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Kinematics, Localization and Control of Differential Drive Mobile Robot

Sandeep Kumar Malu ^α & Jharna Majumdar ^σ

Abstract- The present work focuses on Kinematics, Localization and closed loop motion control of a differential drive mobile robot which is capable of navigating to a desired goal location in an obstacle free static indoor environment. Two trajectory planning approaches are made (i) the robot is rotated to eliminate orientation error and then translate to overcome distance error (ii) Both rotational and translational motion is given to the robot to overcome orientation and distance error simultaneously. Localization is estimated by integrating the robot movement in a fixed sampling frequency. The control law is based on kinematics model which provides updated reference speed to the high frequency PID control of DC motor. Stability of proposed control law is validated by Lyapunov Criterion. Both experimental and simulation results confirm the effectiveness of the achieved control algorithms and their efficient implementation on a two wheeled differential drive mobile robot using an 8-bit microcontroller.

Keywords: kinematics, odometric localization, PID speed control, differential drive robot, lyapunov stability theory.

I. INTRODUCTION

Autonomous mobile robots have various applications in the field of industry, military and security environment. The problem of autonomous motion planning and control of wheeled mobile robots have attracted lot of research interest in the field of robotics. Consequently engineers working on design of mobile robots have proposed various drive mechanisms to drive such robots. However the most common way to build a mobile robot is to use two-wheel drive with differential steering and a free balancing wheel (castor). Controlling the two motors independently make such robots to have good manoeuvring and work well in indoor environment. Mobile robots with such drive systems are a typical example of non-holonomic mechanisms due to the perfect rolling constraints on a wheel motion (no longitudinal or lateral slipping).

An asymptotic stable controller using Backstepping method for posture tracking and its stability has been validated [1] [7]. An Adaptive Controller [2] to compensate errors can further improve its stability. A non-linear control design [6] using feedback linearization can provide better performance

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than conventional linearized controller. Many authors [2] [3] have proposed methods to reduce odometry error caused by kinematic imperfection. A different approach of localization using RFID technology [5] is efficient, fast and cheap.

A neural network [4] based reactive navigation algorithm for mobile robot in unstructured environment while avoiding obstacles is found to be optimized in computation.

In the present paper, kinematics model and localization using optical encoder coupled with the DC motor of a differential drive robot is presented. This model itself is used as a motion controller in a closed loop control scheme. In the absence of workspace obstacles, the basic motion tasks assigned to wheeled mobile robots may be reduced to moving between two robot postures and following a given trajectory.

The paper is organized as follows: In section II, basic equations of Kinematics and Motion Model of the robot are reported. The Localization in indoor environment using optical encoder coupled with the motor is described in brief. In section III, Control law is proposed and its stability analysis is carried out based on Lyapunov theory. In section IV, PID speed control of motor is presented. Section V includes some simulation results. Section VI highlights implementation strategies of control and localization using an 8-bit ATmega 32 microcontroller while Section VI contains conclusion and future work.

II. KINEMATICS OF DIFFERENTIAL DRIVE ROBOT

a) Motion Model

Let Inertial Reference Frame is $\{X_I, Y_I\}$ and Robot Frame is $\{X_R, Y_R\}$. The Robot position $P[x \ y \ \theta]$ is expressed in cartesian co-ordinate system of inertial frame. The relationship between Inertial and Robot frames is represented using basic transformation matrix as follows:

$$\begin{aligned} \dot{\epsilon}_R &= R(\theta)\dot{\epsilon}_I \\ &= R(\theta) \cdot [\dot{x} \ \dot{y} \ \dot{\theta}]^T \end{aligned} \quad (1)$$

Where,

$$R(\theta) = \begin{bmatrix} \cos \theta & \sin \theta & 0 \\ -\sin \theta & \cos \theta & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

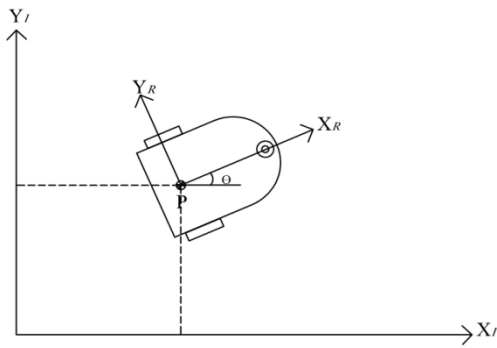


Fig. 1 : Representation of Robot on a Cartesian coordinate system

The robot under consideration is a two wheeled differential drive robot, where each wheel is driven independently. Forward motion is achieved when both wheels are driven at the same rate, turning right is achieved by driving the left wheel at a higher rate than the right wheel and vice-versa for turning left. This type of mobile robot can turn on the spot by driving one wheel forward and second wheel in opposite direction at the same rate. Third wheel is a castor wheel needed for the stability of mobile robot.

Each individual wheel contributes to the robot's motion and at the same time, imposes constraints on robot motion. It is assumed that the wheels of the robot do not slide. It is expressed by Non Holonomic Constraint.

$$\dot{x} \sin \theta - \dot{y} \cos \theta = 0 \tag{2}$$

Also the measure of the traveled distance by each wheel is not sufficient to calculate the final position of the robot. One needs to know how this movement is executed as a function of time. This can be illustrated in Fig. 2.

$$s_{1R} = s_{2R}; s_{1L} = s_{2L} \text{ and } S_1 = S_2$$

$$x_1 \neq x_2; y_1 \neq y_2$$

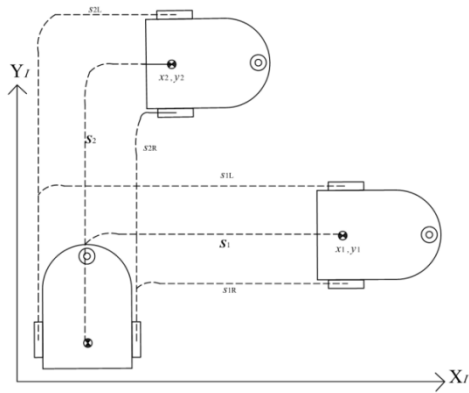


Fig. 2 : Dependence of Robot position on its velocities as function of time

The actual robot motion commands are the angular velocities w_R and w_L of the right and left wheel respectively, rather than robot driving and steering

velocities v and ω . First consider the contribution of each wheel's spinning speed to the translation speed at P in the direction of $+X_R$. If one wheel spins ($v_1 = rw_R$) while the other wheel contributes nothing and is stationary ($v_2 = 0$), Since P is halfway between the two wheels, it will move instantaneously with half the speed i.e. $v_X = (1/2)rw_R$. In a differential drive robot, these two contributions can simply be added to calculate the v_X . Consider, for example, a differential robot in which each wheel spins with equal speed but in opposite directions. The result is a stationary, spinning robot. As expected, v_X will be zero in this case. The value of v_Y is even simpler to calculate. Neither wheel can contribute to sideways motion in the robot's reference frame, hence v_Y is always zero. Finally, we must compute the rotational component w of the robot. Once again the contributions of each wheel can be computed independently and just added. Consider the right wheel (we will call this wheel 1). Forward spin of this wheel results in counter clockwise rotation at point P . Recall that if wheel 1 spins alone, the robot pivots around wheel 2. The rotation velocity w_1 at P can be computed because the wheel is instantaneously moving along the arc of a circle of radius d . $w_1 = \frac{r}{d}w_R$. The same calculation applies to the left wheel, with the exception that forward spin results in clockwise-rotation at point P :

$$w_2 = -\frac{r}{d}w_L.$$

Mapping between Robot velocities to wheel velocities is given as follows:

$$v = v_X = v_1 + v_2 = r\left(\frac{w_R + w_L}{2}\right); v_Y = 0$$

$$\omega = w_1 + w_2 = r\left(\frac{w_R - w_L}{d}\right) \tag{3}$$

Where, r = radius of wheel and d = axial distance between wheels.

b) Kinematic Equations

Kinematics is the most basic study of how mechanical systems behave. In mobile robotics, we need to understand the mechanical behaviour of the robot both in order to design appropriate mobile robots for desired tasks and to understand how to build control software.

Consider a differential drive robot at some arbitrary position $P(q_c) [x_c, y_c, \theta_c]$ which has a non-zero distance with the goal position $R(q_d) [x_d, y_d, \beta]$ defined with respect to global inertial frame.

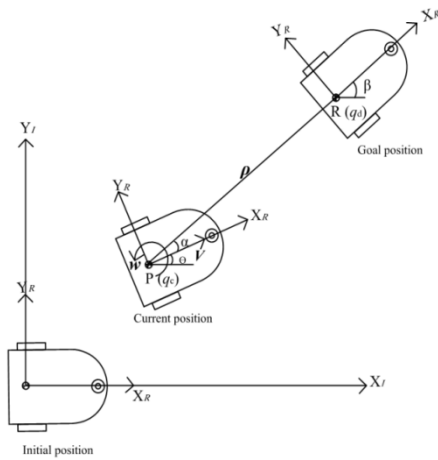


Fig. 3 : Error vectors of robot's position and orientation

The coordinate system of the robot is governed by combined action of both the linear velocity v and the angular velocity w . Using geometrical assumption the robot cartesian is given by:

$$\begin{aligned}\dot{x} &= v \cos \theta \\ \dot{y} &= v \sin \theta \\ \dot{\theta} &= w\end{aligned}\quad (4)$$

Where $v \cos \theta$ and $v \sin \theta$ are the components of v along its X and Y axes and x , y and orientation θ are measured with respect to the reference Inertial Frame.

Same way position of the robot can also be represented in terms of polar coordinates involving error distance $\rho > 0$ as:

$$\begin{aligned}\dot{\rho} &= -v \cos(\beta - \theta) = -v \cos \alpha \\ \dot{\beta} &= v \frac{\sin \alpha}{\rho} \\ \dot{\theta} &= w\end{aligned}$$

Now, $\alpha = \beta - \theta$ be the angle measured between robot axes frame $[X_R, Y_R]$ and the distance vector frame ρ also $\rho = \sqrt{(x_d - x_c)^2 + (y_d - y_c)^2}$ as per distance formula between two coordinates system. Finally we conclude

$$\begin{aligned}\dot{\rho} &= -v \cos \alpha \\ \dot{\alpha} &= -w + \frac{v \sin \alpha}{\rho} \\ \dot{\beta} &= \frac{v \sin \alpha}{\rho}\end{aligned}\quad (5)$$

Kinematics equation based on polar coordinates (5) is valid when $\rho \neq 0$ and above set of equations will be employed in establishing feedback control law for robot manoeuvring as discussed later in section III.

c) Odometric Localization

Localization is one of the most fundamental aspects of a mobile robot. All mobile robot system has to answer the fundamental question, which is "Where Am I", i.e. the current location & orientation of the robot

has to be obtained, so that the robot can easily move from source to destination. There are number of localization techniques with respect to mobile robot, however in the present work we have used dead reckoning method for localization. Dead reckoning method uses odometry to measure the movement of the robot. In the present work, we obtain the data from an incremental encoder (odometry), which is fitted along with a motor of the mobile robot. Incremental encoders measure the rotation of the wheels, which in turn, calculates robot position and orientation using integration approaches of kinematic model over $[t_k, t_{k+1}]$.

Assuming Robot configuration q_k $[x_k \ y_k \ \theta_k]$ and constant velocity inputs V_k and w_k are known at discrete time t_k , then using Euler integration

$$\begin{aligned}x_{k+1} &= x_k + v_k T_s \cos \theta_k \\ y_{k+1} &= y_k + v_k T_s \sin \theta_k \\ \theta_{k+1} &= \theta_k + w_k T_s\end{aligned}\quad (6)$$

where, $v_k T_s = \Delta s$ and $w_k T_s = \Delta \theta$
 $T_s = t_{k+1} - t_k$

The reconstruction of the current robot configuration is based on the incremental encoder data (odometry). Let $\Delta \phi_R$ and $\Delta \phi_L$ be the no. of wheel rotations measured during the sampling time T_s by the encoders. Linear and angular displacements of the robot is given as

$$\Delta s = \frac{r}{2}(\Delta \phi_R + \Delta \phi_L), \quad \Delta \theta = \frac{r}{d}(\Delta \phi_R - \Delta \phi_L)\quad (7)$$

Where, r = radius of wheel and d = axial distance between wheels.

For a differential-drive robot the position can be estimated starting from a known position by integrating the movement (summing the increment travel distances). The estimate of robot configuration at time t_k is computed as:

$$\begin{bmatrix} x_k \\ y_k \\ \theta_k \end{bmatrix} = \begin{bmatrix} x_{k-1} \\ y_{k-1} \\ \theta_{k-1} \end{bmatrix} + \begin{bmatrix} \cos \theta & 0 \\ \sin \theta & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} \Delta s \\ \Delta \theta \end{bmatrix},\quad (8)$$

Robot localization using the above odometric prediction (commonly referred to as *dead reckoning*) is accurate enough in the absence of wheel slippage and backlash.

III. CONTROL LAW AND STABILITY

The control algorithm must now be designed to drive the robot from its current configuration; say (x_c, y_c, θ) to the goal position (x_g, y_g, β) . Here the objective is to find a control $u = [v \ w]$ so that the robot's goal position is reached in finite interval of time. The proposed control law is state dependent i.e. $[v \ w] = f(\rho, \alpha, \beta)$ which guarantees the state to be asymptotically driven to $[0, 0, \beta]$ without attaining $\rho = 0$

in finite time. One of the most commonly used methods to study the asymptotic behaviour is based on the Lyapunov stability theory. Consider a simple positive definite quadratic form of Lyapunov function:

$$\begin{aligned} V &= V_1 + V_2 \\ &= \frac{1}{2}\rho^2 + \frac{1}{2}\alpha^2 \end{aligned} \quad (9)$$

Where the parameters V_1, V_2 represent one half of the squared weighted norms of the "distance error vector" ρ and "orientation error vector" α exhibited by the robot between its current position and goal position defined with respect to the Reference Inertial Frame. Its time derivative is given by:

$$\dot{V} = \dot{V}_1 + \dot{V}_2 = \dot{\rho}\rho + \alpha\dot{\alpha}$$

Using kinematics equation (5),

$$\dot{V} = \rho(-v \cos \alpha) + \alpha(-w + \frac{v}{\rho} \sin \alpha) \quad (10)$$

From the equation (10), the first term can be made non-positive by letting the linear velocity of the form:

$$v = K_\rho \rho \cos \alpha \quad K_\rho > 0 \quad (11)$$

then,

$$\begin{aligned} \dot{V}_1 &= \rho(-K_\rho \rho \cos^2 \alpha) \\ &= -K_\rho \rho^2 \cos^2 \alpha \leq 0 \end{aligned} \quad (12)$$

This means that \dot{V}_1 term is always non-increasing in time and consequently, since it is asymptotically converges to non-negative finite limit.

Similarly the second term \dot{V}_2 can be made be non-positive by letting the angular velocity w take the form of:

$$w = K_\alpha \sin \alpha \cos \alpha + K_\alpha \alpha \quad K_\alpha > 0 \quad (13)$$

Then,

$$\begin{aligned} \dot{V}_2 &= \alpha(-K_\rho \sin \alpha \cos \alpha - K_\alpha \alpha + \frac{K_\rho \rho \sin \alpha \cos \alpha}{\rho}) \\ &= -K_\alpha \alpha^2 \leq 0 \end{aligned} \quad (14)$$

Finally leading to the following expression for the time derivative of the Lyapunov function V

$$\dot{V} = \dot{V}_1 + \dot{V}_2 = -K_\rho \rho^2 \cos^2 \alpha - K_\alpha \alpha^2 \leq 0 \quad (15)$$

The result in (15) is a negative semi-definite form. By applying Barbalat's Lemma, it follows that \dot{V} necessary converges to zero for increasing time; thus in turn implying convergence of the state vector $[\rho, \alpha, \beta]$ to $[0, 0, \beta]$. Hence it can be concluded that control vectors expressed by (11) and (13):

$$v = K_\rho \rho \cos \alpha \text{ and } w = K_\rho \sin \alpha \cos \alpha + K_\alpha \alpha,$$

makes the robot motion behaviour to be smooth and stable.

The overall feedback control architecture for Go-to-Goal motion can be summarized in Fig. 4.

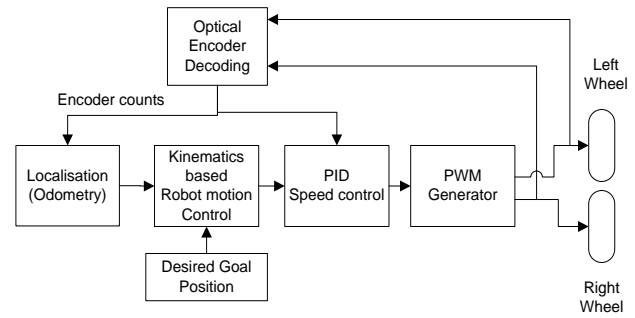


Fig. 4: Block diagram of proposed Kinematics based motion controller

IV. LOW LEVEL PID SPEED CONTROL

The speed of DC motor can be adjusted to a great extent so as to provide easy control and high performance. There are several conventional and numeric controller types intended for controlling the DC motor speed. Recently, many modern control methodologies such as nonlinear control [8], optimal control [9], variable structure control [10] and adaptive control [11] have been extensively proposed for DC motors. However, these approaches are either complex in theory or difficult to implement. PID controller algorithm involves three parameters denoted P , I and D interpreted in terms of time. P depends on present error, I on the accumulation of past errors, and D is a prediction of future errors. PID control with its three term functionality covering both transient and steady-states response, offers the simplest and yet most efficient solution for many real world control problems. In spite of the simple structure and robustness of this method, optimally tuning gains of PID controllers have been quite difficult.

The electric equivalent circuit of the armature and the free-body diagram of the rotor are shown in the Fig. 5 below

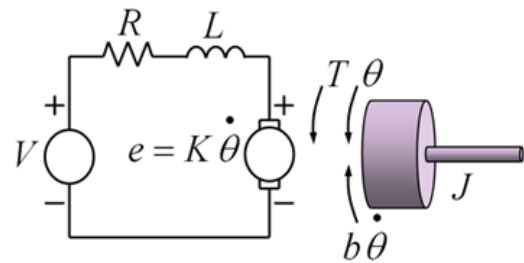


Fig. 5: Circuit Diagram of DC motor

The dynamic equations and the open loop transfer function of the DC motors are:

$$\begin{aligned} s(Js + b)\theta(s) &= KI(s) \\ (Ls + R)I(s) &= V(s) - Ks\theta(s) \\ P(s) &= \frac{s\theta(s)}{V(s)} \\ P(s) &= \frac{K}{(Js + b)(Ls + R) + K^2} \left[\frac{\text{rad/sec}}{V} \right] \end{aligned} \quad (16)$$

where,

- J = moment of inertia of the rotor
 - b = motor viscous friction constant
 - K_e = electromotive force constant
 - K_t = motor torque constant
 - R = electric resistance
 - L = electric inductance
 - V = voltage across motor terminals
 - I = current flowing in the circuit
- Here, $K_t = K_e = K$.

And the system schematic looks like:

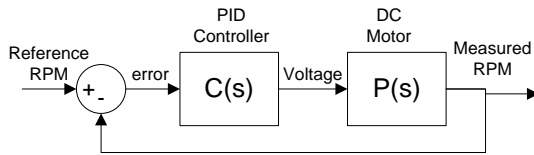


Fig. 6 : PID for DC motor speed control

The PID control design criteria are (i) less settling time (<1 s) (ii) overshoot less than 5% (iii) steady state error less than 1%. The PID speed control design is incorporated into the system. The transfer function for a PID controller given is by:

$$C(s) = K_p + \frac{K_i}{s} + K_d s \quad (17)$$

Where, K_p , K_i and K_d are gains. PID is simulated in Matlab and its response is plotted. Further, gain values are tuned manually to obtain desired response.

V. SIMULATION RESULTS

Simulation results of DC motor characteristics incorporating PID for unit step input at different gains are shown in figure 7:

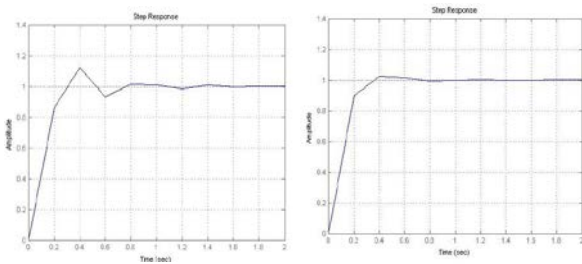


Fig. 7 : Unit step response at a) $K_p=0.2, K_i=39, K_d=0.25$; b) $K_p=5, K_i=8, K_d=1.5$

Simulation results are performed to illustrate the effectiveness of the proposed control law. Here sets of five position coordinates $[0, 0, 0], [5, 0, 90^\circ], [5, 5, 180^\circ], [0, 5, -90^\circ]$ and $[0, 0, 0]$ are considered to traverse a square path. Both the approaches of reaching the goal position are depicted below.

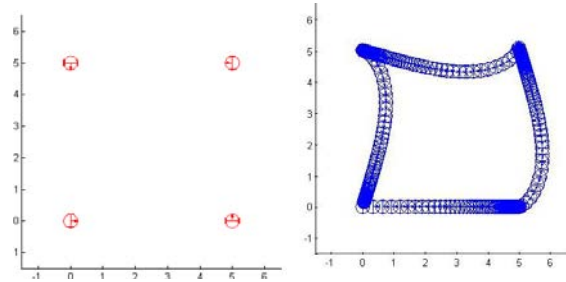


Fig. 8 : Posture tracking with simultaneous translation and rotation motion

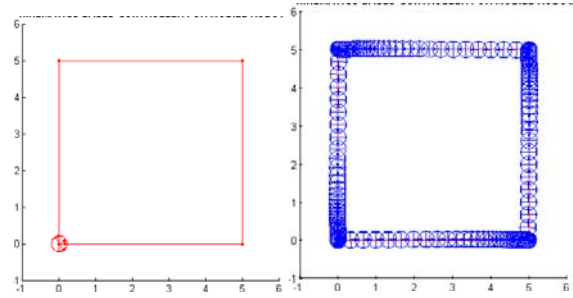


Fig. 9 : Posture tracking with first rotate then translate motion

From the plots, it can demonstrated that the proposed control law show good convergence in both the cases and is equally applicable in all conditions of robot movement between initial and goal position.

VI. IMPLEMENTATION

The robot which we used for our experiments is fabricated in-house. It has a differential-drive mechanism consisting of high torque Dynaflex DC motors each of which is equipped with high resolution Jayashree-15S optical quadrature shaft encoder for precise position and speed feedback to controller. The robot uses two 12V DC batteries in series to power motors, while 5V for Microcontroller development board. The power electronics module used to drive the DC motor is Super Hercules 9V-24V, 15A Motor Driver from Nex-robotics. The algorithm of localization and control has been implemented on 8-bit Atmega32 Microcontroller.

The output signal from one channel of encoder is fed to the rising edge enabled external interrupt pin. On interrupt, the status of other channel decides the direction and equivalent rotation counts of robot motion. These counts on accumulation at fixed sampling frequency (in our case 33 Hz) resolve robot position estimation using the expression for odometric localization explained in section II. Also these counts acts as feedback to PID speed control executing at 100 Hz frequency.

The control algorithm runs at 33 Hz wherein distance error vector and orientation error vector between robot instantaneous position and goal (target) position are calculated and robot control vectors viz.

linear velocity v and angular velocity w are obtained. Two approaches to reach the goal position have been implemented. In first one, robot is made to rotate until orientation error is eliminated and then translated to overcome distance error. While in other method, robot exhibits both linear and angular velocities to overcome distance and orientation error simultaneously.

Using Kinematics motion models the linear and angular velocities of the robot are transformed to right and left wheel speeds and fed as reference speed to PID speed control. The upper limit of robot motion is set to 40 RPM.



Fig. 10 : Experimental Differential Drive Robot

VII. CONCLUSION AND FUTURE WORK

In this paper, Kinematics based feedback path controller for a differential drive mobile robot has been presented. Both the approach of Go-to-Goal motion is implemented using 8-bit Atmega 32 microcontroller. The robot motion behaviour in real time is in line with the behaviour observed in simulation. Also the proposed control law is validated as per Lyapunov stability criterion. The control algorithm proposed in this paper will be extended for human following application, where set points (goal locations) are updated regularly through computer vision algorithm of detecting and tracking human.

The present work will be extended to include obstacle avoidance in Go-to-Goal motion and dynamics consideration of robot for control which can enhance better stabilization. Further, Umbmark calibration needs to be performed to avoid localization error due to irregular wheel diameter and wheel axial alignment. Also Attitude compensation has to be incorporated in odometric localization whenever robot travels in irregular surface.

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