags to be used in the XHTML page, and for composite components the namespace is:

#### xmlns:cc="http://java.sun.com/jsf/composite"

The interface contains configurable features, like input and output data, whereas the implementation contains JSF tags needed to build the new component.

Inside the interface we declared variables that are needed to pass parameters between applications through the *attribute* tag:

#### <cc:attribute name="drums">

In the implementation, instead, we embedded the flash object thanks to JavaScript and the open source script *SWFObject* [14]. It offers methods to insert flash contents in HTML and XHTML pages, using only a small JavaScript file. It is more optimized and flexible than any other currently available method for the insertion of flash content and it is easy to use. Indeed, we need to use only this function to embed our SWF object:

swfobject.embedSWF(swfUrl, ``800", ``600", ``11.1.0",
flashvars);

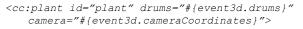
This method accepts as parameters the URL of an SWF file, the size of the window that will show the flash content, the flash player version, and the *flashVars* variables that include data from Managed Bean.

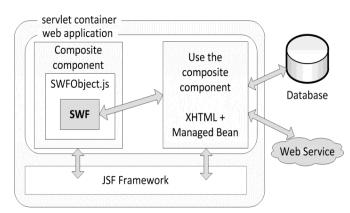
Fig. 5 shows the schematic approach of this integration, describing the use and the configuration of composite components.

In order to use a composite component we only need to declare a namespace with the name of the directory that contains it:

xmlns:cc="http://java.sun.com/jsf/composite/flash"

and just a line is needed to embed it in our code and to pass the parameters in the classic JSF notation:





## *Fig. 5* : Scheme of the use and configuration of composite components

The flash object needs dynamic data to work that come from the database. As seen before, they will be passed to the flash object by means of *flashVars* variables. The attributes *drums* and *camera*, in the code shown above, are declared in the interface section of composite component and they will be used in the implementation code and inside the definition of flashVars variable:

var flashVars = {drums: "\${cc.attrs.drums}", camera: "\${cc.attrs.camera}"};

In order to retrieve these data, on the flash application side, we need to use only the *LoaderInfo* class from *flash.display* package:

var input:String = LoaderInfo(this.root.loader
Info).parameters.drums;

var data : Array = input ? input.split(/[^a-zA-ZO-9\-]+/) : []; Hence, the data structure coming from flashVars variable, will be decrypted as a string, that we changed into an array structure for convenience.

To interact with the flash object, clicking on a fiber or on a drum, we needed the implementation of a bidirectional communication system, then of parameter passing between a flash object and a managed bean. For this reason we used JavaScript like an interface between the two different technologies. As seen in the II Chapter, with ActionScript 3 we only need to use the *ExternalInterface* class to call the JavaScript function that resides in the implementation section of composite component:

```
ExternalInterface.call("clickPlant", drums, camera
Coordinates);
```

We pass two parameters to the JavaScript function: the ID of the clicked drum and the current camera coordinates to retain the position and orientation of the camera after navigation. Finally, these parameters will be associated to managed bean attributes through hidden input text called by JavaScript:

At this point, with managed bean attributes set, we can show the charts related to the clicked drum or fiber in the traditional manner in JSF paradigm.

#### IV. Conclusion and Prospects

In this paper we presented a new candidate system to show and to monitor a radioactive waste storage plant with 3D graphics interactions. It is useful to have a simulated perception of the waste storage at a glance, and a realistic way to see the elements that belong to the repository.

We investigated how to take advantage of new features in JSF 2.0, as composite components, and integrate Flash into our JSF application. These new capabilities in JSF free us from the need to take care about data encoding and decoding for transfer. Furthermore, to pass variables and parameters we used JavaScript variables and methods, that allow an easy way to exchange data between JSF and Flash.

Finally, to embed the flash content we used SWFObject, that detects the Flash Player version and determines whether Flash content or alternative content should be shown, to prevent outdated Flash plugins to break Flash content. It offers a simple way to embed Flash content and it is more optimized and flexible than any other Flash Player embed method around.

The future work will consist of the graphic representation of different sectors of larger repositories, to be shown dynamically depending on the user selection. Moreover, we will investigate about additional graphical improvements, especially with the use of the new release of *Away3D*, the version 4.0, which will completely support the release 11 of Adobe Flash Player. Indeed, the latest version of this plugin allows to take full advantage from graphic acceleration of the Graphics Processing Unit, releasing the CPU from such a heavy computational task.

A possible improvement will consist in correlating the virtual navigation procedure with a real inspection performed by means of a robotic arm, currently under development, equipped with a CCD color camera and a spectroscopic gamma ray detector [15].

A pilot DMNR system, which will include the 3D virtual navigation interface described in this paper, is going to be installed soon for tests in a real radioactive waste storage site.

#### V. Acknowledgment

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## 3D Wavelet Transformation for Visual Data Coding with Spatio and Temporal Scalability as Quality Artifacts: Current State of the Art

By Shaik. Jumlesha & Dr.Ch.Sathyanarayana

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*Abstract* - Several techniques based on the three–dimensional (3-D) discrete cosine transform (DCT) have been proposed for visual data coding. These techniques fail to provide coding coupled with quality and resolution scalability, which is a significant drawback for contextual domains, such decease diagnosis, satellite image analysis. This paper gives an overview of several state-of-the-art 3-D wavelet coders that do meet these requirements and mainly investigates various types of compression techniques those exists, and putting it all together for a conclusion on further research scope.

Keywords : Discrete Cosine Transform, Discrete Wavelet Transform, Image Compression, spacio scalability, temporal scalability.

GJCST-F Classification: I.2.10



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# 3D Wavelet Transformation for Visual Data Coding with Spatio and Temporal Scalability as Quality Artifacts: Current State of the Art

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*Abstract* - Several techniques based on the three–dimensional (3-D) discrete cosine transform (DCT) have been proposed for visual data coding. These techniques fail to provide coding coupled with quality and resolution scalability, which is a significant drawback for contextual domains, such decease diagnosis, satellite image analysis. This paper gives an overview of several state-of-the-art 3-D wavelet coders that do meet these requirements and mainly investigates various types of compression techniques those exists, and putting it all together for a conclusion on further research scope.

*Keywords : Discrete Cosine Transform, Discrete Wavelet Transform, Image Compression, spacio scalability, temporal scalability.* 

#### I. INTRODUCTION

n various image and video applications compression is indispensable to guarantee interactivity during the streaming and consultation in particular about huge volume of medical images, for probing contextdependent full visual structures and/or quantitative analysis of the measurements. As a consequence, trading-off visual quality and/or implementation difficulty against bit-rate introduces exact constraints. On one hand, it is unbearable to drop any information when handling context exact visual data such as medical data.

On the other hand, a model likes progressive data transmission [52] and, thus, naturally support for lossy coding is equally important. This methodology allows for example to prioritize low-resolution edition of the requested motion based or still visuals and to increasingly filter the resolution of the visualized data by move additional data. This scalability mode is often referred to as resolution scalability. In a quality scalability scheme, the visual media is decoded instantaneously to the complete resolution but with a reduced visual quality. Additionally, by choose regions that are applicable for the context such as diagnosisi.e., the regions-of-interest (ROIs)-parts of the image can be assess in a very early transmission stage at full guality. Meanwhile, the background information will be further developed.

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Author σ : Associate Professor in Cse & Hod Jntu Kakinada. E-mail : chsatyanarayana@yahoo.com Moreover, it should be clear that we target best ratedistortion presentation over the complete range of bitrates that is demanded by the application. For example, JPEG2000 [53] (based on the wavelet transform) clearly outperforms its predecessor PEG [based on the discrete cosine transform (DCT)] [54] at low bit-rates and has as a significant property its lossy-to-lossless coding functionality; that is the ability to start from loss density at a very high density ratio and to increasingly refine the data by sending detail information, finally up to the stage where a lossless decompression is obtained.

Systems based on technologies other than the wavelet transform have been proposed, but they only partially assist the requested set of functionalities. Nonetheless, those techniques do superb for the subclass of applications they are designed for. Examples are context-based predictive coding (CALIC) [55] for lossless compression and region-based coding for very low bit-rate coding. Although these coders are competitive in their application domain, they lack support for the other functionalities.

Additionally, the increasing use of threedimensional (3-D) imaging modalities, like magnetic resonance imaging (MRI), computerized tomography (CT), ultrasound (US), single photon emission computed tomography (SPECT), and positron emission tomography (PET) triggers the require for capable techniques to transport and store the associated volumetric data. In the classical approach, the image volume is careful as a set of slices, which are accumulate consecutively compressed and or transmitted. Since modern transmission techniques need the use of concepts like rate scalability, quality scalability, and resolution scalability, multiplexing mechanisms require to be introduced to select from each slice the right layer(s) to support the actually essential quality-of-service (QoS) level. However, a disadvantage of the slice-by-slice mechanism is that potential 3-D correlations are ignored.

#### II. QUALITY ARTIFACTS FOR VISUAL DATA CODING

In the applications of distributing motion based visuals such as video streaming or still visuals such as volumetric image sets over networks, distributing server has to deal with different network environments and client devices. The very diverse connection in mixed networks, ranging from hundreds of mega-bos to a number of tens of kilo-bps, and the fluctuations in bandwidth, need that video bit-streams are flexible in adapting itself to dynamic channel during transmission. The different client devices, with their different display capabilities and their computational and memory limitations, also need that bit-streams are flexible in decoding resolution and complexity. For these reasons, in recent years, apart from continuously improving the density effectiveness of non scalable video coding, a lot of research efforts have been paid to providing different scalabilities as quality artifacts in compressed bitstream, including spatial resolution, frame rate, quality and temporal scalabilities.

Among several scalable video coding schemes, some based on three-dimensional wavelet transform have concerned much attention [56],[57],[21], [20], [60]. In these schemes, wavelet filtering is useful in both spatial and temporal directions. The multi resolution property of wavelet representation makes it a nature solution for spatial-temporal resolution scalability. The resolution scalability is usually realized by dropping needless sub bands. Hence the fallowing two considered as quality artifacts for 3D wavelet based coding strategies.

- 1. Spatial scalability
- 2. Temporal scalability

#### III. SPATIAL SCALABILITY AS QUALITY ARTIFACT FOR 3D WAVELET CODING

Spatially scalable or hierarchical video coders generate two bit-streams: a base layer bit-stream, which represents low-resolution pictures, and an improvement layer bit-stream, which provides additional data wanted for reproduction of pictures with full resolution. A significant feature is that the base- layer bit-stream can be decoded separately from an enhancement layer. Therefore, low-resolution terminals are capable to decode only the base-layer bit-stream in order to display low-resolution pictures. Such density techniques are of great interest recently, because of growth of communication networks with different transmission bit rates [62]-[65]. Moreover, scalable transmission is useful in error-prone environments where base-layer packets are well secluded against transmission errors and losses, while the security of the enhancement layer packets is lower. In such a system, a receiver is able to copy at least low-resolution pictures if quality of service decreases.

There were several attempts to develop spatially scalable coding of video. The proposed schemes were based on pyramid decomposition [61] or subband/wavelet decomposition [62], [63], [65]. Among different proposals, the latter approach should be each image into four spatial sub-bands. The sub-band of lowest frequencies comprises a base laver, while the other three sub-bands are jointly transmitted in an improvement layer (Fig. 1). Nevertheless, this approach often leads to allotment of much higher bit rates to a base layer than to an improvement layer, which is disadvantageous for practical applications. Recently, Benzler [66] has proposed to keep away from this problem by combining spatial and SNR scalability and abandon the obligation of the full MPEG-compatibility in the base layer. Here, our goal is to use a fully MPEGcompatible coder in the base layer. For the essential codecs, spatio-temporal scalability is proposed [67], [68]. Here, a base layer corresponds to the bit-stream of the pictures with compact both spatial and temporal resolutions. Therefore, in the base layer, the bit rate is decreased as compared to a encoder with spatial scalability only. Now, it is easy to get the base layer bit rate equal or even less than that of the development layer. The development layer is used to transmit the information required for restoration of the full spatial and temporal resolution.

considered especially promising. The idea is to divide

Embedding of sub-band decomposition into a motion-compensated encoder leads to in- or out-band motion reparation performed on individual sub-bands or on the whole image, respectively. The latter will be used here, because some experimental results show that it is more capable [62], [63], [65].

Here, the term of spatio-temporal scalability is proposed for a functionality of video compression systems where the base layer corresponds to pictures with compact both spatial and temporal resolution. An improvement layer is used to transmit the information required for return of the full spatial and temporal resolution.

The authors have already considered two basic approaches related to spatio-temporal scalability [67], [68]. The first approach exploits 3-D sub-band analysis while the second approach is based on B-frame data partitioning.

#### a) First Approach

The input video sequence is analyzed in a 3-D separable filter bank, i.e., there are three successive steps of analysis: temporal, horizontal, and vertical. For temporal analysis, very simple linear-phase two-tap filters are used similarly as in other papers on three-dimensional sub-band coding [69], [70]

$$H(z) = 0.5.(1 \pm z^{-1})$$

Where "+" and"-"correspond to low- and high pass filters, respectively. This filter bank has a very simple implementation, wants to store one frame only and exhibits small group delay. Temporal analysis results in two sub-bands  $L_{t}$ 

and  $H_t$  of low and high temporal frequencies, respectively. In both sub-bands, the temporal sampling frequency is compact by factor two. Therefore, these two sub-bands correspond to two video sequences with compact frame frequency. The two sub-bands are separated into four spatial sub-bands (LL, LH, HL, and HH) each. For spatial analysis, both horizontal and vertical, independent FIR filters are used. The 3-D analysis results in eight spatio-temporal sub-bands (Fig. 2). Three high-spatial-frequency sub-bands (LH, HL and HH) in the high-temporal-frequency sub-band  $H_t$  are discarded, as they correspond to the information being less applicable for the human visual system.

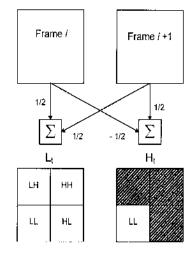


Fig. 2 : 3-D sub-band analysis

According to the experimental authors' tests for 720 x 576 progressive 50-Hz test sequences, it reduces PSNR often to about 32-33 dB, and has unimportant control on subjective quality of the decoded video. Thus, five sub-bands are encoded:

- In a base layer—the spatial sub-band LL of the temporal sub-band L<sub>r</sub>;
- The improvement layer includes the spatial subbands LH, HL, and HH from the temporal sub-band  $L_t$  and the spatial sub-band LL of the temporal subband  $H_t$ .

#### b) Second Approach

In the second alternative, the technique employs data structures already designed for standard MPEG-2 coding. Reduction of temporal resolution is obtained by elimination of each second frame. It is unspecified that groups of pictures (GOPs) consist of even number of frames. Moreover, it is unspecified that each second frame is a B-frame, i.e., it can be removed from a sequence without moving the decoding of the remaining frames. Reduction of spatial resolution is obtained by use of sub-band decomposition. Suitable design of the filter bank results in negligible spatial aliasing in the LL sub-band, which constitutes the base layer. Unfortunately the technique does not provide any means to suppress temporal aliasing. The effects of temporal aliasing are like as those related to frame skipping in hybrid encoders.

The base-layer data are used to create lowquality images; therefore, it is sensible to perform more rough quantization here than in the improvement layer. On the other hand, quality of the sub-band LL is strongly related to the quality of the full sized picture. The small quality of the LL sub-band restricts the full-sized picture quality to a relatively low level, in spite of the amount of information in the remaining sub-bands. Therefore, it is important to transmit additional information  $\Delta LL$  in the improvement layer. This information is used to get better quality of the sub-band LL when used to synthesize fullsized images in the improvement layer.

## IV. Temporal Scalability as Quality Artifact for 3D Wavelet Coding

Temporally scalable video coders can be classified intone of two types depending on the manner in which temporal redundancy is exploited. The first is the motion-compensated analytical coder (e.g. MPEG, [71]), and the second is the sequential sub band coder both without ([72], [73]) and with ([74], [75]) motion compensation (MC). The coding competence and the degree of temporal scalability are a function of size of a group-of-frames (GOF), defined as the number of successive frames that can be decoded separately from the rest of the video sequence. Temporal scalability is achieved by decoding subsets of the GO consisting of equally-spaced frames.

MCP exploits temporal redundancy by forming a(closed loop) prediction of the current frame via MC from a reference frame. The coding competence of MCP is needy on the success of the MC. Good MC considerably increases the association coefficient among pixels, which then yields less energy in the residual. Temporal scalability from MCP video is provided by strategic placement of orientation frames and selective decoding of frames. Recursive prediction with a GOF of length allows lower frame rates of 1/2, 1/4, 1/2v. The result is a simple temporal sub sampling of the original sequence.

Temporal sub band coders and motioncompensated TSB coders exploit redundancy by applying a sub band or wavelet analysis in time. The most usually used filters are the 2-tap Haar wavelet filters applied hierarchically, which result in GOFs with lengths that are powers of 2.Temporal scalability is provided by decoding and synthesizing chosen temporal sub bands.

To include block-based motion recompense with TSB, MC is performed on individual blocks prior to each application of the low-pass filter and therefore local motion of different objects in a scene can be compensated well. To make sure inversion ability of the wavelet transform, MC must be full-pel. The resulting scene-aligned pixels are temporally sub band filtered. However, regions of pixels can now be precious by the occlusion problem which occurs when a one-to-one correspondence among pixels in two frames does not exist in the MC operation. These regions must be especially coded. The standard solution issue analytical coding to maintain reconstruct capacity for these pixels [4]. Regions in the prior frame not found in the present frame are placed in the low-pass sub-band and coded directly. Regions in the current frame not found in prior frame are subtracted from a prediction and placed in the high-pass sub band. Since quantization occurs after forming the temporal sub band representation, such coding is efficiently an open-loop prediction system.

An important difference between MC-TSB and MCPc ding is the effects of full- pel and half-pel MC. MC can be considered as preprocessing previous to temporal sub-band filtering. While the temporal sub band filtering (and predictive coding for covered/uncovered pixels) is invertible, the preprocessing step must also be invertible to allow exact synthesis of the original frames in the absence of quantization. Inverting capacity is only provided by full-pel motion compensation.

#### V. Nomenclature of 3-d Wavelet Coding

#### a) 3-D DCT Coding

The first coder introduced in the 3-D test bed is a JPEG-alike, 3-D DCT-based coder. This coder was designed in order to have a good reference for DCTbased systems. The 3-D JPEG-based coder is composed of a DCT, followed by a scalar quantize and finally a combination of RLC and adaptive arithmetic encoding. The basic principle is simple: the volume is separated in cubes of 8x 8 x 8 pixels (N=8) and each cube is separately3-D DCT-transformed, similar to a classical JPEG-coder.

Thereafter, the DCT-coefficients are quantized using a quantization matrix. In order to derive this matrix, one has to consider two options. One option is to construct quantization tables that create an optimized visual quality based on psycho-visual experiments. It is valuable mentioning that JPEG uses such quantization tables, but this approach would need complicated experiments to come-up with sensible quantization tables for volumetric data. The simplest solution, adopted in this work, is to create a uniform quantization matrix—as reported in[76], [77], and [78]. This option is motivated by the fact that uniform quantization is optimum or quasi-optimum for most of the distributions

$$DCT(u) = \sum_{x1=0}^{N-1} \sum_{x2=0}^{N-1} \sum_{x3=0}^{N-1} f(x) \prod_{i=1}^{3} C(u_i) \cos[\frac{(2x_i+1)\pi}{2N}u_i]$$

With

$$C(u_i) = \begin{cases} \sqrt{\frac{1}{N}}u_i = 0\\ \sqrt{\frac{2}{N}}u_i > 0 \end{cases}$$

The quantized DCT-coefficients are scanned using a 3-Dspace-filling curve, i.e., a 3-D instantiation of the Morton-curve[81], to allow for the alignment of zerovalued coefficients and, hence, to get better the performance of the RLC. This curve was opted for, due to its simplicity compared with that of 3-D zigzag curve [82]. The nonzero coefficients are encoded using the same classification system as for JPEG. The coefficient values are grouped in 16 main magnitude classes (ranges), which are subsequently encoded with an arithmetic encoder [83]. Finally, the remaining bits to refine the coefficients within one range are added without further entropy coding.

The adopted entropy coding system is partially based on the JPEG architecture [54], although the Huffman coder is replaced by an adaptive arithmetic encoder [83]. Consequently, the big look-up tables mentioned in annex K of the standard [54] are extra and moreover, adaptive arithmetic encoding tends to have a higher coding efficiency. The dc coefficients are encoded with a predictive scheme: apart from the first dc coefficient, the entropy coding system encodes the difference between the current dc coefficient and the previous one. For this distinction, the range is determined and encoded with an arithmetic encoder that has a dc model supporting16 ranges. Simply transmitting the remaining bits of the coefficient refines the range specification without any further entropy coding. The latter is possible since the probability distribution of all possible values can be seen as uniform, hence, entropy coding will not be capable to further reduce the bit consumption.

The ac coefficients are encoded by specifying first the amount of zeros preceding the encoded symbol, i.e., the run. The runs of zeros are encoded using an arithmetic encoder with a separate model. Runs of up to 15 zeros are supported. Note that to indicate the situations in which 16 or more zeros precede a important coefficient, an extra symbol "OVF" (overflow) is used. After encoding this symbol, the remaining zeros are directly encoded to avoid confusing situations involving a succession of several OVF encodings. Finally, the range of the encountered important symbol is encoded, using an arithmetic encoder with a similar (AC) model as in the case of the dc coefficients, followed by the essential refinement bits.

#### b) The 3-D Wavelet Transform

Before describing in the following sections the proposed 3-Dwavelet-based techniques, it is significant to notice that these techniques support lossless coding, all the necessary scalability modes as well as ROI coding and this is a important variation with respect to the 3-D DCT-technique presented above, which is not able to provide these features.

For all the 3-D wavelet-based coders included in this study, a common wavelet transform module was designed that supports lossless integer lifting filtering, as well as finite-precision floating-point filtering. A mixed selection of filter types and a different amount of decomposition levels for each spatial direction (x-, y-, or z-direction) are supported by this module. This allows for adapting the size of the wavelet pyramid in each spatial way in case the spatial resolution is limited.

For example, fewer levels will be required along the slice axis if the amount of slices or the resolution along the axis is limited. The supported lossless integer lifting filters include the (S+P), (4,2), (5,3), and (9,7) integer wavelet transforms. This selection is based on current publications [85], [86], as well as investigations performed in the context of the JPEG2000 compression standard.

A typical problem encountered with 3-D lossless integer wavelet transforms is the complexity wanted to make them unitary, which is not the case for floating-point transforms. This property is essential in order to achieve a good lossy coding performance. By calculating the L2 norm of the low- and high-pass filters, the normalization factors can be determined. In two dimensions, this is not a problem, since the typical scaling factors to obtain a unitary transform are about powers of two [87].

However, in three dimensions, the problem pops up again and it only disappears if one takes care that the sum of all decompositions influencing each individual wavelet coefficient (i.e., decompositions in both slice directions and in the axial direction) is an even number. Hence, some proposals have been formulated [88], [89] that make use of a wavelet packet transform [90] to achieve this goal, while assuming that the L2based normalization factors for the supported kernels

scale-up 
$$\sqrt{2}$$
 with for the low-pass and  $\frac{1}{\sqrt{2}}$  for the high-

pass kernels. In practice this seems to be an acceptable approximation. Nevertheless, in the presented study, whenever possible, unitary transforms will be used (and it will be explicitly mentioned if not).

#### c) 3-D SPIHTs

In the test set of wavelet coders, a 3-D SPIHT encoder [91] was included as a reference. An early version of this coder [89] has already established to beat the performance of a context-based octave zerotree coder [85]. The source code was made presented by the authors so it could be equipped with the proposed wavelet transform front-end.

The SPIHT implementation in this study uses balanced 3-Dspatial orientation trees. Therefore, the same number of recursive wavelet decompositions is necessary for all spatial orientations. If this is not respected, several tree nodes do not refer to or be linked with the same spatial location and, as a result, the dependencies among different tree-nodes are destroyed and, hence, the compression performance is reduced. Thus, a packet-based transform is not working to obtain a unitary transform with this embedded coding system. Therefore, the SPIHT coder was equipped with a no unitary transform. It is, however, worthwhile mentioning that solutions have been proposed utilizing unbalanced spatio-temporal orientation trees in the context of video coding [92].

The examined 3-D SPIHT algorithm [91] follows the same procedure as its 2-D homologous algorithm, with the exception that the states of the tree nodes—each embracing eight wavelet coefficients are encoded with a context-based arithmetic coding system during the significance pass. The selected context models are based on the significance of the individual node members, as well as on the state of their descendents. Consequently, for each node coefficient four state combinations are possible. In total 164 different context models are used.

#### d) Cube Splitting (CS)

The CS technique is derived from the 2-D SQP coder proposed in Section II-C. In the context of volumetric encoding, the SQP technique was comprehensive to a third dimension: from square splitting toward CS. CS is applied on the wavelet image in order to isolate smaller entities, i.e., sub cubes, possibly containing important wavelet coefficients.

During the first significance pass  $Sp^{\max}$ , the significance of the wavelet image (volume) W is tested for its highest bit plane  $p^{\max}$  with the significance operator  $\sigma^p$ . If  $\sigma^{p^{\max}}$  (W)=1, the wavelet image W is spliced in eight sub cubes (or octants),  $Q_b^{p^{\max}}(k^q, v^q/2), 1 \le q \le 8$ , with top-left coordinates

 $k^q = (k_1^q, k_2^q, k_3^q)$  and of size  $v^{\frac{q}{2}} = (v_1^{\frac{q}{2}}, v_2^{\frac{q}{2}}, v_3^{\frac{q}{2}})$ . The descendent "significant" cube (or cubes) is (are) then more spliced until the significant wavelet coefficients

 $\sigma^{p^{\text{max}}}$  (w(k)) = 1 are isolated. Thus, the significance pass  $Sp^{\text{max}}$  registers sub cubes and wavelet coefficients, newly identified as important, using a recursive tree structure of octants. The result is an octtree-structured description of the data meaning against a given threshold. As might be noticed, equal significance weights are given to all the branches. When a important coefficient is isolated, also its sign for which two code symbols are conserved is immediately encoded. When the complete bit-plane is encoded with the significance pass  $Sp^{\text{max}}$ , p is set to  $p^{\text{max}}$ -1 and the refinement pass R  $p^{\text{max}}$ -1 is initiated for this bit-plane, refining all coefficients marked as significant in the octtree.

Thereafter, the significance pass is restarted to update the octtree by identifying the new significant wavelet coefficients for the present bit-plane. During this stage, only the before non significant nodes, i.e.,  $\sigma^{p+1}(Q(k^q, v^q/2^j)) == 0, 0 < j \leq J$ , are checked for significance and the important ones, i.e.  $\sigma^{p+1}(Q(k^q, v^q/2^j)) == 1$ ,are unnoticed since the decoder already received this information. The described procedure is frequent, until the complete wavelet image W is encoded, i.e., p=0 or until the desired bit-rate is obtained.

То encode the generated symbols professionally, a context-based arithmetic encoder was integrated. The context model is simple. For the significance pass four context models are distinguished, namely one for the symbols generated at the intermediate cube nodes, one for the pixel nodes having no significant neighbors for the earlier threshold, one for the pixel nodes having at least one significant neighbor for the earlier threshold and finally one for encoding the sign of the isolated significant pixel nodes. Only two contexts are used for the refinement pass: one for the pixel nodes having no significant neighbors for the earlier threshold, one for the pixel nodes having at least one important neighbor for the previous threshold.

Other 2-D techniques, like NQS [84] and sub band block(SB) SPECK [59], have been proposed that use similar quadtree decomposition techniques. These coders divide the wavelet space in blocks and activate for each block disjointedly a quad-tree coding mechanism. In case of SB-SPECK, the block sizes are also depending on the sub band sizes, forcing each block to reside in one sub band. Each block is individually encoded and thereafter an EBCOT-alike rescheduling takes place to restore the scalability functionality. SB-SPECK was also partially extended to 3-D i.e., 3-D SB-SPECK coding [59] by equipping the coder with a 3-D wavelet transform front-end. The transform is activated on discrete chunks of slices [groups of frames (GOFs)], to maintain the accessibility of the data (typical GOF sizes are 8, 16, or 32 planes). The option is not implemented in the coders we designed. SB-SPECK does not use arithmetic encoding. However, the 3-D SB-SPECK coder delivers competitive results and we will refer to it whenever possible.

#### e) Three-Dimensional QT-L

The QT-L coder has also been extended toward 3-D coding. The octtrees corresponding to each bit-plane are constructed following a similar strategy as for the CS coder. However, the partitioning process is limited in such a way that once the volume of a node  $V_n = \prod_{j=1}^{3} v^q / 2^j, 0 < j \le J$ , becomes smaller than a predefined threshold  $V_{{\scriptscriptstyle th}}$  , the splitting process is stopped and the entropy coding of the coefficients within such а significant leaf node  $\sigma^{p}(Q(k^{q}, v^{q}/2^{j})) = 1$  is activated. Similar to the 2-D version, the octtrees are scanned using depth-first scanning. In addition, for any given node, the eight descendant nodes are scanned using a 3-D instantiation of the Morton-curve [81]. For every bitplane, the coding procedure consists of the non significance, importance and modification passes adapted for 3-Dcoding; also, for the maximum bit-plane, the coding process consists of the significance pass only. Notice that the sum number of neighbors  $N_{\rm tot}$  in (2) is set to 26 in 3-D coding.

#### f) 3-D CS-EBCOT

The CS-EBCOT coding [58] join the principles utilized in the CS coder with a 3-D instantiation of the EBCOT coder [43]. In the next paragraphs the interfacing of the CS coder with a edition of EBCOT modified to 3-D is discussed.

To begin with, the wavelet coefficients are separation EBCOT-wise in separate, uniformly sized cubes, called code-blocks. Normally, the first size of the code-blocks is 64x64x64 elements. Additional sizes (even different ones for every dimension) can be chosen, depending on the image characteristics and the request requirements. The coding module CS-EBCOT again consists of two major units, the Tier 1 and Tier 2 parts.

The Tier 1 of the proposed 3-D coding architecture is a hybrid module joins two coding techniques: CS and fractional bit-plane coding by context-based arithmetic encoding. The Tier 2 part is equal to the one used in the 2-D coding system.

1. CS: The CS pass S is resulting from the CS technique presented in Section III-D. In the proposed coding system, the CS is useful on the individual code-blocks in arrange to separate smaller entities, i.e., sub cubes, possibly containing major wavelet coefficients. The least sub cube size

that is supported is  $4x \ 4x \ 4$ .We will refer to these least sub cubes as the leaf nodes.

During the initial CS pass  $Sp_i^{\max}$ , the importance of code-block  $B_i$  is tested for its maximum bit-plane  $p_i^{\max}$  with the significance operator  $\sigma^p$ . If  $\sigma^{p_i^{\max}}(B_i) = 1$ , the code-block  $B_i$  is join until the important leaf nodes  $Q_b^{p_i^{\max}}(k^q, v^q/2^G)$  are isolated, where G state the highest total of CS levels. When every significant leaf nodes are isolated, the fractional bit-plane coding part is activated for the present bit-plane and only for the important leaf-nodes.

When the total bit-plane is encoded utilizing the fractional bit-plane coding,  $p_i$  is set to  $p_i^{\text{max}}$ -1 and the succeeding CS pass,  $s_i^{\text{max}}$ -1 is activated. The explain procedure is repetitive, until the total block is encoded, i.e.,  $p_i = 0$ . Due to the limited quantity of code-symbols and their allocation, arithmetic coding is not useful.

2. Fractional Bit-Plane Coding: The fractional bit-plane coder encodes just those leaf nodes that have been recognized as important throughout the CS pass. Three passes are defined per bit-plane like in the 2-D case: the importance transmission pass, the magnitude refinement pass and the normalization pass. Moreover, these coding passes call numerous coding operations (primitives), i.e., the ZC, SC, MR, and RLC primitives. These primitives enable the choice of suitable3-D situation models for the successive arithmetic coding or RLC stages. The chosen adaptive arithmetic encoder is based on an implementation by Said and Pearlman of the algorithm proposed by Witten et al. [45], which is equal to those utilized in the earlier state encoders.

The data exist in in each leaf-node is scanned applying as lice-by-slice scanning pattern. Inside one slice the pattern is equal to the JPEG2000 scanning: the vowels are read in-groups of four vertically associated voxels. When a total slice is stripe-wise processed, the subsequent segment is processed.

The fractional coding passes perform in an equal way as for the original EBCOT execution. However, the chosen neighborhood  $\Theta_k$  refers now to the twenty-six voxels approximately the voxel being coded (i.e., the immediate neighbors). For every bitplane, sequentially the significance propagation pass, the MR pass, the CS pass and the normalization pass are called, excluding for the first bit-plane where the initial two passes are discarded.

3. Coding Primitives: As for EBCOT, four coding primitives are defined to support the encoding procedure in the dissimilar coding passes: the ZC

primitive, the RLC primitive, the SC, and the MR primitive. For arithmetic encoding, the contextmodel choice is based on the condition of the adjacent voxels of the voxel being encoded, i.e., the preferred neighborhood and the sub band type in which the voxel is situated. The preferred neighborhood  $\Theta_{\nu}$  is separated in 7 orthogonal subsets according to their position to the voxel [58], [80]. Every coding primitive has got its individual look-up table to identify the probability model that has to be utilized by the arithmetic coder for a identified situation state [58], [80]. Additionally, we have to state that the complexity of this part of the coding engine increases heavily compared with the original 2-D implementations, appropriate to the enlarged preferred neighborhood (from 8 to 26 neighbors)and, consequently, the augmented complexity of the look-up tables[58].

4. Tier 2—Layer Formation: The followed process, i.e., PCRD optimization [43], is equal to the original one. However, we have to state one feature that is of key importance. The PCRD routine allocates compensating for the fact that a no unitary transform has been used. By accurate the calculated distortions for each pass  $n_i$  with a scaling factor  $\zeta b_i$ , the coding method will execute as if a unitary transform was used (or approximated when using integer powers of  $\sqrt{2}$ ).Hence, the distortion will be currently described by:

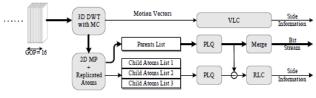
$$D_{i}^{n_{i}} = \zeta b_{i} \sum_{k \in B_{i}} (S_{i}^{n_{i}} t[k] - s_{i}[k])^{2}$$

Where  $s_i[k]$  indicate the magnitude of element

k in code-block  $B_i$  and  $S_i^{n_i}[k]$  provide the quantized illustration of that element connected with truncation point  $n_i$ . This improvement allows support for a unitary transform without obstructing the possibility of lossless coding, a difficulty that does occur with classical zero tree-based coders. The original 2-D algorithms maintain multiple components (e.g., color), but this feature is not retained in the future 3-D implementation. Hence, only images (volumes) gray-scale are supported. Nevertheless, the dissimilar bit-stream chunks are currently grouped into separate packets, every packet contributing to one quality layer and one resolution level. The code-block addition information is again encoded by means of the tag-tree concept. The only alter that has been complete was extending the tag-tree idea to the third dimension, i.e., moving from a quad tree structure to an octtree structure.

#### VI. CURRENT STATE OF THE ART

Matching Pursuits (MP) has been recognized as an useful technique of over entire transform coding for 2D images, originally for 2D motion compensated residuals by Neff and Zakhor [12], and additional recently extended to still images by the application of a spatial wavelet transform [13] and the utilize of additional effective dictionaries and embedded coding [14]. The addition of MP to three dimensions is natural, as has been established [15]. However MP is a computationally strong method and the utilize of 3D bases increases this cost significantly. In this context Yuan Yuan[1] proposed a novel 3D WAVELET VIDEO CODING WITH REPLICATED MATCHING PURSUITS. The model can be briefed in figure 1 and description follows



*Figure 1 :* 3D Replicated Matching Pursuits Scheme for Video Coding

Replicated MP coding is shown in Figure 1. For a group of 16 input frames, the motion compensated 3D DWT is implemented subsequent the scheme of Taubman and Secker [5A, 6A]. The related motion vectors are coded as side information by a Variable Length Code (VLC). This movement overhead affects the low down bit rate performance of all video coded by related schemes. The MP algorithm is useful to the 3D Group of Temporal Pictures (GOTP) using the Intra 8 Codebooks of Monro [14]. The benefit of 3D RMP is that the collection of Child atoms related with a Parent can provide up to three times the image power of the Parent alone, but expenses on average less than 30% of the bits to code in this implementation. MP is a computationally costly algorithm, with the best inner product over the entire data set with all bases in the dictionary essential for selecting each atom.

Sun et al[2] proposed a novel content adaptive rate-distortion optimization scheme, which might effectively differentiate texture region, edge region and flat region by means of directional field technique. There have been numerous efforts in the past trying to include perceptual procedures into video encoding. In [16], the focus was mostly on determination of suitable quantization steps with sub-band Just Noticeable Distortion (JND). In recent H.264/MPEG-4 advanced video coding standard, maximum coding efficiency is achieved by introducing the rate distortion optimization (RDO) procedure to provide the best coding outcome by maximizing image quality and minimizing data rate at the same time. In [17], a new adaptive RDO scheme has been proposed which exploits motion and texture masking property to correct the Lagrangian multiplier and achieves generally bit rate reduction by allowing additional distortion in the less noticeable background random texture area. The RDO technique is also an significant part in wavelet-based scalable video coding (SVC) scheme that is presently under examination by MPEG-21, Part 13 [18]. The SVC scheme requires an embedded bit stream to be formed from which bit streams with different bit rate, resolution and frame rate could be extracted with reasonably fine quality. Here in this job sun et al proposed an effective directional field based visual significance map in the context of capturing the features associated to pre attentive processing such as edges and curves. Based on the visual significance map, the regions with additional preattentive features are likely to get more distortionreduction by assigning a smaller Lagrangian multiplier. Rate balance was also measured as a factor and attempted to attain by assigning a relatively larger Lagrangian multiplier to the random texture area so that additional distortion is permitted without noticeable visual degradation to the image. Since HVS is also sensitive to distortions in flat regions, a little Lagrangian multiplier was also used for flat regions.

Seran et al[3] proposed a 3D BASED VIDEO CODING to carry out the two-dimensional spatial filtering initial and then perform motion-compensated temporal filtering by lifting in the Over total Discrete Wavelet Transform domain. In practice essentially the three-dimensional wavelet decomposition can be performed in two ways: two-dimensional spatial filtering followed by temporal filtering (2D+t) [19, 20, 21] or, temporal filtering followed by two-dimensional spatial filtering (t+2D) [22, 23, 24]. In this work, Seran et al[3] proposed a new temporal filter set to reduce delay in 3D wavelet based video coding, that attempted to increase performance at par with existing longer filters. In this proposal the filter set haven't include any boundary effects at the group of frames (GOF). The length of the GOF can differ from five to any number of frames depending on interruption requirements. This proposed model also illustrated a novel technique of assigning priorities to temporal sub-bands at dissimilar levels to manage distortion fluctuation inside a GOF.

Mavlankar et al[4] considered a multiple description (MD) video coding scheme based on the motion compensated (MC) lifted wavelet transform, which is to carry out the temporal decomposition of a collection of pictures and then make multiple descriptions for every temporally transformed frame. The benefit of basing MD video coding on motioncompensated lifted 3D wavelet decomposition is that it does not need any difference control similar to in a hybrid codec which was earlier achieved by distribution drift compensation data. Earlier to this proposal an important number of MD coding schemes for video have been proposed (e.g., [25], [26], [27]). Most of the proposals for MD video believe the usual hybrid video coding structure with motion compensated prediction and DCT as their major building blocks. In difference the model proposed by Mavlankar et al a multiple explanation video coding scheme that is based on a video encoder that uses the recently emerging motion compensated lifted 3D wavelet change as its basis [27], [28], [29], [30]. Since recursive temporal prediction is restoring by a motion-compensated transform, it also remove the dependency quantization framework which is an inherent part of hybrid video codecs. In usual hybrid codecs quantization is fixed in the recursive prediction loop, whereas in 3- D wavelet codecs quantization and spatial encoding are applied following the temporal decorrelation of a group of pictures (GOP). Figure 2 shows our proposed MD video codec which is based on MC lifted wavelet change. This scheme expands the Drift Compensation Multiple Description Video Codec (DC MDVC) proposed for a hybrid video codec in [31] to MC-lifted 3-D wavelet coding.

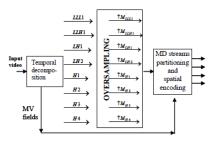


Figure 2 : Overview of the proposed MD coding scheme

A huge benefit of employing the 3-D wavelet scheme is that there is not require to send additional drift compensation data. DC MDVC has to send the drift compensation data to compensate for difference between the recursive predictions loops at the encoder and the decoder in case of loss. In fact the number of drift compensation streams rose exponentially with the total amount of descriptions. It is (2N-2), where N is the total number of descriptions. This can be simply understood by considering that a drift compensation stream has to be formed for each scenario excluding when all or no descriptions are received.

Seran et al[5] focused on the difficulty of controlling unpredictable variation of distortion in 3D coders, which aimed at exploring the MCTF filter properties and we present an entire analysis of the filter and mathematical derivations. The temporal wavelet filter properties are recognized to be a main factor contributing to distortion variation. The problem of controlling the temporal distortion fluctuations has been addressed in a little design [32], [33]. In [32], distortion variation control is considered for the bi-directional unconstrained motion compensated temporal filtering and the distortion in the decoded frame is expressed as a function of the distortions in the reference frames at

the equal temporal level. In [33], the association among the distortion in temporal wavelet subbands and the reconstructed frames are study for the modified 5/3 filter (ignoring sqrt(2)). Based on the association, a distortion ratio model is theoretically developed and an easy rate control algorithm is used to place priorities for the temporal subbands according to the distortion ratio.

The association among the distortion in the reconstructed frames and the filter coefficients study by seran et al[5]. On this foundation, scaling coefficients for the filter are calculated to control the distortion fluctuation. In this circumstance the model proposed by seran et al [5] considered the mainly popular biorthogonal 5/3 filter and quoted that this model can be directly extended for additional longer filters.

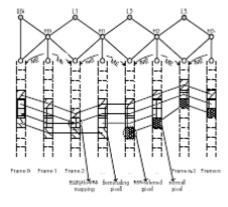
Zefeng Ni et al[6] proposed and develops a 3D wavelet codec based on MCTF and JPEG2000 for a novel advance for stable quality aimed bit allocation between T-bands for the applications of adaptive stored video streaming. The reconstructed structure is divided into dissimilar groups according to the types of their related temporal bands. We suggest an estimated mathematical model to describe the relationship between the T-band distortions and the distortions of the reconstructed frames. Since we consider stored video in this investigation, we can offline produce the model parameters. Throughout the online transmission period, given the existing network bandwidth, we initial perform the conventional JPEG2000-like optimum truncation. After that, a two-step process is used for reducing the PSNR fluctuation, where the fundamental idea is to alter the energy gains to balance the dissimilar contributions from diverse types of T-bands and then more or less equally allocate distortion among the Tbands at the identical level. Along with model proposed by Zefeng Ni et al[6], little of the articles[34], [35], [36], [37] in latest literature have looking into that how to assign bits to every temporal band (T-band) so that definite degree of stable quality can be achieved. Although the difficulty of stable quality aimed bit distribution has been well calculated for conventional hybrid codecs [38, 39], it is still an open question for MCTF based codecs. In the popular MC-EZBC codec [34], the authors suggest to discontinue bit plane scanning of every the GOPs at the identical fractional bit plane. However, it can only help accomplish similar distortion among T-bands, which does not lead to stable quality in reconstructed frames. This is because the distortion in T-bands propagates unequally into reconstructed frames, which is hard to model mathematically. In [35, 36], in order to smooth PSNR performance, an optimized quantization step is obtained by analyzing the motion performance throughout the temporal filtering. In [37], the authors suggest to utilize an adaptive update step for the temporal filtering. Compared with conventional implementation, these technique indeed help dropping the serious PSNR fluctuation in reconstructed frames.

However, they did not clearly address the problem of bit allocation among T-bands.

Yongjian Man et al<sup>[7]</sup> presented a new 3D-WT algorithm for video coding, which can considerably decrease the processing memory and attain a high coding presentation. This paper is organized as follows: in section 2, we explain the traditional 3D-WT algorithm and its shortage. According to the decomposition structure, the model proposed is dissimilar from traditional 3D wavelet coding. As the input sequence for the proposed new 3D wavelet decomposition structure is based on numerous groups, for every group, only the little frequency frame is remained in processing memory for the final temporal decomposition. According to the analysis of the proposed model, only the high frequency frames are exported following temporal decomposition, while the low frequency frame remains in memory for every group.

Yu Liu et al [8] projected an expansion to the lifting-based activity threading procedure from the frame-based coding to the object-based coding, involved by the inimitable compensation of the objectbased coding that do not survive in other coding proposal. Object-based coding allows convenience and impressionability of object within a video series, and allows the organization of video content to endure the process of attainment, editing and allocation, which is useful for content-based search and recovery in MPEG-7. As an substitute to established video coding normal, 3D wavelet video coding has conventional much awareness recently. A main benefit of 3D wavelet video coding is that it can supply complete spatio-temporalquality scalability with non-redundant 3D sub-band disintegration. In 3D wavelet video coding, action reimbursement is regularly integrated into the sequential wavelet make over to accomplish competent coding routine, leading to a class of algorithms normally called as activity rewarded sequential sieve (MCTF). Previous to this occupation, Xu et al. [40] projected a motion yarn (MT) technique that employs longer wavelet filter to develop the long-term association across border along movement route. The aim of MT is to shape as lots of long clothes as likely since too much small gear will considerably augment the figure of reproduction borders. Luo et al. [41] planned an advanced MT method to reduce the numeral of many-to-one drawing pixels and non-referred pixels in the innovative MT. though the trouble of frontier effects grounds by the truncation of many-to-one drawing case in the inventive MT can be well explain by the advanced MT, the non referred pixel case, which is allocate to use the action vector from neighboring movement strand in the higher MT, is not well resolve since the allocate movement vector may be not precise or even mistaken for nonreferred pixel and may source some poverty on coding presentation.

Due to the boundary effects in spatial and sequential wavelet reconstruction, these artificial limitations will mortify greatly the coding recital. Consequently, it is improved to solve the problem of border line property, which survive in spatial and sequential makeover of object-based system concurrently, in a unified structure.



*Fig. 3 :* The object-based motion threading technique with lifting structure

To obtain more accurate proposal trajectory of each pixel and the functionality of object-based coding with illogical province of support, the object-based action threading with the exciting structure is projected, as shown in Fig.3. The object-based motion yarn using exciting structure aims to condense boundary possessions of synthetically terminating/emerging outfit in the preceding MT procedure.

Chen-Wei Deng et al[9] projected a new structure for scalable video convention. A mesh-based activity inference model is calculated and functional in this proposal, which engender a continuous proposal field. The earthly relationship is broken by Barbell thrilling [42]. In 3D wavelet video coding, wavelet change are exploit temporally across frames, and straight and up and down with each border, correspondingly. [34], [44] perform motioncompensated sequential strain (MCTF) in the inventive spatial field pursue by a spatial change on each sequential sub band. This is typically denote as a t+2D scheme. This preceding work was principally in the framework of a slab activity model, which poorly represent complex motion in real video sequences. Y. Andreopoulos et al [45] pertain spatial convert before the sequential one, which is typically referred to as 2D+t method. 2D+t can answer the spatial scalability concert issues of t+2D, but it experience from the shift-variant natural history of the DWT. [46] is a 2D+t+2D scheme. In such structure, spatial balance description of video succession are acquire opening from the superior resolution, while the coding system include inter scale prediction (ISP) machine in organize to develop the multiscale symbol idleness [8A]. These formats have

archetypal inadequacy of over absolute change [47], the regulations competence is hard to achieve.

With reference to the limits claimed touching the representation [34], [44], [45], [46], [48] Chen-Wei Deng et al [9] projected a novel scalable 3D wavelet video coding support, which is a t+2D scheme. A mesh-based motion model is incorporated into the planned scaffold, which is helpful to perk up solidity concert. In totaling, due to the fact that different temporal sub bands have different characteristics, two different wavelet system algorithms were planned for the low and high-pass activist sub bands, correspondingly.

Ke Xu[10] planned a novel scheme that is forced by circulated cause system to correct the errors of momentous division in 3D wavelet video tributary. In this projected replica Extra in rank is engender by Wyner-Ziv codec and is throw to the decoder. While errors occur, the relevant parts in the same frames from EZBC decoder are used as side in sequence to decipher the Wyner-Ziv bits to construct a refined substitution of the dishonored ones. to terminate the rest parts of these border are shared with superior parts from the Wyner-Ziv portrayal to yield a correct succession.

Later to this application Sehoon Yea et al[49]planned a motion reimbursement entrenched zero-block coder (MC-EZBC) and shows that MC-EZBC can achieve higher concert evaluate with H.264 at a high rate. Furthermore, it is easy to extract the various rates, declaration and eminence succession from the raw video succession in decoder. However, MC-EZBC is still disposed to errors and encodes the variously noteworthy parts uniformly. A number of scheme, such as Auto Repeat demand (ARQ) or Forward Error improvement codes (FEC) or grouping of both can be used to truthful errors in conjecture.

FEC methods are shaped by adding together extra ensure bits in the facade of the packet to check the error bits and moderate the likelihood of error. While, they can neither acceptable the fracture errors nor inefficiently adapt to the guide. ARQ scheme can progress the quality of decode movies by a response guide to send the lost packet again. Nevertheless, it is very disposed to time stoppage. Time delay will be engorged if the waterway is concerned by noises more commonly.

Both FEC and ARQ applied in established codec can be engaged in MC-EZBC shielding the bit tributary supposedly. Nevertheless, the decoder of MC-EZBC can work well within at least 4 frames indoors, so the importance is more time delay. Recently, Wyner-Ziv coding has been planned on error rigidity [50], [51]. They encode the source using established codec and Wyner-Ziv codec, which is used to suitable the errors in other report in that order. Ke Xu[10]extensive the same idea to 3D wavelet coding.

Nobuhara et al[11] planned A video coding method using max-plus algebra based three dimensional wavelet convert (3DMP-Wavelets) . The recompense of MP wavelets those definite by max-plus algebra are

- a. Since no hovering point calculation is requisite, the addition speed of MP-Wavelets is high.
- b. Since no duplication operation is required, MPWavelets are hardware implementation oriented.
- c. Since the problem of round-off errors is completely eliminated, MP-Wavelets are appropriate for digital watermarking, i.e., they be applied to exclusive rights fortification.

Hence 3DMP-Wavelets computational charge is very small since no balanced point calculations are done, max is computationally less expensive than the sum, and sum is computationally less costly than the product. The projected 3DMP-wavelets do better than a three dimensional linear wavelet in terms of velocity of the computation. Also, 3DMP-wavelets are hardware accomplishment oriented. This stimulates their study, mostly in view of some watch applications. It is easy to observe that in observation request it is important to have cheap strategy that can program a video succession at a low cost with rational eminence.

## VII. Conclusion

This paper has provided a picture of various tools that have been designed in recent literature for visual data compression, in particular 3D wavelet transformation and coding. It has focused on multiresolution representations with the use of the wavelet transform and its extensions to handle motion and ROI in visual sequences. For image compression, WT based approaches are showing quite competitive performance due to the energy compaction ability of the WT to handle piecewise polynomials that are known to well describe many natural images. In video sequences, the adequacy of such model falls apart unless a precise alignment of moving object trajectories can be achieved. This might remain only a challenge, since as for any segmentation problem; it is difficult to achieve it in a robust fashion, due to the complex information modeling which is often necessary. Most of the models studied in this paper are not generalized in the context of quality artifacts. Hence it remain an issue and providing research scope to identify the effective lifting schemes and 3D wavelet coding models in the context various contextual parameters such as region of interest, motion in streaming visuals and quality artifacts such as spatial scalability and temporal scalability.

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## Performance Analysis of Intensity Averaging by Anisotropic diffusion Method for MRI Denoising Corrupted by Random Noise By Ms. Ami vibhakar, Mukesh Tiwari, Jaikaran singh & Sanjay rathore

Ragiv Gandhi Technical University

*Abstract* - The two parameters which plays important role in MRI (magnetic resonance imaging), acquired by various imaging modalities are Feature extraction and object recognition. These operations will become difficult if the images are corrupted with noise. Noise in MR image is always random type of noise. This noise will change the value of amplitude and phase of each pixel in MR image. Due to this, MR image gets corrupted and we cannot make perfect diagnostic for a body. So noise removal is essential task for perfect diagnostic. There are different approaches for noise reduction, each of which has its own advantages and limitation. MRI denoising is a difficult task as fine details in medical image containing diagnostic information should not be removed during noise removal process. In this paper, we are representing an algorithm for MRI denoising in which we are using iterations and Gaussian blurring for amplitude reconstruction and image fusion, anisotropic diffusion and FFT for phase reconstruction. We are using the PSNR(Peak signal to noise ration), MSE(Mean square error) and RMSE(Root mean square error) as performance matrices to measure the quality of denoised MRI .The final result shows that this method is effectively removing the noise while preserving the edge and fine information in the images.

Keywords : MRI, Random noise, iteration, Gaussian blur, image fusion, anisotropic diffusion, PSNR, MSE, Image denoising.

GJCST-F Classification: 1.4.3



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# Performance Analysis of Intensity Averaging by Anisotropic diffusion Method for MRI Denoising Corrupted by Random Noise

Ms. Ami vibhakar<sup>a</sup>, Mukesh Tiwari<sup>o</sup>, Jaikaran singh<sup>o</sup> & Sanjay rathore<sup> $\omega$ </sup>

Abstract - The two parameters which plays important role in MRI (magnetic resonance imaging), acquired by various imaging modalities are Feature extraction and object recognition. These operations will become difficult if the images are corrupted with noise. Noise in MR image is always random type of noise. This noise will change the value of amplitude and phase of each pixel in MR image. Due to this, MR image gets corrupted and we cannot make perfect diagnostic for a body. So noise removal is essential task for perfect diagnostic. There are different approaches for noise reduction, each of which has its own advantages and limitation. MRI denoising is a difficult task as fine details in medical image containing diagnostic information should not be removed during noise removal process. In this paper, we are representing an algorithm for MRI denoising in which we are using iterations and Gaussian blurring for amplitude reconstruction and image fusion, anisotropic diffusion and FFT for phase reconstruction. We are using the PSNR(Peak signal to noise ration), MSE(Mean square error) and RMSE(Root mean square error) as performance matrices to measure the quality of denoised MRI .The final result shows that this method is effectively removing the noise while preserving the edge and fine information in the images.

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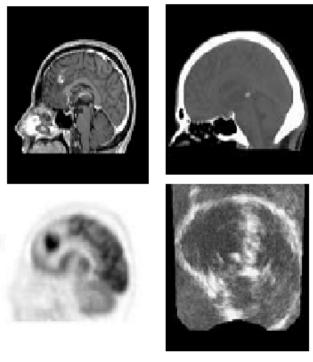
#### I. INTRODUCTION

RI Stands for Magnetic Resonance Imaging; once called Nuclear Magnetic Resonance Imaging. The "Nuclear" was dropped off about 15 years ago because of fears that people would think there was something radioactive involved, which there is not. MRI is a way of getting pictures of various parts of your body without the use of x-rays, unlike regular x-rays pictures and CAT scans. A MRI scanner consists of a large and very strong magnet in which the patient lies. A radio wave antenna is used to send signals to the body and then receive signals back. These returning signals are converted into pictures by a computer attached to the scanner. Pictures of almost any part of your body can be obtained at almost any particular angle.

Medical information, acquired from MRI and composed of clinical data, images and other

physiological signals, has become an essential part of a patient's care for diagnosis in medical field. Over the past three decades, there is a vast development in information technology (IT) & Medical Instrumentation, which has improved the level of medical imaging. This development are Computed Tomography (CT), Magnetic Resonance Imaging(MRI), the different digital radiological processes for vascular, cardiovascular and contrast imaging, mammography, diagnostic ultrasound imaging, nuclear medical imaging with Single Photon.

Emission Computed Tomography (SPECT) and Positron Emission Tomography (PET). All these methods generate good quality of medical image [1] and each has its own specific features corresponding to the physical and physiological phenomena studied, as shown in "Fig.1



*Figure 1 :* Sagittal slices of the brain by different imaging modalities: a) magnetic resonance imaging (MRI), b) computed tomography (CT), c) positron emission tomography (PET), d) ultrasound (US)

All medical images contain Random noise. The presence of noise gives an image a mottled, grainy,

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textured, or snowy appearance. Image noise comes from a variety of sources. No imaging method is free of noise, but noise is much more prevalent in certain types of imaging procedures than in others. Noise is also significant in MRI (Medical Resonance Imaging), CT, and ultrasound imaging. In comparison to these, radiography produces images with the least noise. Fluoroscopic images are slightly noisier than radiographic images. The presence of noise degrades the image quality and decreases visibility of lower contrast image. So there is a need for noise removal technique to improve the image quality and to recover the fine details of image which is required for perfect diagnostic. This paper is divided into seven sections. Section one gives idea about MRI and denoising. Section two shows a literature survey .Section three defines implementation of algorithm .Section four and five gives idea about Gaussian blur and anisotropic diffusion. Section six defines pproposed algorithm for denoising while section seven is conclusion

#### II. Related Work

Various algorithms for image denoising are discussed in [2]. The de-noising of Magnetic Resonance Images using wave atom shrinkage is proposed in [3] and also proved that this approach achieves a better SNR compared to wavelet and curvelet shrinkages. A NL-Denoising method for Rician noise reduction is proposed in [4 & 5].In [6], Total Variation Wavelet-Based technique is used to remove a noise from MR image. The method to improve image quality using adaptive threshold based on contourlet transform is given in [7]. A new filter to reduce random noise in multicomponent MR images by spatially averaging similar pixels and a local principal component analysis decomposition using information from all available image components to perform the denoising process is proposed in [8]. An estimator using a priori information for devising a single dimensional noise cancellation for the variance of the thermal noise in magnetic resonance imaging (MRI) systems called ML estimator has been proposed in [9]. A noise removal technique using 4th order PDE is introduced in [10] to reduce noise in MRI images. A phase error estimation scheme based on iteratively applying a series of non-linear filters each used to modify the estimate into greater agreement with one piece of knowledge, until the output converges to a stable estimate is introduced in [11].

#### III. Implementation

Fig. 3 shows the block diagram, gives general idea for MRI denoising using intensity averaging method.

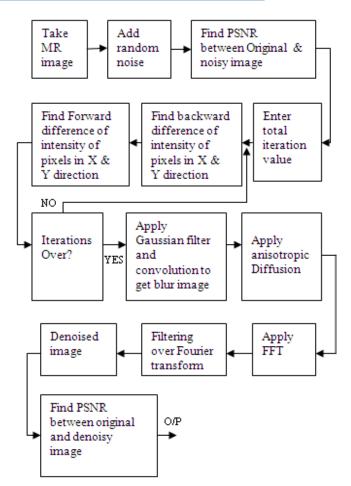


Fig. 3 : Block diagram of intensity averaging algorithm

In proposed algorithm we have taken the image of [fig.2] for evaluating our method. First we will apply amplitude correction on noisy MR image by finding forward and backward difference of intensity of pixels in X and Y direction. This gives average type of value to each pixel and then image is blurred by Gaussian filter and convolution. After completion of this amplitude correction, we apply a phase correction algorithm. Here, we are splitting amplitude corrected image into its red, green and blue band and then we are rotating each band by appropriate amount to correct the phase of MR image. After this, we are applying anisotropic diffusion and FFT to remove the noise from image.



Fig. 2 : MRI of knee

### IV. GAUSSIAN BLUR

Gaussian blur is also known as Gaussian smoothing used to blur (smooth) the image. Typically it is used to reduce a random noise from the image. Mathemetically, Gaussian blur is equivalent to applying a convolution between image and Gaussian function [12, 13]. Gaussian distribution in 1-D is given as,

$$G(x) = \frac{1}{\sqrt{2\pi}} e^{-\frac{x^2}{2\sigma^2}}$$

Where  $\sigma$  is the standard deviation of distribution. In 2-D, an isotropic Gaussian has the form,

$$G(x, y) = \frac{1}{2\pi\sigma^2} e^{-\frac{x^2 + y^2}{2\sigma^2}}$$

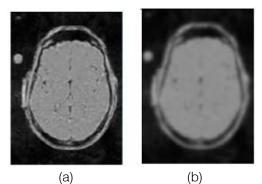
Here. we are producing a discrete approximation of the Gaussian function before we perform the convolution as image is considered as a collection of pixels. Ideally we require an infinitely large convolution kernel because the Gaussian distribution is non-zero everywhere, but in practice we can truncate the kernel as Gaussian distribution in it is effectively zero, more than about three standard deviations from the mean. The degree of smoothing depends on the value of standard deviation. The Gaussian outputs a `weighted average' of each pixel's neighborhood, with the average weighted towards the value of the central pixels. So,

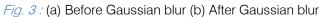
$$I(x,y) = IO(x,y) * G(x,y)$$
(1)

I(X,Y) = Gaussian blurred image I0 (x,y) = Noisy image G(x,y,t) = Gaussian filter function

This is in contrast to the mean filter's uniformly weighted average.[14] Because of this, a Gaussian provides gentler smoothing and preserves edges better than a similarly sized mean filter. The main problem with Gaussian filter is,

- Loss of fine detail
- Smoothing across boundaries as shown in fig.3.





This problem can be solved by anisotropic diffusion as discussed below.

### V. Image Fusion and Anisotropic Diffusion

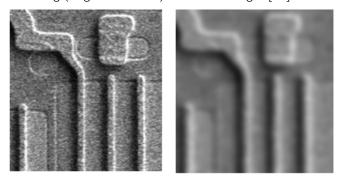
Image fusion describes the concept of combining multiple images into one image which gives more information compared to individual one [15]. Linear diffusion provides over smoothing of image as shown in fig. 3, we will use non-linear smoothing in which each pixel is treated with varying intensity depending on its neighboring value. In general,

if (x,y) is a part of an edge  $\rightarrow$  apply little smoothing if not a part of an edge  $\rightarrow$  apply full smoothing

This idea can be implemented by using a gradient function as given below.

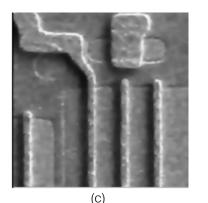
$$Grad(I) = \left(\frac{\partial}{\partial x}, \frac{\partial}{\partial y}\right)$$

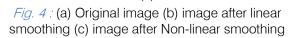
So non linear smoothing gives good intraregion smoothing as well as doesn't do much with interregion smoothing (edges and lines) as shown in fig.4 [16]



(a)

(b)

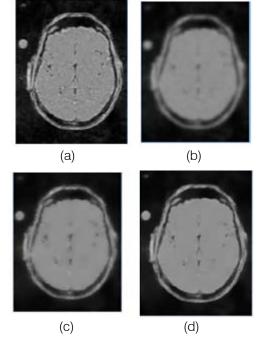




This problem can be solved with anisotropic diffusion [17] when equation no.1 can be viewed as heat equation as shown below,

$$It = \Delta I = (Ixx,Iyy)$$

The matter in an image is not heat, but brightness level. So, an image could be generalized to be a surface, where bright spots are "hot" and dark spots are "cold". So the idea is to use a varying size of kernel. Comparison of linear, non-linear and anisotropic diffusion is shown in fig.5



*Fig. 5 :* (a) original image (b) image after linear fusion (c) image after non-linear fusion (d) image after anisotropic fusion

### VI. Proposed Algorithm

- 1. I/P Image.
- 2. Add Random noise
- 3. Find PSNR between Original and Noisy Image.
- 4. Steps for applying Magnitude Reconstruction using iteration method on MR image:
  - a. Enter iteration value.
  - b. Find the backward difference of intensity of pixels in X and Y direction till the iteration ends.
  - c. Find the forward difference of intensity of pixels in X and Y direction till the iteration ends.
  - d. Find PSNR between Original and Denoisy Image after iteration process [psnr2].
  - e. Apply Gaussian filters to blur the image.
  - f. Perform convolution.
  - g. Find PSNR between Original and Denoisy image after Gaussian blurring. [psnr3].
- 5. Steps for applying Phase Reconstruction on Noisy MR image.
  - a. Apply anisotropic diffusion.
  - b. Find PSNR between Original and Denoisy image after iteration Process for Smoothing [psnr4]
  - c. Apply FFT on Diffused Image.
  - d. Perform the Filtering over Fourier transform.
  - e. Find PSNR between Original and Denoisy image after FFT on filtering [psnr5].
  - f. Apply Image Sharpening Using Filtering.
- 6. Denoised Image

Now, we will evaluate the algorithm by taking different values of iterations (A).





Fig. 6: (a) Input image

(b) Noisy input image

Now apply different values of "A" (no. of iterations) on noisy MR image as shown in fig.6 (b). As the value of "A" increases, we are getting more and more noise removal from noisy image as shown in fig. (7), (8) and (9). We are taking following parameters to evaluate the algorithm.

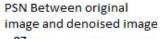
- Psnr2:- PSNR between Original and Denoisy image after iteration process
- psnr3:- PSNR between Original and Denoisy image after Gaussian blurring Image.
- Psnr4:- PSNR between Original and Denoisy image after iteration process for Smoothing using anisotropic diffusion.
- Psnr5:- PSNR between Original and Denoisy image After FFT on filtering
- Mse2:-Difference between Original and Denoisy image after iteration process
- Mse3:-Difference between Original and Denoisy image after Gaussian blurring Image.
- Mse4:- Difference between Original and Denoisy image after iteration process for Smoothing using anisotropic diffusion.
- Mse5:-Difference between Original and Denoisy image After FFT on filtering.

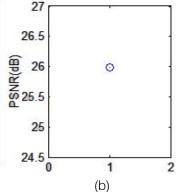
### FOR A=10:-

### Denoised image



(a)





### Denoised image after gaussian Blur



(C)

Denoised omage after iteration process for smoothing





Denoised image after FFT on filtering



(g)

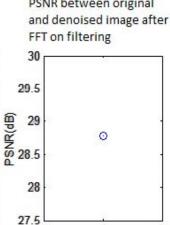
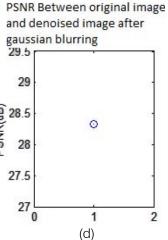
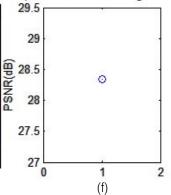


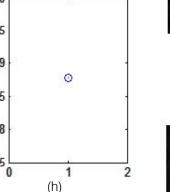
Fig. 7: Denoised image after (a) iterations (c) Gaussian blur (e)iteration process for smoothing using anisotropic diffusion (g) FFT on filtering



PSNR between original and denoised image after iteratior process for smoothing



PSNR between original



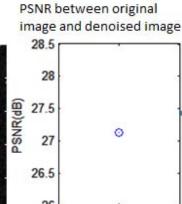
### Table 1 : Obtained Result for A=10

| Psnr2=25.98 | Mse2=15.51 |
|-------------|------------|
| Psnr3=28.35 | Mse3=17.86 |
| Psnr4=24.37 | Mse4=17.95 |
| Psnr5=28.81 | Mse5=18.2  |

### FOR A=15:-

Denoised image







Denoised image after gaussian Blur



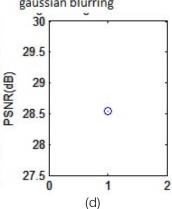
(C)

Denoised omage after iteration process for smoothing

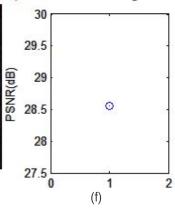


(e)

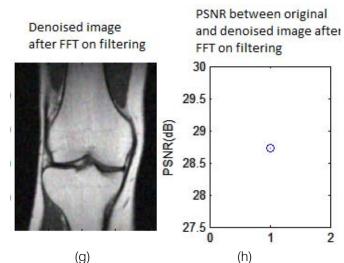
26 0 1 2 (b) PSNR Between original image and denoised image after gaussian blurring



PSNR between original and denoised image after iteration process for smoothing







*Fig. 8 :* Denoised image after (a) iterations (c) Gaussian blur (e)iteration process for smoothing using anisotropic diffusion (g) FFT on filtering

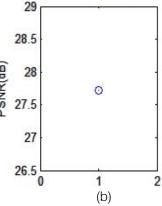
*Table 2 :* Obtained Result for A=15

| Psnr2=27.10 | Mse2=16.63 |
|-------------|------------|
| Psnr3=28.53 | Mse3=18.06 |
| Psnr4=27.60 | Mse4=18.07 |
| Psnr5=28.73 | Mse5=18.25 |

FOR A=20:-



PSNR between original and denoised image

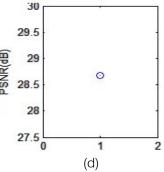


Denoised image after gaussian Blur



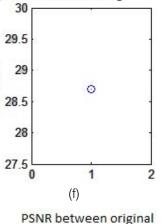
(C)

PSNR Between original image and denoised image after gaussian blurring



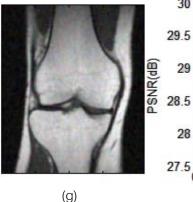
Denoised omage after iteration process for smoothing PSNR between original and denoised image after iteration process for smoothing

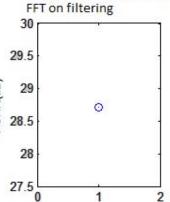




Denoised image after FFT on filtering

(e)





(h)

and denoised image after

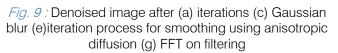


Table 3 : Obtained Result for A=20

| Psnr2=27.75 | Mse2=17.28 |
|-------------|------------|
| Psnr3=28.69 | Mse3=18.22 |
| Psnr4=28.70 | Mse4=18.23 |
| Psnr5=28.75 | Mse5=18.29 |

### VII. Conclusion

From the above result we conclude that, our algorithm is efficiently removing the noise from MR image. As number of iterations increases ("A"), we are getting more and more improved image. Experimental results show that, we are getting good Result in terms of PSNR and image quality. This algorithm is capable of removing noise from images and at the same time it is preserving fine details of images too. We also conclude that, for large value of iteration (A>25), increment in PSNR is less compared to small values of iterations (A<25).

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## Robust Watermarking Method for Color Images Using DCT Coefficients of Watermark

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*Abstract* - Digital technologies are playing a vital role in the present communication system. This paper presents a robust and secure watermarking method to protect the copyright information of multimedia objects. In the proposed method, Discrete Wavelet Transform and Discrete Cosine Transform are applied on the cover image and then Discrete Cosine Transform coefficients of watermark image are embedded into transformed cover image. The experimental result shows the performance evaluation of the proposed method by the quality metrics as PSNR for watermarked image and NC for extracted watermark image and we have compared the results with the existing transformation methods in frequency domain based on attacks.

Keywords : Discrete Wavelet Transform (DWT);Discrete Cosine Transform (DCT);PSNR(Peak Signal to Noise Ratio);NC (Normalized Correlation)

GJCST-F Classification: 1.4.0



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### Robust Watermarking Method for Color Images Using DCT Coefficients of Watermark

R.Eswaraiah<sup> a</sup> & E.Sreenivasa Reddy<sup> o</sup>

*Abstract* - Digital technologies are playing a vital role in the present communication system. This paper presents a robust and secure watermarking method to protect the copyright information of multimedia objects. In the proposed method, Discrete Wavelet Transform and Discrete Cosine Transform are applied on the cover image and then Discrete Cosine Transform coefficients of watermark image are embedded into transformed cover image. The experimental result shows the performance evaluation of the proposed method by the quality metrics as PSNR for watermarked image and NC for extracted watermark image and we have compared the results with the existing transformation methods in frequency domain based on attacks.

*Keywords : Discrete Wavelet Transform (DWT);Discrete Cosine Transform (DCT);PSNR(Peak Signal to Noise Ratio);NC (Normalized Correlation)* 

### I. INTRODUCTION

ow days the rapid growth of the technology is a cause for the requirement of the protection of copyright information from unauthorized access. Hence an advanced authentication method is essential to the data security. Digital watermarking is one such method that offers data security. At present various kinds of digital watermarking methods are in use to protect the information from unauthorized access. Any security method come under either spatial [1] like LSB or frequency domain [2] like Discrete Wavelet Transform (DWT) [5], Discrete Cosine Transform (DCT) [3], Discrete Fourier Transform (DFT) [6] and the combination of them [7, 8] as well as any method can come under either blind or non blind watermarking. In blind watermarking the cover image is not used to decipher the watermark [3]. The cover image is used to decipher the watermark in the non-blind watermarking [4]. This paper proposes a non blind watermarking method based on the frequency domain. This method uses the transformations to improve the robustness of the system. The necessary features are perceptual quality and robustness to determine the quality of watermarking scheme.

In the frequency domain, DWT decomposes an image into multi-resolution components i.e. horizontal, vertical and diagonal [5] and DCT segregate each block into three frequency sub-bands: low, mid and high [3]. DFT requires an input image that is discrete. Such inputs

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are often created by sampling a continuous function. But, in this correspondence, DWT- DCT combination [7] and DCT coefficients of watermark are used in the proposed method. It improves all security factors in the data transfer. The strength of this paper is, analysis of the proposed scheme based on the standard metrics such as PSNR and NC and it sustains the general attacks like Gaussian noise, salt & pepper noise, Poisson noise, Gaussian blur, sharpening and common image processing operations like cropping, JPEG compression. The rest of the paper is organized in the following way. Section 2 consists of proposed method. Section 3 holds the description of performance metrics. Results and analysis are illustrated in Section 4.

### II. Proposed Method

The main aim of the proposed method is to improve the quality of watermarked image and robustness of the watermark. This approach consists of two major processes.

- Applying DWT and DCT on RGB cover image to get transformed image.
- Embed DCT coefficients of watermark image into transformed image.

### a) Watermark Embedding Algorithm

In watermark embedding process, till now researchers are using either DWT or DCT or both transformations. Here, we are using both DWT and DCT to transform the cover image as transformed image and a new and efficient embedding algorithm which embeds DCT coefficients of watermark image into transformed image to get the watermarked image as shown in Fig 1. It is elucidated in the following way.

### Algorithm

4.

Step 1: Consider any color image as cover image denote it by 'I'. Get R, G, B channels of cover image 'I'.

Step 2: Apply DWT to blue channel 'B' to get the multi-resolution sub-bands LL1, HL1, LH1, and HH1. When compared to red and green channels blue channel is more resistant to changes.

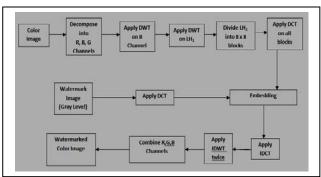
Step 3: Apply DWT again to HL1 sub-band of B channel and choose HL2 sub-band.

Step 4: Divide the HL2 sub-band into blocks of size  $8{\times}8.$ 

Step 5: Apply DCT to the blocks obtained in step

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Step 6: Consider any gray scale image as watermark, denote it by 'w' and then apply DCT to the watermark 'w'.



*Figure 1 :* Watermark embedding process

Step 7: To embed the watermark, select any four coefficients in the mid frequency band of each DCT block of HL2 sub-band.

Step 8: Four DCT coefficients of watermark are stored in each DCT block of HL2 sub-band.

Step 9: The DC component from the DCT coefficients of watermark is normalized before embedding.

Step 10: Apply IDCT to the blocks of HL2 sub-band.

Step 11: Apply IDWT for 2 levels.

Step 12: Combine R, G, B channels to get watermarked image 'WI'.

It helps to improve the copyright protection of the cover image and robustness of the watermark in the watermarked image.

### b) Watermark Extraction Algorithm

Extraction of watermark image from watermarked image is explained as follows and it is shown in Fig 2.

### Algorithm

4

Step 1: Get R, G, B channels of watermarked image 'WI'.

Step 2: Apply DWT to blue channel 'B' to get the multi-resolution sub-bands LL1, HL1, LH1, and HH1.

Step 3: Apply DWT again to HL1 sub-band of B channel and choose HL2 sub-band.

Step 4: Divide the HL2 sub-band into blocks of size  $8 \times 8$ .

Step 5: Apply DCT to the blocks obtained in step

Step 6: Obtain four DCT coefficients of watermark from the four selected coefficients of mid frequency band of each DCT of HL2 sub-band.

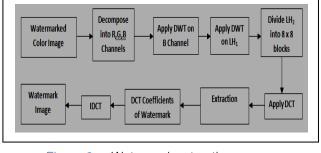
Step 7: To get the watermark apply IDCT on the set of DCT coefficients obtained from previous step.

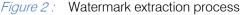
### III. Performance Evaluation Metrics

To measure the quality of the watermarked image peak signal to noise ratio (PSNR) is used. The quality of extracted watermark is measured using

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Normalized Correlation (NC) between the extracted and the original watermark.





### a) Peak Signal to Noise Ratio

It is a metric which is used regularly to find the quality of the watermarked image. It is calculated by considering the following formula

$$\mathsf{PSNR} = \log \frac{(2^n - 1)^2}{\mathsf{MSE}}$$

Where n denotes the number of bits used for color representation and MSE refers to the Mean Square Error between original and water marked image and is calculated with the formula.

$$MSE = \frac{\sum_{R,G,B} \sum_{i=1}^{M} \sum_{j=1}^{N} (I[i, j] - I'[i, j])^2}{3MN}$$

Here, M and N are the height and width of image respectively. I[i, j] denotes the (i, j)th pixel value of the original image and l'[i, j] denotes the (i, j)th pixel value of watermarked image.

### b) Normalized Correlation (NC)

It is one of the metrics used to find the quality of extracted watermark image with respect to the original watermark image. It is found by using the following formula.

$$NC = \frac{\sum_{i=1}^{M} \sum_{j=1}^{N} w(i, j) w'(i, j)}{\sqrt{\sum_{i=1}^{M} \sum_{j=1}^{N} w(i, j)^{2}} \sqrt{\sum_{i=1}^{M} \sum_{j=1}^{N} w'(i, j)^{2}}}$$

Here, w(i, j) is the original watermark, w'(i, j) is the extracted watermark.

### IV. Results and Analysis

As per this proposed method, a Lena color image of size  $512 \times 512$  is considered as cover image as shown in Fig. 3 and Fig. 4 is a gray scale image of skeleton with size  $32 \times 32$  is chosen as watermark.





Figure 3 : Original Image



*Figure 5*: Watermarked image

After applying the proposed method the resultant watermarked image is shown as Fig.5.

Table 1 summarizes PSNR value of watermarked image and NC value of extracted watermark image when the watermarked image is not attacked and undergoes any attack like Gaussian noise, salt & pepper noise, Poisson noise, Gaussian blur, sharpening and image manipulation operations like cropping, JPEG compression are applied.

| Attack / Operation  | PSNR  | NC    |
|---------------------|-------|-------|
| No attack           | 85.76 | 1.000 |
| Gaussian noise      | 65.46 | 0.996 |
| Salt & Pepper noise | 65.45 | 0.998 |
| Poisson noise       | 85.76 | 1.000 |
| Gaussian blur       | 79.18 | 0.998 |

68.73

60.28

63.45

0.995

0.920

0.910

Sharpening

JPEG compression

Cropping

*Table I* : PSNR and NC values

Table 2 shows comparison results of the proposed method with the existing transformation methods such as DWT [5], DWT-DCT [9] and Bior [10] based on the NC value between original and extracted watermark when the watermarked image undergoes any attacks.

| Attack / Operation  | DWT    | DWT   | Bior   | DCT   |
|---------------------|--------|-------|--------|-------|
|                     |        | -DCT  |        | Coeff |
| Gaussian noise      | 0.7415 | 0.993 | 0.8575 | 0.996 |
| Salt & Pepper noise |        | 0.997 | 0.8518 | 0.998 |
| Poisson noise       |        |       | 0.8567 | 1.000 |
| Gaussian blur       | 0.8968 | 0.997 |        | 0.998 |
| Sharpening          |        | 0.992 |        | 0.995 |
| Cropping            | 0.9031 | 0.861 | 0.8484 | 0.920 |
| JPEG compression    | 0.8775 | 0.756 | 0.7505 | 0.910 |

### Table II : Comparison Results

### V. Conclusion

This robust watermarking method is proposed for increasing the security of data hiding as well as quality compared with the existing algorithms. Authenticity is incorporated by embedding the DCT coefficients of watermark image into the transformed image. Experimental results based on attacks confirm that the proposed algorithm performs better than the DWT, DWT-DCT and Bior based scheme and robustness is achieved by hiding DCT coefficients of watermark image in transformed image by the frequency transformations as DWT and DCT.

### VI. Acknowledgments

We are very much thankful to the various reviewers who helped us to bring the paper effectively. The valuable suggestions of Dr. Ch.Rupa could not be underestimated in several situations of developing the paper. We sincerely thank our other research friends for giving their precious cautions. For the invaluable research facilities provided by the magnanimous management for achieving the work, we convey our deepest thanks.

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# An Impressive Method to Get Better Peak Signal Noise Ratio (PSNR), Mean Square Error (MSE) Values Using Stationary Wavelet Transform (SWT)

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*Abstract* - Impulse noise in images is present because of bit errors in transmission or introduced during the signal acquisition stage. There are two types of impulse noise, they are salt and pepper noise and random valued noise. In our proposed method, first we apply the Stationary wavelet transform for noise added image. It will separate into four bands like LL, LH, HL and HH. The proposed algorithm replaces the noisy pixel by trimmed median value when other pixel values, 0's and 255's are present in the selected window and when all the pixel values are 0's and 255's then the noise pixel is replaced by mean value of all the elements present in the selected window. This proposed algorithm shows better results than the Standard median filter (MF), decision based algorithm (DBA). The proposed method performs well in removing low to medium density impulse noise with detail preservation up to a noise density of 70% and it gives better Peak signal-to-noise ratio (PSNR) and mean square error (MSE) values.

Keywords : Peak signal-to-noise ratio (PSNR), mean square error (MSE) values.standard median filter (MF), decision based algorithm (DBA).

GJCST-F Classification: 1.4.5



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### An Impressive Method to Get Better Peak Signal Noise Ratio (PSNR), Mean Square Error (MSE) Values Using Stationary Wavelet Transform (SWT)

N. Naveen Kumar & Dr.S.Ramakrishna

Abstract - Impulse noise in images is present because of bit errors in transmission or introduced during the signal acquisition stage. There are two types of impulse noise, they are salt and pepper noise and random valued noise. In our proposed method, first we apply the Stationary wavelet transform for noise added image. It will separate into four bands like LL, LH, HL and HH. The proposed algorithm replaces the noisy pixel by trimmed median value when other pixel values, 0's and 255's are present in the selected window and when all the pixel values are 0's and 255's then the noise pixel is replaced by mean value of all the elements present in the selected window. This proposed algorithm shows better results than the Standard median filter (MF), decision based algorithm (DBA). The proposed method performs well in removing low to medium density impulse noise with detail preservation up to a noise density of 70% and it gives better Peak signal-to-noise ratio (PSNR) and mean square error (MSE) values.

Keywords : Peak signal-to-noise ratio (PSNR), mean square error (MSE) values.standard median filter (MF), decision based algorithm (DBA).

### I. INTRODUCTION

n the transmission of images over channels, images are corrupted by impulse noise, due to faulty communications. The objective of filtering is to remove the impulses so that the noise free image is fully recovered with minimum signal distortion. Noise removal can be achieved, by using a number of existing linear filtering techniques which are popular due to their mathematical simplicity and the existence of the unifying linear system theory. A new and efficient algorithm Javaraj etal[1] proposed high-density salt and pepper noise removal in images and videos. The non-linear filter like standard median filter (SMF), adaptive median filter (AMF), decision based algorithm (DBA) and robust estimation algorithm (REA) shows better results at low and medium noise densities. At high noise densities, their performance is poor.

Salt and pepper impulse noise is one commonly encountered noise type during image and video communication. So far the state of the art methods can reasonably restore images corrupted by salt and pepper noise whose level is up to 90%. A novel

quadratic type variation formulation of this noise removal problem is reported by wanetal [2]. This approach first uses a simple yet fast method to eliminate all salt-andpepper noise pixels as well as possibly some clean pixels, then the clean image is efficiently reconstructed from the remaining clean pixels by minimizing a carefully designed functional. Because the functional is quadratic type, fast unconditional convergence is guaranteed.

Recently a new adaptive weight algorithm Jianetal<sup>[3]</sup> for the removal of salt and pepper noise in. It consists of two major steps, first to detect noise pixels according to the correlations between image pixels, then use Different methods based on the various noise levels. For the low noise level, neighborhood signal pixels mean method is adopted to remove the noise, and for the high noise level, an adaptive weight algorithm is used. A switching bilateral filter (SBF) with a texture and noise detector for universal noise removal is proposed by Tsaietal [4]. Here the operation was performed in two stages: detection followed by filtering. For detection, the sorted guadrant median vector (SQMV) scheme, this includes important features such as edge or texture information. This information is utilized to allocate a reference median from SQMV, which is in turn compared with a current pixel to classify it as impulse noise, Gaussian noise, or noise-free. The SBF removes both Gaussian and impulse noise without adding another weighting function. The range filter inside the bilateral filter switches between the Gaussian and impulse modes depending upon the noise classification result.

To remove salt and pepper noise in video a new algorithm is proposed by Esakkirajan etal [5]. This proposed adaptive decision algorithm first checks whether the selected pixel in the video sequence is noisy or noise free. Initially the window size is selected as  $3 \times 3$ . If the selected pixel within the window is 0's or 255's, and some of other pixels within the window are noise free, then the selected pixel value is replaced by trimmed median value. If the selected pixel is 0 or 255 and other pixel values in a selected window ( $3 \times 3$ ) all are 0's and 255's, then change the selected window size

as 5  $\times$  5, then the selected pixel value is replaced by trimmed median value. In the selected new window (5  $\times$  5), all the elements are 0's or 255's then the processing pixel is replaced by the previous resultant pixel.

A modified simple edge preserved de-noising algorithm to remove salt and pepper noise in digital color images is discussed in [6]. This proposed algorithm has three steps: noisy pixel detection, replacement of noisy pixels, and confirmation by comparing with a threshold. In addition a median filtering is added to improve the quality of the image. It prevents the smoothing of edges in the noise removal process, by predicting the possible edges and taking the mean value from the predicted edge. Furthermore, Rahaman etal. [7] reported a new algorithm to remove Salt and Pepper noise from grayscale images. This is an enhanced adaptive median filtering algorithm which initially calculates median without considering noisy pixels in the processing window. If the noise-free median value is not available in the maximum processing window, the last processed pixel value is used as the replacement. However, in extreme situations such as noise corrupted pure black and white images, a threshold value is used to determine the pixel value.

A new wavelet-based de-noising method for medical infrared images was reported by Rabbani etal [8]. Since the dominant noise in infrared images is signal dependent utilize local models for statistical properties of (noise-free) signal and noise. In this base, the noise variance is locally modeled as a function of the image intensity using the parameters of the image acquisition protocol. In the next step, the variance of noise-free image is locally estimated and the local variances of noise-free image and noise are substituted in a wavelet-based maximum a posterior (MAP) estimator for noise removal.

A methodology based on median filters for the removal of salt and pepper noise by its detection followed by filtering in both binary and gray level images has been proposed by sarath etal [9]. Linear and nonlinear filters have been proposed earlier for the removal of impulse noise; however the removal of impulse noise often brings about blurring which results in edges being distorted and poor quality. Therefore, the necessity to preserve the edges and fine details during filtering is the challenge faced by researchers today. It consists of noise detection followed by the removal of detected noise by median filter using selective pixels that are not noise themselves. The noise detection is based on simple thresholding of pixels.

Impulse noise removal is one of the important image preprocessing techniques since the noise will lead the image processing procedures into an unexpected direction. The candidate-oriented strategy that detects the corrupted pixels (noise candidates) and then updates the intensity value of those pixels can achieve better performance than the brute-force strategy. Recently a novel region feature is presented in Ref.[10] to avoid the misclassification problem. In this method the noise pixels are treated as the small-sized regions, and labeled by the multi-scale connected component labeling algorithm. In this way, the region size can be considered as a clue during the noise detection procedure. This newly developed region feature can be easily utilized to the current noise removal algorithms.

### II. METHODOLOGY

The proposed image denoising corrupted by salt and pepper noise is built on Stationary wavelet transform. The following section described the Stationary Wavelet Transform (SWT) and Proposed Algorithms.

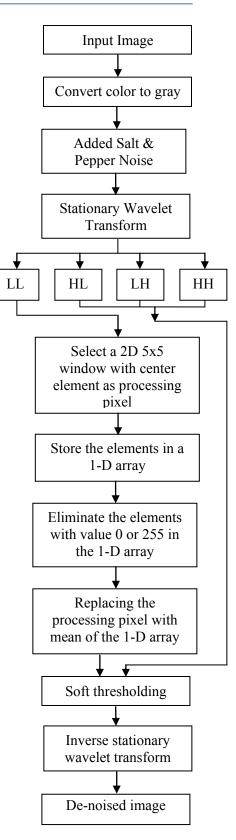
### a) Stationary Wavelet Transform

The SWT provides efficient numerical solutions the signal processing applications. It was in independently developed by several researchers and under different names, e.g. the un-decimated wavelet transform, the invariant wavelet transforms and the redundant wavelet transform. The key point is that it gives a better approximation than the discrete wavelet transform (DWT) since, it is redundant, linear and shift invariant. These properties provide the SWT to be realized using a recursive algorithm. Therefore, the SWT is very useful algorithm for analyzing a linear system. A brief description of the SWT is presented here. Figure shows the computation of the SWT of a signal x(k), whereW jk, and V jk are called the detail and the approximation coefficients of the SWT. The filters H j and G j are the standard low pass and high pass wavelet filters, respectively. In the first step, the filters H 1 and G 1 are obtained by up sampling the filters using the previous step (i.e. H j-1 andG j-1).

### III. Proposed Method

The Proposed method shown in Figure 1. In order to overcome the above mentioned difficulties, a new two-stage cascaded filter is proposed in this paper which removes the noise as high as possible, without blurring and retains the fine edge details. The proposed algorithm removes these drawbacks at high noise density. This algorithm processes the corrupted images by first detecting the impulse noise. The processing pixel is checked whether it is noisy or noisy free. That is, if the processing pixel lies between maximum and minimum gray level values then it is noise free pixel, it is left unchanged. If the processing pixel takes the maximum or minimum gray level then it is noisy pixel. It gives better Peak signal-to-noise ratio (PSNR) and image enhancement factor (IEF) values than the existing algorithm.

In adaptive modified filter the pixels are processed using 5 X 5 windows. During processing if a pixel is '0' or '255' then it is processed else it is left unchanged. In decision based algorithm (DBA) the corrupted pixel is replaced by the median of the window. At higher noise densities the median itself will be noisy, and the processing pixel will be replaced by the processed neighborhood pixel. This repeated replacement of neighborhood pixels produces streaking effect. To overcome the above drawbacks, the adaptive modified decision-based unsymmetric trimmed median filter is proposed. The corrupted processing pixel is replaced by a median or mean value of the pixels in the 5 X 5 window after trimming impulse values. The corrupted pixel is replaced by the median or mean of the resulting array. In this, the median value is replaced only when both the processing pixel and all the neighboring pixels are noisy pixels. In other case, if the processing pixel is noisy and all the neighboring pixels are not noisy pixel then the mean value is used for replacement. If the processing pixel itself is not a noisy pixel, then it does not require further processing. Thus the processing pixel value can be modified as median or mean value according to the cases and hence named as adaptive modified decision based unsymmetric trimmed median filter.



*Figure 1 :* Block diagram of proposed method

For all higher bands (LH, HL and HH) the denoising can be achieved by applying a thresholding operator to the wavelet coefficients in the transform domain followed by reconstruction of the signal to the  $T = MAD * \sqrt{2logn}$ 

original image in spatial domain. In the proposed method, soft shrinkage and median absolute difference

(MAD) are used. The scaled MAD noise estimator is calculated by (6).

(2)

by equation (3). Finally, the noise free image is obtained

$$MAD = \frac{median (|X|)}{0.6745} \tag{1}$$

Where X is the high frequency sub-bands coefficients. From the estimated noise, the non linear threshold T is calculated by equation (2)

Where N is the size of the high frequency subband array. Then the soft thresholding is applied to remove the noise and the soft shrinkage rule is defined

The noise free sub-bands are extracted using adaptive thresholding. Finally, the noise free image is obtained by taking the inverse SWT using the modified high frequencies sub-bands and the low frequency sub band of SWT.

### IV. Algorithm

Step 1: Select 2 -D window of size 5 X 5. Assume that the pixel being processed is  $P_{ij}$ .

Step 2: If  $0 < P_{ij} < 255$ then  $P_{ij}$  is an uncorrupted pixel and its value is left unchanged.

**Step 3:** If  $P_{ij}=0$  or  $P_{ij}=255$  then  $P_{ij}$  is a corrupted pixel then two cases are possible.

**Case i:** If the selected window contain all the elements as 0's and 255's. Then replace  $P_{ij}$  with the mean of the element of window.

**Case ii:** If the selected window contains not all elements as 0's and 255's. Then eliminate 255's and 0's and find the median value of the remaining elements. Replace  $P_{ii}$  with the median value.

Step 4: Repeat steps 1 to 3 until all the pixels in the entire image are processed.

Move the window by one step and repeat from step 1 to step 4. The above steps are repeated, until the processing is completed for the entire image.

#### a) Algorithm description

Each and every pixel of the image is checked for the presence of salt and pepper noise. Different cases are illustrated in this Section. If the processing pixel is noisy and all other pixel values are either 0's or 255's is illustrated in Case i. If the processing pixel is noisy pixel that is 0 or 255 is illustrated in Case ii. If the processing pixel is not noisy pixel and its value lies between 0 and 255 is illustrated in Case iii. If the processing pixel value is 0 or 255, then it is a corrupted pixel and it is processed by two cases:

**Case (i):** If the selected window contains salt or pepper noise as processing pixel (i.e., 255/0 pixel value)

 $\rho_T(x) = \begin{cases} x - T, & if \quad x \ge T \\ x + T, & if \quad x \le -T \\ 0, & if \quad |x| < T \end{cases}$ (3)

by taking the inverse SWT.

and neighboring pixel values contains all pixels that adds salt and pepper noise to the image. Consider the matrix [0 255 0 0 255 255 0 0 255 0 0 255 255 0 255 0 255 0 0 0 255 0 255 255 0 255]. Since all the elements surrounding Pij are 0's and 255's. If one takes the median value it will be either 0 or 255 which is again noisy. To solve this problem, the mean of the selected window is found and the processing pixel is replaced by the mean value.

**Case (ii):** If the selected window contains salt or pepper noise as processing pixel (i.e. 255/0 pixel value) and neighboring pixel values contains some pixels that adds salt (i.e. 255 pixel value) and pepper noise to the image. Now eliminate the salt and pepper noise from the selected window. That is, elimination of 0's and 255's. The 1-D array of the matrix is [78 90 108 0 120 0 255 97 255 73 112 95 0 83 75 56 180 0 78 0 55 43 255 93 87]. After elimination of 0's and 255's the pixel values in the selected window will be [78 90 108 120 97 73 112 95 83 75 56 180 78 55 43 93 87]. Here the processing pixel is replaced by median value of the remaining pixels.

If the processing pixel ( $P_{ij})$  value is not 0 or 255, then it is an uncorrupted pixel and it is processed by following case.

**Case (iii):** If the selected window contains a noise free pixel as a processing pixel, it does not require further processing. For example, 1-D array of the matrix is [43 67 70 55 75 108 112 143 164 85 97 45 80 95 100 45 87 43 190 87 90 79 85 81 66]. Here the processing pixel is 80 then it is noise free pixel. Since "80" is a noise free pixel it does not require further processing.

Repeat the steps until all the pixels in the entire image are processed.

### V. EXPERIMENTAL RESULTS

The proposed algorithm foe tested for 256x256 images. It is tested for various levels of noise values and also compared with Standard median filter (SMF).

Figure.2 shows the de-noising performance of the proposed algorithm. Table.1 show the PSNR values of

the proposed method and based soft shrinkage and SWT method with different noise variance.



Figure 2: (a) Input image, (b) Noise added Image, (c) Median filtered Image and (d) Our proposed method

| Noise density | PSNR in db    |       |       |                        |  |  |
|---------------|---------------|-------|-------|------------------------|--|--|
| in %          | Median filter | PSMF  | DBA   | Our Proposed<br>Method |  |  |
| 10            | 31.60         | 26.42 | 39.78 | 39.91                  |  |  |
| 20            | 31.38         | 26.42 | 37.29 | 37.31                  |  |  |
| 30            | 31.04         | 26.42 | 35.42 | 35.63                  |  |  |
| 40            | 30.64         | 26.42 | 34.09 | 34.42                  |  |  |
| 50            | 30.63         | 26.42 | 34.08 | 34.44                  |  |  |
| 60            | 29.59         | 26.42 | 32.06 | 32.54                  |  |  |
| 70            | 28.86         | 26.42 | 31.14 | 31.57                  |  |  |
| 80            | 28.21         | 26.42 | 30.14 | 30.54                  |  |  |
| 90            | 27.55         | 26.42 | 29.49 | 29.51                  |  |  |

*Table 1 :* PSNR value for the proposed method

### VI. CONCLUSION

In this work, image denoising based on Stationary wavelet transform (SWT) and soft threshold method are discussed. Experimental results show that the proposed method restore the original image much better than standard non linear median-based filters and some of the recently proposed algorithms. The proposed filter requires less computation time compared to other methods. The visual quality results clearly shows the proposed filter preserve fine details such as lines and corners satisfactorily. This filter can be further improved to apply for the images corrupted with high density impulse noise up to 90% and random valued impulse noise.

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### Performance Analysis of Modified Lifting Based DWT Architecture and FPGA Implementation for Speed and Power By C.Chandrasekhar & Dr.S.Narayana Reddy

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*Abstract* - Demand for high speed and low power architecture for DWT computation have led to design of novel algorithms and architecture. In this paper we design, model and implement a hardware efficient, high speed and power efficient DWT architecture based on modified lifting scheme algorithm. The design is interfaced with SIPO and PISO to reduce the number of I/O lines on the FPGA. The design is implemented on Spartan III device and is compared with lifting scheme logic. The proposed design operates at frequency of 280 MHz and consumes power less than 42 mW. The pre-synthesis and post-synthesis results are verified and suitable test vectors are used in verifying the functionality of the design. The design is suitable for real time data processing.

*Keywords : Lifting scheme, low power, high speed, FPGA implementation. GJCST-F Classification: 1.4.5* 

### PERFORMANCE ANALYSIS OF MODIFIED LIFTING BASED DWT ARCHITECTURE AND FPGA IMPLEMENTATION FOR SPEED AND POWER

Strictly as per the compliance and regulations of:



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### Performance Analysis of Modified Lifting Based DWT Architecture and FPGA Implementation for Speed and Power

C.Chandrasekhar<sup>a</sup> & Dr.S.Narayana Reddy<sup>o</sup>

*Abstract* - Demand for high speed and low power architecture for DWT computation have led to design of novel algorithms and architecture. In this paper we design, model and implement a hardware efficient, high speed and power efficient DWT architecture based on modified lifting scheme algorithm. The design is interfaced with SIPO and PISO to reduce the number of I/O lines on the FPGA. The design is implemented on Spartan III device and is compared with lifting scheme logic. The proposed design operates at frequency of 280 MHz and consumes power less than 42 mW. The presynthesis and post-synthesis results are verified and suitable test vectors are used in verifying the functionality of the design. The design is suitable for real time data processing.

Keywords : Lifting scheme, low power, high speed, FPGA implementation.

### I. INTRODUCTION

iscrete wavelet transforms (DWT) decomposes image into multiple subbands of low and high frequency components. Encoding of subband components leads to compression of image. DWT along with encoding technique represents image information with less number of bits achieving image compression. Image compression finds application in every discipline such as entertainment, medical, defense, commercial and industrial domains. The core of image compression unit is DWT. Other image processing techniques such as image enhancement, image restoration and image filtering also requires DWT and Inverse DWT for transformations. DWT-IDWT is one of the prominent transformation techniques that are widely used in signal processing and communication applications. DWT-IDWT computes or transforms signal into multiple resolution bands [1][2][3][4][5]. sub DWT is computationally very intensive and consumes power due to large number of mathematical operations. Latency and throughput are other major limitations of DWT as there are multiple levels of hierarchy [6][7][8]. DWT has traditionally been implemented by convolution. Digit serial or parallel representation of input data further decides the architecture complexity. Such an implementation demands a large number of computations and a large storage that are not desirable

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for either high-speed or low-power applications. Recently, a lifting-based scheme that often requires far fewer computations has been proposed for the DWT. The main feature of the lifting based DWT scheme is to break up the high pass and low pass filters into a sequence of upper and lower triangular matrices and convert the filter implementation into banded matrix multiplications. Since DWT requires intensive computations, several architectural solutions usina special purpose parallel processor have been proposed, in order to meet the real time requirement in many applications. The solutions include parallel filter architecture, SIMD linear array architecture, SIMD multigrid architecture, 2-D block based architecture, and the AWARE's wavelet transform processor (WTP) [9][10][11]. Several versions of lifting scheme architecture have been compared and reported in literature. In terms of hardware complexity, the folded architecture in [12] is the simplest and the DSP-based architecture in [13] is the most complex. All other architectures have comparable hardware complexity and primarily differ in the number of registers and multiplexor circuitry. The control complexity of the architecture in [14] is very simple. In contrast, the number of switches, multiplexors and control signals used in the architectures of [15] is quite large. The control complexity of the remaining architectures is moderate. In terms of timing performance, the architectures in [14, 12, 16–18] are all pipelined, with the architectures in [17] having the highest throughput (1/Tm). The architecture in [19] has fewer cycles since it is RPA based, but its clock period is higher. The architecture in [17] has the lowest computation delay.

In this paper, we propose, design, model, implement and compare the performances of three different DWT architectures. Section II briefly discusses the Lifting Scheme DWT algorithm for image processing, Section III discusses modified lifting base DWT and Section IV presents the FPGA implementation and compares the results of modified lifting algorithm. Conclusion is presented in Section VI.

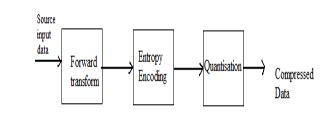
### II. Dwt

The influx of sophisticated technologies in the field of image processing is affiliated with that of

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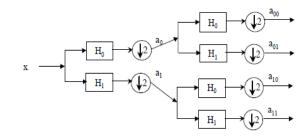
*Author σ* : Prof & Head in Dept.of ECE, S.V.University, Tirupathi. E-mail : snreddysvu@yahoo.com

digitization in the computers arena. Image Compression plays an important role of all the Image Processing techniques. The compression techniques are of two types: Lossless and Lossy. The most common image format that uses a lossy compression scheme is JPEG (Joint Photographic Experts Group) format. JPEG 2000 structure is wavelet based compression methodology that provides a number of benefits over the Discrete Cosine Transformation (DCT) compression method, which was used in JPEG format. Wavelet compression converts the image into a series of wavelets that can be stored more efficiently than pixel blocks. The Wavelet compression is accomplished through the use of JPEG 2000 encoder as shown in the figure 1.



### Figure 1 : JPEG 2000 Block Diagram

The problem statement in the present section deals

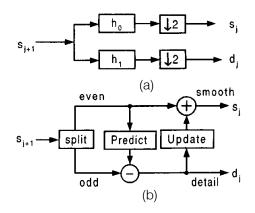


### Figure 2 : Two-level DWT decomposition [6]

With the design of the modified two-level DWT architecture for decomposition. The Discrete Wavelet Transform (DWT), which is based on sub-band coding is found top yield a fast computation of Wavelet Transform. It is easy to implement and reduces the computation time and resources required. In DWT, a time-scale representation of the digital signal is obtained using digital filtering techniques [6]. The signal to be analyzed is passed through filters with different cut-off frequencies at different scales as shown in figure 2.

### Lifting Scheme:

The Lifting Scheme is a well known method for constructing bi-orthogonal wavelets. The main difference with the classical construction is that it does not rely on the Fourier transform. The lifting scheme is an efficient implementation of a wavelet transform algorithm. It was primarily developed as a method to improve wavelet transform, and then it was extended to a generic method to create so-called second-generation wavelets. Second-generation wavelets are much more flexible and powerful than the first generation wavelets. The lifting scheme is an implementation of the filtering operations at each level [6]. The figure 3 represents the classical and lifting based implementations of DWT.



*Figure 3 :* a) Classical Implementation, b) Lifting scheme based DWT [6]

Lifting Scheme consists of three steps: SPLIT, PREDICT and UPDATE, as shown in the figure 3 (b).

- **SPLIT:** In this step, the data is divided into ODD and EVEN elements.
- **PREDICT:** The PREDICT step uses a function that approximates the data set. The differences between the approximation and the actual data, replace the odd elements of the data set. The even elements are left unchanged and become the input for the next step in the transform. The PREDICT step, where the odd value is "predicted" from the even value is described by the equation [6].

$$Odd_{j+1, i} = odd_{j, i} - P(even_{j, i})$$

• UPDATE: The UPDATE step replaces the even elements with an average. These results in a smoother input for the next step of the wavelet transform. The odd elements also represent an approximation of the original data set, which allows filters to be constructed. The UPDATE phase follows the PREDICT phase. The original values of the odd elements have been overwritten by the difference between the odd element and its even "predictor". So in calculating an average the UPDATE phase must operate on the differences that are stored in the odd elements [6]:

$$Even_{j+1, i} = even_{j, i} + \boldsymbol{U} (odd_{j+1, i})$$

The equations for the lifting based implementation of the bi-orthogonal wavelet are:

Predict P1:  $d_i^{1} = \alpha (x_{2i} + x_{2i+2}) + x_{2i+1}$ Update U1:  $a_i^{1} = \beta (d_i^{1} + d_{i-1}^{-1}) + x_{2i}$ Predict P2:  $d_i^{2} = \gamma (a_{i1} + a_{i+1}) + x_{2i+1}$ Update U2:  $a_i^{2} = \delta (d_i^{2} + d_{i-1}^{-2}) + a_i^{1}$  Scale G1:  $a_i = \zeta a_i^2$ Scale G1:  $d_i = d_i^2 / \zeta$ 

The figure 4 shows the lifting scheme architecture to realise the equations shown above.

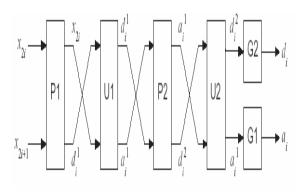


Figure 4 : Lifting Scheme Architecture

The input data x is first split into even and odd samples and each of the samples are taken through predict and update stages as per the architecture shown above. As the data moves from first stage to the last stage, data switching occurs at the input and output of every stage. Every stage consists of multipliers and adders. For the given set of Predict and Update stages, assuming the value of i = 0, the equation can be finalized.

### III. MODIFIED LIFTING SCHEME

By re-arranging all the values and the constant co-efficient, the final equation can be derived.

$$\begin{split} ai &= (\underline{3 * \gamma . \beta . \delta . \zeta + \delta . \zeta + \beta . \zeta}) \left[ \alpha \left( x_0 + x_2 \right) + x_1 + \alpha \left( x_0 + x_{-2} \right) + x_{-1} \right] + \underline{\zeta . \delta . \beta . \gamma} \left[ \alpha (x_2 + x_4) + x_3 + \alpha (x_{-2} + x_{-4}) + x_{-3} \right] + \underline{\zeta . \delta . \gamma} \left( x_0 + x_2 + x_0 + x_{-2} \right) + \underline{\zeta} * x_0 \end{split}$$

 $\begin{array}{l} di = 1/\zeta \left[ (\underline{2 \ast \gamma . \beta + 1}) \left\{ \begin{array}{l} \alpha \left( \begin{array}{c} x_0 + x_2 \right) + x_1 \right\} + \underline{\gamma . \beta} \\ x_{.2} \right) + x_{.1} + \alpha (x_2 + x_4) + x_3 \right\} + \underline{\gamma} (x_0 + x_2) \right]. \end{array}$ 

Being a dedicated DWT core for JPEG 2000, the filter coefficients are fixed. The filter coefficients are:  $\alpha = 1.586134342$ ,  $\beta = 0.05298011854$ ,  $\gamma = 0.8829110762$ ,  $\delta = 0.4435068522$ ,  $\zeta = 1.149604398$ . By substituting the above values in the modified equation, the coefficient values obtained then are also decimals, by multiplying them with constants they form integers as: 1 \* 32 = 57, 2 \* 256 = 6, 3 \* 64 = 30, 4 \* 32 = 35, 5 \* 256 = 12, 6 \* 32 = 26, 7 \* 32 = 50.

Thus the above integers are the values of the underlined coefficients in above equations. From the equations it is observed that there are common lifting coefficients to compute ai and di coefficients and there are input terms. The architecture realised by the above equations considering the constant coefficients is shown in the figure 5.

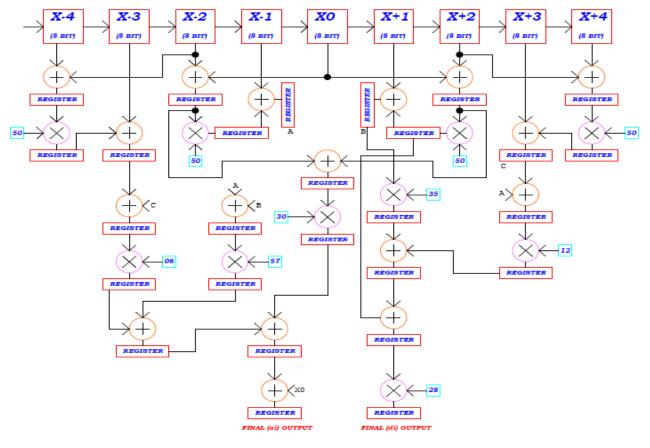


Figure 5 : Modified Lifting Scheme Architecture for DWT

The FPGA implementation of the modified lifting based DWT is designed based on the following:

- The input data X should be of 8 bit signed data.
- Output should be of 16 bit signed representation (including a<sub>i</sub> and d<sub>i</sub>).
- The lifting coefficients should be of 8 bit signed representation.
- The architecture should work on the streaming input sequence (Serial input).
- The intermediate outputs should be stored in a memory.
- The input data flows into the architecture through one input line and the output should be read out through one output line.

Thus by considering the above design specifications the architecture shown in the figure 5 is designed as per the requirements.

- The blocks from X -4 to X +4 resemble the input 9 samples designed in form of SIPO, each of 8-bit signed representation (serial in parallel out).
- Here the input stream is given through the single input line to 9 SIPOs. The outputs of those are taken in parallel to perform the addition and multiplication operations.
- The addition and multiplication operations are of 8 bit signed operators.
- The intermediate results of these addition and multiplication operations are stored in registers than preferring memory, as the data can be stored in registers with ease and in random, but in memory the storage (write operation in particular) should be done in orderly fashion.
- These intermediate registers are of PIPO structure and of 8 bit signed representation.
- Though final outputs ai and di are single bit, those are stored in the registers of PISO structure as the output should be taken for 8 bits.

The inputs can be 8-bit signed at any point of time. But the outputs should not be a signed number and can be more than 8-bits as every time the adders create an extra bit and the multipliers create more than one bit of data. That might be the major cause for the failure of the hardware; the architecture might not work properly. In order to minimize the error, suitable modifications are carried out.

### Modifications to minimize the Errors:

The few possible modifications that can be done for the calculations which can minimize the errors are:

1. An adder performs the addition of two 8-bit numbers and gives the result as a 9-bit number. Instead of a 9-bit number the LSB is discarded. As the Least Significant Bit is discarded the value of the number might not change drastically and the output still is an 8-bit data which is used for further operations.

- 2. The multiplier performs multiplication of one 8-bit number and the other is coefficient numbers. For each multiplication the hardware will be different so the final architecture requires a lot of multipliers which are of different width and again gives different output values.
- 3. The lifting co-efficient which should be of 8-bit signed number goes in decimal numbers like **0.458** so that the computation will become very difficult. For multiplying this number multiplier takes more time to compute and the final output would be a decimal as **57.35**.

The lifting coefficient is multiplied with an integer as **57** so that representing it might be much easier. These coefficients should be multiplied such that the final values should be obtained as 8-bit signed number and it should not have any decimal value as **57.000**. This can be achieved by taking only the positive values and discarding all other decimal point values (e.g.XY.xy). Thus the values of all the lifting coefficients have integer without any decimal values so that the calculations can be much easier. From the architectural calculations, the values of ai and di are **65** and **39** respectively, and match with theoretical calculations.

By comparing those values we can come to know that:

- 1. The architecturally calculated values are of 8-bit singed representation while theoretically calculated are unsigned.
- 2. The architectural values do not have any decimal values.
- 3. The architectural values do not exceed more than 8bit.
- 4. The intermediate calculations will be always 8-bit and signed instead of 9 or more bits.
- 5. The outputs of the adder in architectural calculations are 8-bit by discarding the LSB than having 9-bits which will be continued to increase for next level of addition.

Estimation of Power, Area and Delay of Sub-Blocks of Architecture:

The main sub-blocks of the modified lifting scheme architecture are:

- ✓ Adders
- ✓ Multipliers (Constant Coefficient-IP Cores)
- ✓ Registers

The table 1 represent the estimation of Power, Delay and Area of these sub blocks.

| Sl  | Sub Block      | Power | Area(no of    | Delay(n |
|-----|----------------|-------|---------------|---------|
| No  |                | (W)   | slices,utiliz | S)      |
|     |                |       | ation of      |         |
|     |                |       | slice%)       |         |
| 1)  | Adder          | 0.142 | 5(1%)         | 3.369   |
| 2)  | Multiplier(50) | 0.007 | 9 (1%)        | 1.999   |
| 3)  | Multiplier(30) | 0.037 | 7(1%)         | 0.881   |
| 4)  | Multiplier(12) | 0.037 | 6(1%)         | 0.881   |
| 5)  | Multiplier(35) | 0.037 | 13(1%)        | 0.881   |
| 6)  | Multiplier(57) | 0.037 | 11(1%)        | 0.881   |
| 7)  | Multiplier(6)  | 0.037 | 6(0%)         | 0.881   |
| 8)  | Multiplier(26) | 0.037 | 7(1%)         | 0.881   |
| 9)  | PISO           | 0.011 | 5(1%)         | 1.216   |
| 10) | SIPO           | 0.037 | 36(1%)        | 0.915   |
| 11) | PIPO           | 0.037 | 1(0%)         | 0.916   |

*Table 1 :* Estimation of Power, Delay and Area of sub blocks

The table 1 represents the estimation of Power, Delay and Area of the Sub-blocks. Here every block is of 8 bit signed representation from +127 to -128. As observed from the table there are one type of adder, multipliers of 7 types of constant coefficient type to reduce the complexity of multiplication the IP cores in XILINX are used. Of all the sub-blocks the Adder has the highest delay and the highest utilisation of the resources. Thus by instantiating these sub-blocks the area utilised by the DWT architecture is 12% and the delay is 3.313ns. From the table 1 the delays of individual blocks are known. Almost all work at different clock frequencies, as the delay mentioned in the table is the minimum period of the design clock. But the whole or the top level design should work at one clock frequency, thus the concept of synchronising the clocks arise. The clock frequency of top level architecture should synchronise with the sub modules, in general the problem of Synchronisation is addressed by any of these below:

- Increase system clock period (usually not feasible).
- Decrease tcomb (use no combinational logic).
- Decrease tsu (use fast flip-flops)
- Increase synchroniser clock period.

The figure 6 represents the clock synchroniser.

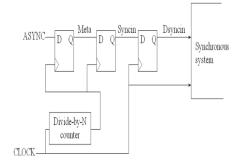
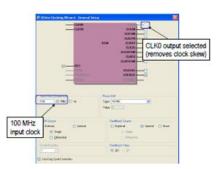


Figure 6 : Clock Synchroniser

In applying the same for the implementation of the present top-level design architecture, the DCM is used. The Divide-by-N counter block in the figure 7 can be replicated by using DCM. The figure 2.8 represents how to configure the DCM.



*Figure 7 :* Configuring DCM [2]

Thus by configuring the DCM in the frequency mode the tool generates the instantiation template and thus that instantiation template is used in the design to make the design run on same clock. The operating frequency of the present design runs at 280MHz.

### Observations:

- The equation of the lifting scheme for two-level DWT is simplified based on the basic equations mentioned.
- The simplified equation is made into an architecture such that both the ai and di is implemented using the same architecture.
- The mathematical and the architectural computation of the equation are computed and compared, and observed that the architectural computations are the modified version of the mathematical, where the discarding of the LSBs result to scaling down of the original values.
- The power, area and the delay of the sub-blocks are observed and noticed that the Adder takes the maximum delay i.e. 3.39ns and maximum utilisation of resources i.e. 13%, and registers SIPO and PIPO takes the least delay and least utilisation of resources i.e. 1%

### IV. HDL MODELING AND FPGA Implementation

The top level module or block of the DWT architecture is shown in the figure 8. The figure explains the input and output ports. The input ports are **clk**, **en**, **piso\_load**, **rst and ser\_in** and the output ports are **ai** and **di**. The input 9 samples each of 8 bit signed data is entered into the design through the **ser\_in** input. The **rst** signal is used to reset the design when the signal is high. When the **en** signal is high, loading of the input data in all the 9-8 bit registers for 280 clock cycles is done. The **piso\_load** signal is used to take the output at **ai** and **di**, and this signal is kept high for 8 clock cycles

as the 8 bit is to be taken out through the single line. The HDL models of the sub-block can be understood from the internal hardware of the RTL schematic shown in the figure 9. The figure 9 represents the schematic of the DWT architecture where all the sub blocks can be viewed. Thus the sub blocks are modelled in such a way that the multipliers used are the IP cores from the XILINX library, and the adder that is designed for 8 bit signed addition is instantiated wherever necessary. The simulation of the top level module is shown in the figure 10 where the intermediate signals gives the performance of the sub blocks in the total simulation.

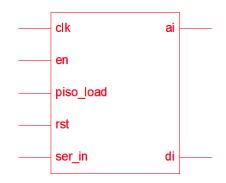


Figure 8 : Top-level DWT architecture

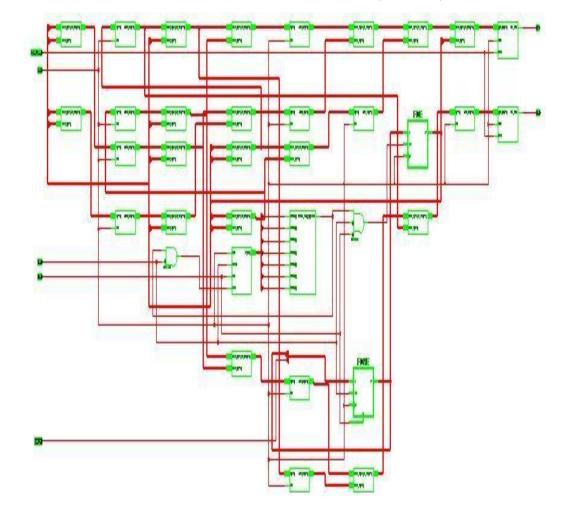
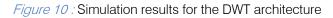


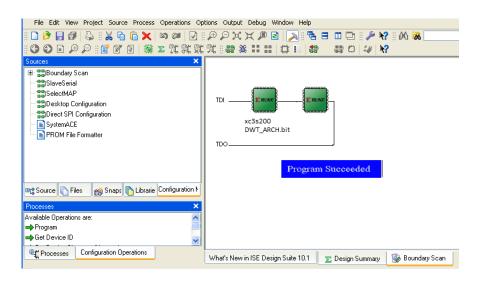
Figure 9 : RTL schematic of DWT showing sub-block

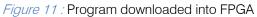
The figure **explains the integration of the sub blocks** in the main top level architecture. Initially the sub blocks are designed by considering the DWT equation, the multiplier used in the design is a **constant coefficient multiplier** as it is faster than any other for the application required. For the present design, the constant coefficient multipliers are taken as a **IP core** from the XILINX library for different coefficients. The **adder** is 8-bit signed operator designed or modelled in the HDL and instantiated where it is necessary. The registers that are used in the design covers all the types SIPO, PIPO, PISO. SIPO at the initial stage while giving the inputs, PIPO while performing the operations intermediately, and the PISO at the output stage to take the outputs serially i.e. one bit for 8 clocks, as the required is two outputs of 8 bits taking serially. From the figure 9, the top level ports are shown; the serial input data is given in a random way. This is loaded inthe registers (SIPO), when enable signal is high, after 72 clock cycles the enable is made low, and for four clock cycles the operation is performed and the output is taken when the piso load signal is high for 8 clock cycles as the output is taken for 8 bit. Thus the same procedure follows for 8 (load) + 4 (operation) + 8 (output) = 22 clocks. To program a single device using iMPACT, all needed is a bitstream file. To program several devices in a daisy chain configuration, or to program devices using a

PROM, iMPACT is used to create a PROM file. iMPACT accepts any number of bitstream and creates one or more PROM files containing one or more daisy chain configurations.

|  |                  | signal or<br>to load                    | Serial input data |                          | a, for 8 clock<br>cycles           | d; for 8 clock<br>cycles         |
|--|------------------|---|-------------------|--------------------------|------------------------------------|----------------------------------|
| kihora_twoj 🎸<br>bihora_twoj 🎸<br>kihora_twoj 🎸<br>bihora_twoj 🎸 | Pisc             | _load signal<br>output 8 clocks         | Sitte             |                          |                                    |                                  |
| dwt_archin   | n -No Data-      |   |                   |                          |                                    |                                  |
| <pre>dowt_Arch/p</pre>   |                  |   |                   | -01111010                | ¥01000001                          | ¥10000000                        |
| B- IDWT_ARCHIX   |                  |   |                   | 0110110                  | 101111010                          | 10000001                         |
|  |                  |   |                   | 01100110                 | 201101110                          | 201111010                        |
| D  |                  |   |                   | 01011011                 | 201100110                          | 201101110                        |
|  | z -No Data-      |   |                   | 01001000                 | 201011011                          | D1100110                         |
|  | po -No Data-     |   |                   | 00110110                 | 101001000                          | 201011011                        |
| . JDWT_ARCH/x  | ptw -No Date-    |   |                   | 00110000                 | 200110010                          | 101001000                        |
| DWT_ARCH/x   | _pth _No Data-   |   |                   | 00011011                 | 00110000                           | 200110110                        |
| DWT_ARCH/x   | _trf =No Data⊷   |   |                   | 00000110                 | 200011011                          | 200110000                        |
| JDWT_ARCH/s  | po_out -No Data- | 000000000000[]]]]]]]]]]]]]]]]]]]]]]     |                   | 111111111110000011000011 | 211021 []]]]]][2000110]10011000000 | 1101 []]]][]][001100000011011001 |
| DWT_ARCHIO   | ount -No Data-   | 0000000 1111111111111111111111111111111 |                   | 1111111111111111000000   | 111111111000000                    | [111111111000000                 |
| 💀 🤣 jDWT_ARCHIa  |                  |   |                   | 00111100                 | 101001000                          | 201010111                        |
| 🖬 🎝 /DWT_ARCHI:  |                  | 00000000                                |                   | 00101110                 | 200111000                          | 101000011                        |
| JDWT_ARCH/a  |                  |   |                   | 00110010                 | 1201000000                         | 1201001111                       |
| JDWT_ARCH/a  |                  |   |                   | 01010111                 | \$01100100                         | D1110000                         |
| 🖬 🍫 /DWIT_ARCH/in  |                  | 00000000                                |                   | 01000011                 | 201001110                          | 101010111                        |
| 🖬 🌗 JDWIT_ARCH/a   |                  |   |                   | 01001111                 | 101011010                          | 101100010                        |
| 🖪 🍫 JOWIT_ARCH/a   | dd_4 -No Data-   |   |                   | 01001001                 | 101010110                          | 201100011                        |
| 🖬 🎝 JOWT_ARCHI   | 30m1 -No Data-   | 000000003                               |                   | 01000100                 | 201010000                          | 201011100                        |
| 🖬 🥠 /DWT_ARCH/a  |                  |   |                   | 00011011                 | 100101000                          | 200111100                        |
| 🖬 🍫 /DWT_ARCH/s  |                  | 0000000                                 |                   | 00010101                 | 100011111                          | 100101110                        |
| 🖬 📣 /DWT_ARCH(6  |                  |   |                   | 00011000                 | ()00100111                         | 100110010                        |
| 🖬 🔩 įdwit_archia   |                  |   |                   | 01110000                 | 201010111                          | D1111101                         |
| B- /DWT_ARCHIN   |                  | 00000000                                |                   | 01010111                 | 201000011                          | 11111101                         |
| DWT_ARCH/a   |                  |   |                   | 01100010                 | 101011110                          | 1000011111                       |
| E- /DWIT_ARCH/a  | dd_9 -No Data-   |   |                   | 00110011                 | 101000000                          | 1201001010                       |
| DWT_ARCHI  | 12m1 -No Data-   | 00000000                                |                   | 01001100                 | D01100000                          | 101101111                        |





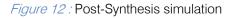


### Comparison of Post and Pre- Synthesis results:

The Post-synthesis result is that obtained after the place and route process has done. Here the delays of the LUTs and also the interconnection delay are added. But the Pre-Synthesis is nothing but the behavioral simulation. The comparison between the Post and Pre Synthesis results is shown in the figure 12 and 13.

### Performance Analysis of Modified Lifting Based DWT Architecture and FPGA Implementation for Speed and Power

| Messages                               | 2 h                    |  |
|--|------------------------|--|
| /DWT_ARCH/dk                           | -No Data-              | עריי היי היי היי היי היי היי היי היי היי   |
| 👉 /DWT_ARCH/ai                         | -No Data-              |  |
| 🔷 /DWT_ARCH/ser_in                     | -No Data-              |  |
| 🔷 /DWT_ARCH/di                         | -No Data-              |  |
| 👉 /DWT_ARCH/en                         | -No Data-              |  |
| 🔷 /DWT_ARCH/rst                        | -No Data-              |  |
| 🐓 /DWT_ARCH/piso_load                  | -No Data-              |  |
| /DWT_ARCH/en_IBUF_0                    | -No Data-              |  |
| /DWT_ARCH/piso_load_IBUF_0             | -No Data-              |  |
| /DWT_ARCH/rst_IBUF_0                   | -No Data-              |  |
| 🔷 /DWT_ARCH/ser_in_IBUF_0              | -No Data-              |  |
| /DWT_ARCH/dk_BUFGP                     | -No Data-              | ու առավորությունը հարարարան հարարան հա |
| /DWT_ARCH/\add_12<4>_0                 | -No Data-              |  |
| /DWT_ARCH/\add_12<1>_0                 | -No Data-              |  |
| /DWT_ARCH/\add_12<3>_0                 | -No Data-              |  |
| /DWT_ARCH/\add_12<0>_0                 | -No Data-              |  |
| /DWT_ARCH/\add_12<5>_0                 | -No Data-              |  |
| 🔶 /DWT_ARCH/\m57/m57/BU2/U             | -No Data-              |  |
| /DWT_ARCH/\add_12<6>_0                 | -No Data-              |  |
| /DWT_ARCH/\m57/m57/BU2/U               | -No Data-              |  |
| /DWT_ARCH/\m57/m57/BU2/U               | -No Data-              |  |
| /DWT_ARCH/\add_12<7>_0                 | -No Data-              |  |
| /DWT_ARCH/\m57/m57/BU2/U               | -No Data-              |  |
| 2 /DWT_ARCH/\m57/m57/BU2/U             | -No Data-              |  |
| /DWT_ARCH/\m57/m57/BU2/U               | -No Data-              |  |
| <pre>/DWT_ARCH/\add_1&lt;4&gt;_0</pre> | -No Data-              |  |
| /DWT_ARCH/\m35/m_35/BU2/               | -No Data-              |  |
| /DWT_ARCH/\m35/m_35/BU2/               | -No Data-              |  |
| /DWT_ARCH/\add_1<5>_0                  | -No Data-              |  |
| /DWT_ARCH/\m35/m_35/BU2/               | -No Data-              |  |
| /DWT_ARCH/\add_1<6>_0                  | -No Data-              |  |
| /DWT_ARCH/\m35/m_35/BU2/               | -No Data-<br>-No Data- |  |
| /DWT_ARCH/\m35/m_35/BU2/               | -No Data-              |  |
| /DWT_ARCH/\m35/m_35/BU2/               | -No Data-              |  |



|  | Enable signa                    |               | Serialinput | data         | a for 8 clock<br>cycles | di for 8 clock<br>cycles |
|--|---------------------------------|---------------|-------------|--------------|-------------------------|--------------------------|
| 🖕 (DWT_ARCH)di 🚽   | ko<br>No Deta-                  |               |             |              |                         |                          |
|  | lo Data-<br>lo Data-<br>Piso_lo | ad signal for |             | 17           |                         |                          |
| Contraction of the Contract of | io Data- output                 | 8 clocks      |             |              |                         |                          |
| 🗳 (DWT_ARGH)en 🔄   | lo Data-                        |               |             |              |                         |                          |
| 🚽 (DWT_ARCH(piso_load -1   | io Data-                        | S             |             |              |                         |                          |
| D / DWIT_ARCH/x_mf 4   | io Data-                        |               |             | 01111010     | 101000001               | 10000000                 |
| DWIT_ARCH/x_mth +  | io Data-                        |               |             | 01101110     | 101111010               | 101000001                |
| DWT_ARCH/x_adm -1  | io Data-                        |               |             | 01100110     | 101101110               | 101111010                |
| DWT_ARCH(x_mo -1   | io Data-                        |               |             | 01011011     | 101100810               | 101101110                |
| termine (AARDAR CARDING AND A CARDINE IN CONTRACT OF A CARDINAL OF A   | io Data-                        |               |             | 01001000     | 201011011               | 201100110                |
| DWT_ARCH/x_po 4  | No Evata-                       |               |             | 00110110     | 201001000               | 101011011                |
| Contraction of the Contraction o | io Data-                        |               |             | 00110000     | 200110110               | \$21001000               |
| Contraction of the second s  | io Data-                        |               |             | 00011011     | 200110000               | 200110110                |
|  | io Data-                        |               |             | 00000110     | 100011011               | 20110000                 |
|  |                                 |               |             |              |                         |                          |
| Contraction of the second seco | lo Data-                        |               |             | 111111000000 | 1111111111111000000     | 111111111000000          |
| - Contraction - States - States - States   | io Data-                        |               |             | 00111100     | 01001000                | B1010111                 |
| and the state of t | io Data- 0000000                |               |             | 00101110     | 100111000               | 101000011                |
| Land Control Control Control Control International Control Con | io Data-                        |               |             | 00110010     | 101000000               | 1201001111               |
| In the state of th | ko Data-                        |               |             | 01010111     | 01100100                | X01110000                |
| Contraction of the Contraction o | io Data- 10000000               |               |             | 01000011     | 01001110                | 101010111                |
| A REAL PROPERTY IN COMPANY OF THE  | io Data-                        |               |             | 01001111     | 101011010               | 101100010                |
| Land All Children Anto Children  | lo Data-                        |               |             | 01001001     | 101010110               | 201100011                |
| CARLANCAMED, MADE N  | ko Data-                        |               |             | 01000100     | 201010000               | 201011100                |
| and the second se  | io Data-                        |               |             | 00011011     | 200101000               | 20111100                 |
| Contraction of the second seco | lo Data-                        |               |             | 00010101     | 200011111               | 200101110                |
| Contract of the second s  | ko Data-                        |               |             | 00011000     | 100100111               | 100110010                |
|  | io Data-                        |               |             | 01110000     | 201010111               | Li1111101                |
|  | ko Data- 10000000               |               |             | D1010111     | 01000011                | 11111101                 |
| and the second se  | io Data-                        |               |             | 01100010     | 101011110               | (000011111               |
| Carlos and a second sec | io Data-                        |               |             | 00110011     | 101000000               | 101001010                |
| 🖬 🚮 jDWIT_ARCH/m12m1 🚽   | ki Data- 10000000               |               |             | 01001100     | 101100000               | 101101111                |

Figure 13 : Pre-Synthesis Simulation

### Logic Utilization

Number of Slice Flip Flops: 247 out of 1,536 Number of 4 input LUTs: 295 out of 1,536

### Logic Distribution

Number of occupied Slices: 245 out of 768

Number of Slices containing only related logic: 245 out of 245

Number of Slices containing unrelated logic: 0 out of 245

Total Number of 4 input LUTs: 314 out of 1,536 Number used as logic: 295 Number used as a route-thru: 19 Number of bonded IOBs: 6 out of 140 Number of GCLKs: 1 out of 4 Number of GCLKIOBs: 1 out of 4

### Place and Route

Placement and routing is performed by the PAR program. *Place and route* is the most important and time consuming step of the implementation. It defines how device resources are located and interconnected inside an FPGA. Placement is even more important than routing, because bad placement would make good routing impossible. In order to provide possibility for FPGA designers to tweak placement, PAR has a "starting cost table" option. PAR accounts for timing constraints set up by the FPGA designer. If at least one constraint can't be met, PAR returns an error. The output of the PAR program is also stored in the NCD format. The device utilization summary of the architecture is given below.

### Device Utilization Summary

- Number of GCLKs: 1 out of 4 25%
- Number of External GCLKIOBs: 1 out of 4 25%
- Number of LOCed GCLKIOBs : 0 out of 1 0%
- Number of External IOBs: 6 out of 140 4%
- Number of LOCed IOBs: 0 out of 6 0%
- Number of SLICEs: 245 out of 768 31%

| Parameters           | DWT<br>[9]         | Lifting<br>DWT<br>[18][19] | Modified Lifting<br>DWT |
|----------------------|--------------------|----------------------------|-------------------------|
| No of Slices         | 432 out<br>of 1536 | 358 out<br>of 1536         | 247 out of 1536         |
| No of gates          | 37K                | 27K                        | 15K                     |
| Clock Speed          | 72<br>MHZ          | 156 MHz                    | 268 MHz                 |
| Power<br>dissipation | 81 mW              | 67 mW                      | 42 mW                   |

Table 2 : Performance comparison of DWT architecture

### V. Conclusion

In this work a modified lifting based DWT architecture is proposed, designed, modeled and verified. The design is modeled using HDL and is implemented on FPGA. The interfaces requried for data processing are also designed and is used to synchronize the data transfer operation. The HDL models and simulation of the sub blocks have been done to model the top-level design architecture. The test-bench to verify the functionality and performance of the sub modules and the top level architecture have been done. Implemented the design on FPGA and verified and debugged through the Chip-Scope. The Pre

and Post Synthesis have been done and compared. The design can be further optimized for video signal processing.

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#### You can use your own standard format also. Author Guidelines:

1. General,

- 2. Ethical Guidelines,
- 3. Submission of Manuscripts,
- 4. Manuscript's Category,
- 5. Structure and Format of Manuscript,
- 6. After Acceptance.

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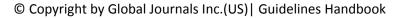
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- Shield the model why did you employ this particular system or method? What is its compensation? You strength remark on its appropriateness from a abstract point of vision as well as point out sensible reasons for using it.
- Present a justification. Status your particular theory (es) or aim(s), and describe the logic that led you to choose them.
- Very for a short time explain the tentative propose and how it skilled the declared objectives.

## Approach:

- Use past tense except for when referring to recognized facts. After all, the manuscript will be submitted after the entire job is done.
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- Do not take in frequently found.
- If use of a definite type of tools.
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## Approach:

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## What to keep away from

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The page length of this segment is set by the sum and types of data to be reported. Carry on to be to the point, by means of statistics and tables, if suitable, to present consequences most efficiently. You must obviously differentiate material that would usually be incorporated in a study editorial from any unprocessed data or additional appendix matter that would not be available. In fact, such matter should not be submitted at all except requested by the instructor.

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- Never confuse figures with tables there is a difference.

## Approach

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Approach:

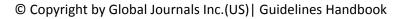
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# INDEX

## Α

Achieved · 4, 13, 23, 24, 25, 28, 52, 55, 59, 68 Algorithm · 17, 21, 23, 25, 27, 39, 40, 41, 43, 45, 46, 50, 52, 55, 56, 57, 58, 59, 60, 61, 62, 64, 65 Analysis · 33, 39, 46, 64 Andreopoulos · 27, 31, 32, 33 Anisotropic · 39, 40, 41, 42, 43, 44, 45, 46 Anisotropic · 39 Appropriate · 21, 28, 41 Architecture · 5, 64, 67, 68, 74, 75 Arithmetic · 15, 16, 17, 19, 20, 21 Artifacts · 9 Averaging · 39

## В

Behavioral · 46 Benelux · 35

# С

Cascaded  $\cdot$  58 Coefficients  $\cdot$  30, 49 Compensated  $\cdot$  11, 13, 15, 23, 24, 25, 27, 31, 32, 34 Compression  $\cdot$  9, 34, 35, 36, 65, 73 Correlation  $\cdot$  49, 51 Corrupted  $\cdot$  39 Cylinder  $\cdot$  3

## D

 $\begin{array}{l} \mbox{Debugged} \cdot 73 \\ \mbox{Decomposition} \cdot 11, 13, 17, 19, 24, 27, 31, 35, 40, 54, 65, \\ 73 \\ \mbox{Denoising} \cdot 39, 40, 46 \\ \mbox{Dimensional} \cdot 9, 10, 11, 24, 28, 32, 33, 35, 36, 40 \\ \mbox{Discarding} \cdot 68, 69 \\ \mbox{Discrete} \cdot 9, 24, 49, 65, 73, 74, 75 \\ \end{array}$ 

# Ε

Extracted · 24, 49, 51, 52, 60

## F

Fractional · 20, 21, 25 Further · 9, 16, 59, 60, 62, 64, 68, 73

## G

Geometric · 3, 52

# Η

Hierarchically · 13 Horizontal · 2, 11, 13, 49

## I

Implementation  $\cdot$  35, 64, 66, 73, 74 Interactive  $\cdot$  1, 5 Interfaces  $\cdot$  1

# М

Modified  $\cdot$  62, 64, 67, 73 Monitoring  $\cdot$  1

## 0

Overwritten · 66

## Ρ

Piecewise · 28 Predisposal · 5 Preserving · 39, 45

## Q

Quality · 9, 10, 11, 13, 30 Quantization · 13, 15, 23, 25, 32, 33, 36

## R

Radioactive · 1, 5 Radwaste · 1 Resolution · 9, 10, 11, 13, 17, 22, 24, 27, 28, 35, 49, 50, 51, 64

# S

Scalability • 9, 11, 13 Schematic • 4, 70 Smoothing • 42, 43 Square • 51, 55 Stationary • 55, 58, 62 Structure • 75 Symposium • 30, 31, 32, 34, 63, 74, 75 Synthesis • 71, 72, 73

# T

Temporal · 9, 11, 13, 14, 23, 35 Transformation · 9, 65

# U

Underlined · 67 Unexpected · 57

## V

Variables · 4, 5 Varying · 33, 42 Visual · 9, 32, 33, 34, 35, 36 Volumetric · 10, 15, 17, 35, 36

# W

Watermarking · 49 Wavelet · 9, 17, 24, 30, 31, 34, 35, 36, 40, 46, 49, 55, 58, 62, 65, 73, 74, 75 William · 33



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