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# General Engineering

Frequency Hopping Signal

Seasonality Smoothing Method

Highlights

Low Probability of Intercept

Spectrogram and the Scalogram

**Discovering Thoughts, Inventing Future** 

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# GLOBAL JOURNAL OF RESEARCHES IN ENGINEERING: J General Engineering

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# Hyperbolic Metamaterial Interface: Propagation of Surface Waves

# By T. Gric, M. Cada & J. Pistora

Dalhousie University, Canada

Abstract- Herein we study propagation of surface waves at a boundary of an isotropic media and a multilayered hyperbolic metamaterial. The structure dispersion is discussed for various cases of a hyperbolic metamaterial. It is demonstrated that it is possible to tune the frequency range of surface waves by varying the thickness of dielectric sheets. It is also shown that this frequency range can be broadened by decreasing the thickness of the dielectric in the metal-dielectric compound or by replacing the isotropic media with a metal. It is also shown that the mentioned effect can be achieved by increasing the doping concentration of the semiconductor if the metamaterial/semiconductor structure is considered. The effective medium approximation is used and its validity in the long-wavelength limit is justified by investigating the electromagnetic field variations over one period of the proposed structure.

Keywords: surface wave; metamaterial; drude metal; surface plasmons; terahertz.

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# Hyperbolic Metamaterial Interface: Propagation of Surface Waves

T. Gric  $^{\alpha}$ , M. Cada  $^{\sigma}$  & J. Pistora  $^{\rho}$ 

Abstract- Herein we study propagation of surface waves at a boundary of an isotropic media and a multilayered hyperbolic metamaterial. The structure dispersion is discussed for various cases of a hyperbolic metamaterial. It is demonstrated that it is possible to tune the frequency range of surface waves by varving the thickness of dielectric sheets. It is also shown that this frequency range can be broadened by decreasing the thickness of the dielectric in the metal-dielectric compound or by replacing the isotropic media with a metal. It is also shown that the mentioned effect can be achieved by increasing the doping concentration of the semiconductor if the metamaterial/semiconductor structure is considered. The effective medium approximation is used and its validity in the long-wavelength limit is justified by investigating the electromagnetic field variations over one period of the proposed structure.

*Keywords:* surface wave; metamaterial; drude metal; surface plasmons; terahertz.

### I. INTRODUCTION

The waves at an interface between an isotropic medium and a uniaxial crystal characterized by its effective permittivities have been first demonstrated in [1]. The existence of these surface waves in some material examples is studied in [2], discussing the challenge posed by their experimental observation. The novelty and potential importance of Dyakonov waves for integrated optics applications were described in a stream of papers [3-5].

In near-infrared and visible wavelengths, the nanolayered behavior of metal-dielectric (MD) compounds is similar to plasmonic crystals. In this case a simplified description of the medium by using the long-wavelength approximation can be justified; the homogenization of the structured metamaterial can also be employed [6-8]. It is interesting to note that the second-rank tensor representing the medium permittivity includes elements of opposite signs, thus yielding extremely anisotropic metamaterials under certain conditions [9, 10]. This category of nanostructured media opens the wide avenues for the practical applications from biosensing to fluorescence engineering [11].

The existence of surface waves when dealing with anisotropic media possessing the indefinite permittivity was reported for the first time in [12]. This study was dedicated mainly to surface waves enabling sub-diffraction imaging in magnifying superlenses [13], where the surface waves exist at an interface between a metal and an all-dielectric birefringent metamaterial. Dealing with hyperbolic media, however, the authors concluded only with an indefinable analysis of surface waves.

In this paper we retake the task and perform a thorough analysis of surface waves traveling in infinite MD lattices. Our approach is to use the effective-medium approximation.



*Fig.* 1 : The model under study consisting of an infinite MD lattice (x < 0) and an isotropic material (x > 0). The periodic structure contains a Drude metal and a dielectric.

The investigated structure is shown in Fig. 1. It should be mentioned that an infinite periodic model consisting of alternating layers of a metal and a conventional dielectric is placed on the left of the homogeneous dielectric. The effective dielectric tensor components parallel ( $\varepsilon_{\parallel}$ ) and perpendicular ( $\varepsilon_{\perp}$ ) to the anisotropy axis are described in [14] as:

Е

$$\frac{1}{\varepsilon_{\parallel}} = \frac{d_m / \varepsilon_m + d_d / \varepsilon_d}{d_m + d_d} \tag{2}$$

where  $d_m$  ( $d_d$ ) is the thickness and  $\varepsilon_m$  ( $\varepsilon_d$ ) is the dielectric constant of the metallic (dielectric) component. The effective medium approximation in Eqs. (1) and (2) is

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valid in the long-wavelength limit. However, the field varies significantly on the scale of one period due the excitation of surface plasmon (SP) polaritons at metal/dielectric interfaces. Therefore, the approximation in Eqs. (1) and (2) may not be applicable in some spectral ranges.

### III. NUMERICAL ANALYSIS OF DISPERSION CHARACTERISTICS

Herein we discuss the numerical solution to the dispersion relation [15], either obtained for metamaterial/air interface or for metamaterial/metal and metamaterial/semiconductor interfaces.

### a) Metamaterial/Air interface

The dispersion relation [15] is graphically represented in Fig. 2 for the MD crystal displayed in Fig. 1 at x < 0, considering different widths of the dielectric  $d_d$  and assuming, that we are dealing with metamaterial/air interface.



*Fig.* 2 : Dispersion curves of TM modes at a metamaterial/air interface for different widths of the dielectric layer  $d_d > 0$  (red dots) in a Drude metal/dielectric compound. The light line is shown (dashed line), and  $d_m = 35$  nm.

Fig. 2 shows the dispersion of surface waves at a metamaterial/air interface. The case  $(d_d > 0)$ corresponds to classic surface waves well known for conductive interfaces [16]. Thus, this example illustrates the verification of our methodology [15]. In addition, the light line in vacuum is plotted in Fig. 2 as dashed line. Our calculated data corresponds to the data of [17] where also a complete discussion of the surface wave characteristics in this case can be found.

As can be seen from Fig. 2, the frequency range for surface waves can be controlled by the fill factors of the air and metal sheets in the MD compound.

If we decrease the thickness of the dielectric  $d_d$ , the dispersion curve moves to a lower frequency. The dependence of the frequency range for the surface

wave existence on the thickness of the dielectric layer provides additional degree of freedom to control surface waves.

Fig. 3 shows the profiles of the magnetic field along the *x* axis for the investigated case, i. e. metamaterial/air. The magnetic field has been obtained when  $\beta = 4 \cdot 10^6$  1/m. The wave field is tightly confined near *x* = 0 for all the thicknesses of the dielectric *d*<sub>d</sub>.



*Fig.* 3 : Magnetic field (normalized) for different values of parameter  $d_d$  for SP modes at metamaterial/air inter face.

The conducted analysis shows, that the frequency range for the surface wave existence seems to be narrow. Thus, we shall leave the geometry unchanged but assume that instead of an interface between a metamaterial and air we will be dealing with an interface between a metamaterial and a metal. Doing so, we can extend the frequency range for the surface wave existence. It should be noticed, that each medium is capable of supporting propagating surface waves separately.

#### b) Metamaterial/Metal interface

The dispersion curves for the matamterial/metal structure are presented in Fig. 4. The calculations were performed using the following parameters for bulk silver:  $\omega_p = 2297.09$  THz,  $\varepsilon_{\infty} = 5.2$  [18]. As in previous cases we discuss the effects of the thickness of the dielectric  $d_d$  on the dispersion curve. It should be noticed that the upper and the lower limits move to lower frequencies when  $d_d$  is decreased. However the lower limit moves faster than the upper one. The mentioned issue leads to a broader frequency range for the surface wave existence.



*Fig.* 4 : Dispersion curves of TM modes at a metamaterial/metal interface for different widths of the dielectric layer  $d_d > 0$  (red dots) in a Drude metal/dielectric compound. The light line is shown (dashed line), and  $d_m = 35$  nm.



*Fig.* 5 : Magnetic field (normalized) for different values of parameter  $d_d$  for SP modes at metamaterial/air interface.

Fig. 5 shows the profile of the magnetic field along the *x* axis for the case under the study. It is interesting to point out, that  $\beta = 6 \cdot 10^6 \text{ 1/m}$ .

It should be mentioned, that the wave field is tightly confined near x = 0 for all the thicknesses of the dielectric  $d_d$ . It is interesting to note that in the case of a metamaterial/air interface (Fig. 3) the tighter confinement near the boundary x = 0 is exhibited in the metamaterial while in the case of a metamaterial/metal interface the tighter confinement is in the metal.

### c) Metamaterial/Semiconductor interface

It is interesting to notice, that he case of heavy doped Si is considered, assuming that the doping level is  $N_1=5\cdot10^{19}$  cm<sup>-3</sup> [19]. An average effective mass  $m_1$  for electrons is  $0.26m_0$  with  $m_0$  being the free-electron mass, and  $\varepsilon_{\infty 1} = 11.68$ .

Figure 6 shows the longitudinal propagation constant  $\beta$  as a function of the propagation frequency, for different values of dielectric layers  $d_{q}$ , employed in the MD lattice.



*Fig.* 6 : Dispersion curves of TM modes at a metamaterial/semiconductor interface for different widths of the dielectric layer  $d_d > 0$  (red dots) in a Drude metal/dielectric compound. The light line is shown (dashed line), and  $d_m = 35$  nm. Points *A*, *B*, *C*, *D*, *E* are used in Fig. 8.

It is interesting to notice, that the behavior of the dispersion curves in the investigated case is consistent with the effect of  $d_d$  on the frequency range of the negative  $\varepsilon_{\parallel}$  [20].

To date it is well known, that the presence of semiconductor in the structure can tune the properties of the investigated system in the easy way. Thus in Fig. 7 we present the dispersion curves of the surface wave dealing with the different concentrations of the Si.



*Fig.* 7 : Dispersion curves of TM modes at a metamaterial/semiconductor interface for different concentration  $N_1$  of Si. The light line is shown (dashed line), and  $d_d = d_m = 35$  nm. Points A, B, C, D are used in Fig. 9.



Fig. 8 : Magnetic field (normalized) of SP modes at metamaterial/semiconductor

Interface for the points A, B, C, D and E highlighted in Fig. 2, when that the doping level is  $N_1 = 5.10^{19} \text{ cm-3}.$ 

Fig. 8 shows the profile of the magnetic field along the x axis for the investigated case, i. e. metamaterial/semiconductor. Fig. 8 shows the magnetic field for the points when  $\beta = 1.5 \cdot 10^6 \text{ 1/m}$ . The tight confinement of the wave field is observable near x = 0for all the thicknesses of the dielectric  $d_d$ . It is interesting to note that the tighter confinement near the boundary x = 0 is exhibited in the semiconductor.

Moreover, it is of particular interest to plot the profile of the magnetic field along the x axis, when the doping level of the silicon changes. The results are depicted in Fig. 9 being  $d_d = d_m = 35$  nm.



Fig. 9 : Magnetic field (normalized) of SP modes at metamaterial/semiconductor interface for the points A, *B*, *C* and *D* highlighted in Fig. 3, if  $d_d = d_m = 35$  nm.

It is interesting to notice, that the lowest confinement in this case is produced for the lowest doping concentration of silicon.

It is valuable to confirm the presence of SP waves for all the investigated cases. In the insets of Figs. 2, 4 and 6 we show the nature of the magnetic fields for all the investigated cases.

#### IV. CONCLUSIONS

The surface waves at the boundary of hyperbolic metamaterial have been studied for three different cases: metamaterial/ air, metamaterial/metal, metamaterial/semiconductor. The dependences of the longitudinal propagation constant on the propagation frequency have been examined. It is established that one can tune the frequency range of surface waves by varying the thickness of dielectric sheets. It is also revealed that this frequency range can be broadened by three different manipulations:

- decreasing the thickness of the dielectric in the metal-dielectric compound;
- dealing with the metamaterial/metal structure; •
- increasing the doping concentration the • of semiconductor. considering the if metamaterial/semiconductor interface.

#### V. **ACKNOWLEDGMENTS**

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# Performance Assessment of SARIMA Model with Holt-Winter's Trend and Additive Seasonality Smoothing Method on forecasting Electricity Production of Australia an Empirical Study

By Md. Matiur Rahman Molla, S.M. Nuruzzaman, Dr. M. Sazzad Hossain & Md. Shohel Rana

Islamic University, Bangladesh

Abstract- Australia is a leading developed country which is indispensable a proper planning and management of power generation. To take a unique planning decision forecasting of electricity production is badly in need so that electricity generation copes with the demand of the electricity smoothly. The main task of this study is to assess the performance of two time series models in forecasting electricity generation in Australia. Two time series forecasting methods such as ARIMA and Holt-Winter's additive trend and seasonality smoothing methods are considered. Applying Theil's U-statistic as the key performance measure, the study concludes that Holt-winter's method is more appropriate model.

Keywords: electricity production, seasonal ARIMA, smoothing, forecasting, time series analysis.

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Md. Matiur Rahman Molla<sup> $\alpha$ </sup>, S.M. Nuruzzaman<sup> $\sigma$ </sup>, Dr. M. Sazzad Hossain<sup> $\rho$ </sup> & Md. Shohel Rana<sup> $\omega$ </sup>

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*Keywords:* electricity production, seasonal ARIMA, smoothing, forecasting, time series analysis.

### I. INTRODUCTION

t present electricity has become a first and precondition foremost of macroeconomic development of a territory. Each day, electricity plays key rolein keeping homes and business running smoothly, powers transportation that take people work, school and other places, and supplies electricity to appliances in all sectors. The demand of electricity especially in for industrial sector need not to say. Without electricity not only a single day but also a moment is unimaginable. A country's economic growth directly related to electricity production. That's why sustainable electricity production badly in needs to fulfill the demand of households as well as industry and communication sectors. To manage such kind of demand of electricity a country's power development board has to take sophisticated decision to produce electricity that can cope with demand with supply of energy.

Being a developed country monthly electricity production of Australia is a seasonal and trending behavior. So, electricity production authority of Australia should take plan for proper management of production with demand. To overcome uncertainty of future production smoothing or forecasting approach time series analysis is the most applied method. For

Author  $\alpha \sigma \rho \omega$ : Islamic University, Bangladesh. e-mail: matiur508@gmail.com predicting Australian electricity production, we will use conventional smoothing methods and well known ARIMA modeling. Hence we want to show the comparative performance of referred model. This paper is divided into six sections. The section one of this study is the introductory part. The second section of the study will present forecasting approach where we present stationarity, Holt's-Winter trend and additive seasonality, Box-Jenkins methodology SARIMA modeling and accuracy measurement approach. Section three is the empirical data analysis and forecasting while sections four is the accuracy measurement and finally conclusion

*Basic Terminologies:* The following keywords are used throughout the research approach.

*Stationarity:* Stationarity means that there is no growth or decline in the data. The data must be horizontal along the axis. A time series is said to be stationary if its mean and variance are constant over time and the value of the covariance between the two time periods depends only on the distance or gap or lag between the two time periods and not the actual time is computed.

Suppose  $y_t$  be a stochastic time series then,

$$E(y_t) = \mu$$
  
var $(y_t) = E(y_t - \mu)^2 = \sigma^2$ 

Holt's-Winter's trend and additive seasonality method

The basic equations of Holt-Winters' trend and additive seasonality method are as follows:

Level	$L_t = \alpha(Y_t - S_{t-s}) + (1 - \alpha)(L_{t-1} + b_{t-1})$
Trend:	$b_t = \beta (L_t - L_{t-1}) + (1 - \beta) b_{t-1}$
Seasonal:	$S_t = \gamma(Y_t - L_t) + (1 - \gamma)S_{t-s}$
Forecast:	$F_{t+m} = L_t + b_t m + S_{t-s+m}$

Where s is the length of seasonality (e.g., number of months or quarters in a year),  $L_t$  represents the level of the series,  $b_t$  denotes the trend,  $S_t$  is the seasonal component, and  $F_{t+m}$  is the forecast for m period ahead.

Box-Jenkin's methodology and ARIMA modeling

The general ARIMA model proposed by Box and Jenkins (1970) is written as ARIMA (p, d, q) but when the characteristic of the data is seasonal behavior then it said to be SARIMA. And the seasonal ARIMA model is written as very formal notation like this

ARIMA 
$$(p, d, q) \times (P, D, Q)_m$$

Non-seasonal Seasonal Part of the model part of the model AR: p = order of the autoregressive part

*I*:d = degree of differencing involved

MA: q = order of the moving average part

m = number periods per season

The basis of the Box-Jenkins modeling in time series analysis is summarized the following figure and consist of three phases: identification, estimation and testing, and application.



Figure 1.1 : Schematic representation of the Box-Jenkins methodology for time series modeling

Assessment Approach: The validity of the forecasting in time series analysis can be assessedvia couples of approaches such as Mean error (ME), root mean square error (RMSE), mean absolute error (MAE), mean percentage error (MPE), mean absolute percentage error (MAPE), mean square error (MSE), Mean absolute scaled error (MASE) and The il's U statistic.

*Percentage Error:* If  $Y_t$  is the actual observation for time period t and  $F_t$  is the forecast for the same period, then the percentage error is defined as

$$PE_t = (\frac{Y_t - F_t}{Y_t}) \times 100$$

Mean Percentage Error (MPE):

$$MPE = \frac{1}{n} \sum_{t=1}^{n} PE_t$$

Mean Absolute Percentage Error:

$$MAPE = \frac{1}{n} \sum_{t=1}^{n} |PE_t|$$

If smaller the any above index is considered the better forecasting technique.

Theil's U Statistic: It is defined as follows:

$$U = \sqrt{\frac{\sum_{t=1}^{n-1} (FPE_{t+1} - APE_{t+1})^2}{\sum_{t=1}^{n-1} (APE_{t+1})^2}}$$

Where  $FPE_{t+1} = \frac{F_{t+1} - Y_t}{Y_t}$  (forecast relative change)

And 
$$APE_{t+1} = \frac{Y_{t+1} - Y_t}{Y_t}$$
 (actual relative change)

If U < 1: the forecasting technique being used is better than the naïve method. The smaller the U statistic is considered the better forecasting technique.

### II. Empirical Results



*Fig: 1. 2 :* Australian monthly electricity productions from January, 1980 to August, 1995

Now, it is revealed to us that the above figure of monthly Australian electricity production exhibits an additive seasonal and steadily increasing trend pattern. Obviously the data series is non-stationary.

Before model building first and foremost task is to differentiate the original data first difference as well as seasonal first difference.



*Figure:* Time series plot of first difference of the original data

Obviously, first difference of original time series data is now of stationary.

The model SARIMA (0, 1, 1) (0, 1, 2) [12] has chosen on the basis AIC & BIC criterion. The minimum of AIC & BIC that model is taken as the ultimate model for forecasting.

### Forecasts from ARIMA(0,1,1)(0,1,2)[12]





Diagnostic Checking: we want to compare the performance of the SARIMA with Holt's-Winter trend and additive smoothing approach.

Method	RMSE	MAE	MAP	MAS	Theil's U
			Е	Е	
SARIMA	163.855	115.020	1.777	0.344	0.368217
*	9	9	8	6	*
SES	457.050	334.720	4.945	1.002	
	0	8	4	9	
HOLT's	456.179	342.033	5.188	1.024	
	1	9	5	8	
SNAIVE	394.986	330.534	5.624	0.990	
	6	8	5	3	
HOLT -	167.402	119.075	1.870	0.356	0.6557*
WINTER	6	7	4	7	
*					

We may say from the above accuracy measurement table that the performance of SARIMA (0, 1, 1) (0, 1, 2) [12] model is better than Holt's-Winter method.

Now, we want to represent the histogram of the respective method sequentially



Figure: Histogram of forecast error of SARIMA (0,1,1)(0,1,2)[12] model



Histogram of forecaster

### Figure: Histogram forecast error of Holt-winter's trend and additive seasonality model

Comment: On the basis of above two histogram of forecast error, it is revealed that the both of two error terms shape is approximately normal distribution. So, the both of the error term represent white-noise. But the SARIMA (0, 1, 1) (0, 1, 2) [12] model exhibits better normality of forecast error than counterpart.

White Noise Test: The following Table represents the white noise assessment of the error term of the fitted model

Test	P-value	H <sub>0</sub>	Decision
Ljung-Box	0.7863	accept	stationary
KPSS	0.1	accept	Stationary
ADF	0.01	Do not accept	Stationary

Above white noise testing approach suggests there is lack of correlation in error term. So, the model is well fitted.

#### CONCLUSION III.

The main goal of this paper was the performance assessment between seasonal ARIMA modeling with Holt-Winters' exponential smoothing approach. The empirical analysis revealed that SARIMA (0, 1, 1) (0, 1, 2) [12] were the better model than counterpart

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# Low Probability of Intercept Frequency Hopping Signal Characterization Comparison using the Spectrogram and the Scalogram<sup>1</sup>

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Abstract- Low probability of intercept radar signals, which are often problematic to detect and characterize, have as their goal 'to see and not be seen'. Digital intercept receivers are currently moving away from Fourier-based analysis and towards classical time-frequency analysis techniques for the purpose of analyzing these low probability of intercept radar signals. This paper presents the novel approach of characterizing low probability of intercept frequency hopping radar signals through utilization and direct comparison of the Spectrogram versus the Scalogram. Two different frequency hopping low probability of intercept radar signals were analyzed(4-component and 8-component). The following metrics were used for evaluation: percent error of: carrier frequency, modulation bandwidth, modulation period, and time-frequency localization. Also used were: percent detection, lowest signal-to-noise ratio for signal detection, and plot (processing) time. Experimental results demonstrate that overall, the Scalogram produced more accurate characterization metrics than the Spectrogram. An improvement in performance may well translate into saved equipment and lives.

GJRE-J Classification : FOR Code: 091599

# LOWPROBABILITY OF INTERCEPTFREDUENCYHOPPINGSIGNALCHARACTERIZATIONCOMPARISONUSING THE SPECTROGRAMAND THE SCALOGRAM

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# Low Probability of Intercept Frequency Hopping Signal Characterization Comparison using the Spectrogram and the Scalogram<sup>1</sup>

Daniel L. Stevens <sup>a</sup> & Stephanie A. Schuckers <sup>o</sup>

Abstract- Low probability of intercept radar signals, which are often problematic to detect and characterize, have as their goal 'to see and not be seen'. Digital intercept receivers are currently moving away from Fourier-based analysis and towards classical time-frequency analysis techniques for the purpose of analyzing these low probability of intercept radar signals. This paper presents the novel approach of characterizing low probability of intercept frequency hopping radar signals through utilization and direct comparison of the Spectrogram versus the Scalogram. Two different frequency hopping low probability of intercept radar signals were analyzed(4-component and 8-component). The following metrics were used for evaluation: percent error of: carrier frequency, modulation bandwidth, modulation period, and time-frequency localization. Also used were: percent detection, lowest signal-to-noise ratio for signal detection, and plot (processing) time. Experimental results demonstrate that the Scalogram produced more overall, accurate characterization metrics than the Spectrogram. An improvement in performance may well translate into saved equipment and lives.

### I. INTRODUCTION

low probability of intercept (LPI) radar that uses frequency hopping techniques changes the transmitting frequency in time over a wide bandwidth in order to prevent an intercept receiver from intercepting the waveform. The frequency slots used are chosen from a frequency hopping sequence, and it is this unknown sequence that gives the radar the advantage over the intercept receiver in terms of processing gain. The frequency sequence appears random to the intercept receiver, and so the possibility of it following the changes in frequency is remote [PAC09]. This prevents a jammer from reactively jamming the transmitted frequency [ADA04]. Frequency hopping radar performance depends only slightly on the code used, given that certain properties are met. This allows for a larger variety of codes, making it more difficult to intercept.

Time-frequency signal analysis involves the analysis and processing of signal s with time - varying

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Author o: Department of Electrical and Computer Engineering, Clarkson University Potsdam, NY 13699. e-mail: sschucke@clarkson.edu frequency content. Such signals are best represented by a time-frequency distribution [PAP95], [HAN00], which is intended to show how the energy of the signal is distributed over the two-dimensional time-frequency plane [WEI03], [LIX08], [OZD03]. Processing of the signal may then exploit the features produced by the concentration of signal energy in two dimensions (time and frequency), instead of only one dimension (time or frequency) [BOA03], [LIY03]. Since noise tends to spread out evenly over the time-frequency domain, while signals concentrate their energies within limited time intervals and frequency bands; the local SNR of a noisy signal can be improved simply by using time-frequency analysis [XIA99]. Also, the intercept receiver can increase its processing gain by implementing timefrequency signal analysis [GUL08].

Time-frequency distributions are useful for the visual interpretation of signal dynamics [RAN01]. An experienced operator can quickly detect a signal and extract the signal parameters by analyzing the time-frequency distribution [ANJ09].

The Spectrogram is defined as the magnitude squared of the Short-Time Fourier Transform (STFT) [HIP00], [HLA92], [MIT01], [PAC09], [BOA03]. For non-stationary signals, the STFT is usually in the form of the Spectrogram [GRI08].

The STFT of a signal x(u) is given in equation 1 as:

$$F_{x}(t,f;h) = \int_{-\infty}^{+\infty} x(u)h(u-t)e^{-j2\pi f u} du$$
 (1)

Where h(t) is a short time analysis window localized around t = 0 and f = 0. Because multipli-cation by the relatively short window h(u - t) effectively suppresses the signal outside a neighborhood around the analysis point u = t, the STFT is a 'local' spectrum of the signal x(u) around t. Think of the window h(t) as sliding along the signal x(u) and for each shift h(u - t) we compute the usual Fourier transform of the product function x(u)h(u - t). The observation window allows localization of the spectrum in time, but also smears the spectrum in frequency in accordance with the uncertainty principle, leading to a trade-off between time resolution and frequency resolution. In general, if the window is short, the time resolution is good, but the frequency resolution is poor, and if the window is long,

<sup>&</sup>lt;sup>1</sup> Approved for Public Release; Distribution Unlimited : 88ABW-2016 2140 20160428

the frequency resolution is good, but the time resolution is poor.

The STFT was the first tool devised for analyzing a signal in both time and frequency simultaneously. For analysis of human speech, the main method was, and still is, the STFT. In general, the STFT is still the most widely used method for studying non-stationary signals [COH95].

The Spectrogram (the squared modulus of the STFT) is given by equation 2 as:

$$S_{x}(t,f) = \left| \int_{-\infty}^{+\infty} x(u) h(u-t) e^{-j2\pi f u} du \right|^{2}$$
(2)

The Spectrogram is a real-valued and nonnegative distribution. Since the window h of the STFT is assumed of unit energy, the Spectrogram satisfies the global energy distribution property. Thus we can interpret the Spectrogram as a measure of the energy of the signal contained in the time-frequency domain centered on the point (t, f) and whose shape is independent of this localization.

Here are some properties of the Spectrogram: 1) Time and Frequency covariance - The Spectrogram preserves time and frequency shifts, thus the spectrogram is an element of the class of quadratic time-frequency distributions that are covariant by translation in time and in frequency (i.e. Cohen's class); 2) Time-Frequency Resolution - The time-frequency resolution of the Spectrogram is limited exactly as it is for the STFT; there is a trade-off between time resolution and frequency resolution. This poor resolution is the main drawback of this representation; 3) Interference Structure - As it is a quadratic (or bilinear) representation, the Spectrogram of the sum of two signals is not the sum of the two Spectrograms (quadratic superposition principle); there is a cross-Spectrogram part and a real part. Thus, as for every quadratic distribution, the Spectrogram presents interference terms: however, those interference terms are restricted to those regions of the time-frequency plane where the signals overlap. Thus if the signal components are sufficiently distant so that their Spectrograms do not overlap significantly, then the interference term will nearly be identically zero [ISI96], [COH95], [HLA92].

The Scalogram is defined as the magnitude squared of the wavelet transform, and can be used as a time-frequency distribution [COH02], [GAL05], [BOA03].

The idea of the wavelet transform (equation (3)) is to project a signal x on a family of zero-mean functions (the wavelets) deduced from an elementary function (the mother wavelet) by translations and dilations:

$$T_{x}(t,a;\Psi) = \int_{-\infty}^{+\infty} x(s)\Psi_{t,a}^{*}(s)ds$$
(3)

Where  $\Psi_{t,a}(s) = |a|^{-1/2} \Psi\left(\frac{s-t}{a}\right)$ . The variable *a* corresponds to a scale factor, in the sense that taking |a| > 1 dilates the wavelet  $\Psi$  and taking |a| < 1 compresses  $\Psi$ . By definition, the wavelet transform is more a time-scale than a time-frequency representation. However, for wavelets which are well localized around a non-zero frequency  $\nu_0$  at a scale = 1, a time-frequency interpretation is possible thanks to the formal identification  $\nu = \frac{\nu_0}{a}$ .

The wavelet transform is of interest for the analysis of non-stationary signals, because it provides still another alternative to the STFT and to many of the guadratic time-frequency distributions. The basic difference between the STFT and the wavelet transform is that the STFT uses a fixed signal analysis window. whereas the wavelet transform uses short windows at high frequencies and long windows at low frequencies. This helps to diffuse the effect of the uncertainty principle by providing good time resolution at high frequencies and good frequency resolution at low frequencies. This approach makes sense especially when the signal at hand has high frequency components for short durations and low frequency components for long durations. The signals encountered in practical applications are often of this type.

The wavelet transform allows localization in both the time domain via translation of the mother wavelet, and in the scale (frequency) domain via dilations. The wavelet is irregular in shape and compactly supported, thus making it an ideal tool for analyzing signals of a transient nature; the irregularity of the wavelet basis lends itself to analysis of signals with discontinuities or sharp changes, while the compactly supported nature of wavelets enables temporal localization of a signal's features [BOA03]. Unlike many of the quadratic functions such as the Wigner-Ville Distribution (WVD) and Choi-Williams Distribution (CWD), the wavelet transform is a linear transformation, therefore cross-term interference is not generated. There is another major difference between the STFT and the wavelet transform; the STFT uses sines and cosines as an orthogonal basis set to which the signal of interest is effectively correlated against, whereas the wavelet transform uses special 'wavelets' which usually comprise an orthogonal basis set. The wavelet transform then computes coefficients, which represents a measure of the similarities, or correlation, of the signal with respect to the set of wavelets. In other words, the wavelet transform of a signal corresponds to its decomposition with respect to a family of functions obtained by dilations (or contractions) and translations (moving window) of an analyzing wavelet.

A filter bank concept is often used to describe the wavelet transform. The wavelet transform can be interpreted as the result of filtering the signal with a set of bandpass filters, each with a different center frequency [GRI08], [FAR96], [SAR98], [SAT98].

Like the design of conventional digital filters, the design of a wavelet filter can be accomplished by using a number of methods including weighted least squares [ALN00], [GOH00], orthogonal matrix methods [ZAH99], nonlinear optimization, optimization of a single parameter (e.g. the passband edge) [ZHA00], and a method that minimizes an objective function that bounds the out-of-tile energy [FAR99].

Here are some properties of the wavelet transform: 1) The wavelet transform is covariant by translation in time and scaling. The corresponding group of transforms is called the Affine group; 2) The signal x can be recovered from its wavelet transform via the synthesis wavelet; 3) Time and frequency resolutions, like in the STFT case, are related via the Heisenberg-Gabor inequality. However in the wavelet transform case, these two resolutions depend on the frequency: the frequency resolution becomes poorer and the time resolution becomes better as the analysis frequency grows;4) Because the wavelet transform is a linear transform, it does not contain cross-term interferences [GRI07], [LAR92].

A similar distribution to the Spectrogram can be defined in the wavelet case. Since the wavelet transform behaves like an orthonormal basis decomposition, it can be shown that it preserves energy:

$$\iint_{-\infty}^{+\infty} |T_x(t,a;\Psi)|^2 dt \frac{da}{a^2} = E_x$$
(4)

where  $E_x$  is the energy of x. This leads us to define the Scalogram (equation (4)) of x as the squared modulus of the wavelet transform. It is an energy distribution of the signal in the time-scale plane, associated with the measure  $\frac{da}{a^2}$ .

As is the case for the wavelet transform, the time and frequency resolutions of the Scalogram are related via the Heisenberg-Gabor principle.

The interference terms of the Scalogram, as for the spectrogram, are also restricted to those regions of the time-frequency plane where the corresponding signals overlap. Therefore, if two signal components are sufficiently far apart in the time-frequency plane, their cross-Scalogram will be essentially zero [ISI96], [HLA92].

For this paper, the Morlet Scalogram will be used. The Morlet wavelet is obtained by taking a complex sine wave and by localizing it with a Gaussian envelope. The Mexican hat wavelet isolates a single bump of the Morlet wavelet. The Morlet wavelet has good focusing in both time and frequency [CHE09].

### II. METHODOLOGY

The methodologies detailed in this section describe the processes involved in obtaining and

comparing metrics between the classical time-frequency analysis techniques of the Spectrogram and the Scalogram for the detection and characterization of low probability of intercept frequency hopping radar signals.

The tools used for this testing were: MATLAB (version 7.12), Signal Processing Toolbox (version 6.15), Wavelet Toolbox (version 4.7), Image Processing Toolbox (version 7.2), Time-Frequency Toolbox (version 1.0) (http://tftb.nongnu.org/).

All testing was accomplished on a desktop computer (HP Compaq, 2.5GHz processor, AMD Athlon 64X2 Dual Core Processor 4800+, 2.00GB Memory (RAM), 32 Bit Operating System).

Testing was performed for 2 different waveforms (4 component frequency hopping, 8 component frequency hopping). For each waveform, parameters were chosen for academic validation of signal processing techniques. Due to computer processing resources they were not meant to represent real-world values. The number of samples for each test was chosen to be 512, which seemed to be the optimum size for the desktop computer. Testing was performed at three different SNR levels: 10dB, 0dB, and the lowest SNR at which the signal could be detected. The noise added was white Gaussian noise, which best reflects the thermal noise present in the IF section of an intercept receiver [PAC09]. Kaiser windowing was used, when windowing was applicable. 50 runs were performed for each test, for statistical purposes. The plots included in this paper were done at a threshold of 5% of the maximum intensity and were linear scale (not dB) of analytic (complex) signals; the color bar represented intensity. The signal processing tools used for each task were the Spectrogram and the Scalogram.

Task 1 consisted of analyzing a frequency hopping (prevalent in the LPI arena [AMS09]) 4component signal whose parameters were: sampling frequency=5KHz; carrier frequencies=1KHz, 1.75KHz, 0.75KHz, 1.25KHz; modulation bandwidth=1KHz; modulation period=.025sec.

Task 2 was similar to Task 1, but for a frequency hopping 8-component signal whose parameters were: sampling frequency=5KHz; carrier frequencies=1.5 KHz, 1KHz, 1.25KHz, 1.5KHz, 1.75KHz, 1.25KHz, 0.75KHz, 1KHz; modulation bandwidth=1KHz; modulation period=.0125sec.

After each particular run of each test, metrics were extracted from the time-frequency representation. The different metrics extracted were as follows:

- 1) *Plot (processing) time:* Time required for plot to be displayed.
- Percent detection: Percent of time signal was detected - signal was declared a detection if any portion of each of the signal components (4 or 8 signal components for frequency hopping) exceeded a set threshold (a certain percentage of

the maximum intensity of the time-frequency representation).

which the signal could be visually detected in the timefrequency representation) (see Figure 1).

Threshold percentages were determined based on visual detections of low SNR signals (lowest SNR at



*Figure 1*: Threshold percentage determination. This plot is an amplitude vs. time (x-z view) of the Spectrogram of a frequency hopping 4-component signal (512 samples, SNR= -2dB). For visually detected low SNR plots (like this one), the percent of max intensity for the peak z-value of each of the signal components was noted (here 98%, 78%, 75%, 63%), and the lowest of these 4 values was recorded (63%). Ten test runs were performed for both time-frequency analysis tools (Spectrogram and Scalogram) for this waveform. The average of these recorded low values was determined and then assigned as the threshold for that particular time-frequency analysis tool. Note - the threshold for the Spectrogram is 60%.

Thresholds were assigned as follows: Spectrogram (60%); Scalogram (50%).

For percent detection determination, these threshold values were included in the time-frequency plot algorithms so that the thresholds could be applied automatically during the plotting process. From the threshold plot, the signal was declared a detection if any portion of each of the signal components was visible (see Figure 2).



*Figure 2 :* Percent detection (time-frequency). Spectrogram of 4-component frequency hopping signal (512 samples, SNR=10dB) with threshold value automatically set to 60%. From this threshold plot, the signal was declared a (visual) detection because at least a portion of each of the 4 FSK signal components was visible.

3) *Carrier frequency:* The frequency corresponding to the maximum intensity of the time-frequency

representation (there are multiple carrier frequencies (4 or 8) for the frequency hopping waveforms).



Y-Z view Spectrogram FSK 4-component fs=5KHz fc=1KHz, 1.75KHz, .75KHz, 1.25KHz modBW=1KHz modper=.025sec #samples=512 SNR=10dB

*Figure 3:* Determination of carrier frequency. Spectrogram of a 4-component frequency hopping signal (512 samples, SNR=10dB). From the frequency-intensity (y-z) view, the 4 maximum intensity values (1 for each carrier

frequency) are manually determined. The frequencies corresponding to those 4 max intensity values are the 4 carrier frequencies (for this plot fc1=996 Hz, fc2=1748Hz, fc3=760Hz, fc4=1250Hz).

 Modulation bandwidth: Distance from highest frequency value of signal (at a threshold of 20% maximum intensity) to lowest frequency value of signal (at same threshold) in Y-direction (frequency).

The threshold percentage was determined based on manual measurement of the modulation bandwidth of the signal in the time-frequency representation. This was accomplished for ten test runs of each time-frequency analysis tool (Spectrogram and Scalogram), for each of the 2 waveforms. During each manual measurement, the max intensity of the high and low measuring points was recorded. The average of the max intensity values for these test runs was 20%. This was adopted as the threshold value, and is representative of what is obtained when performing manual measurements. This 20% threshold was also adapted for determining the modulation period and the time-frequency localization (both are described below).

For modulation bandwidth determination, the 20% threshold value was included in the time-frequency plot algorithms so that the threshold could be applied automatically during the plotting process. From the threshold plot, the modulation bandwidth was manually measured (see Figure 4).

20% Threshold of Max Intensity Spectrogram FSK 4-component fs=5KHz fc=1KHz, 1.75KHz, .75KHz, 1.25KHz modBW=1KHz modper=.025sec #samples=512 SNR=10dB



*Figure 4*: Modulation bandwidth determination. Spectrogram of a 4-component frequency hopping signal (512 samples, SNR=10dB) with threshold value automatically set to 20%. From this threshold plot, the modulation bandwidth was measured manually from the highest frequency value of the signal (top red arrow) to the lowest frequency value of the signal (bottom red arrow) in the y-direction (frequency).

5) *Modulation period:* From Figure 5 (which is at a threshold of 20% maximum intensity), the modulation period is the manual measurement of the width of each of the 4 frequency hopping signals in the x-direction (time), and then the average of the 4 signals is calculated.



*Figure 5*: Modulation period determination. Spectrogram of a 4-component frequency hopping signal (512 samples, SNR=10dB) with threshold value automatically set to 20%. From this threshold plot, the modulation period was measured manually from the left side of the signal (left red arrow) to the right side of the signal (right red arrow) in the x-direction (time). This was done for all 4 signal components, and the average value was determined.

6) *Time-frequency localization:* From Figure 6, the time-frequency localization is a manual measurement (at a threshold of 20% maximum intensity) of the 'thickness' (in the y-direction) of the

center of each of the 4 frequency hopping signal components, and then the average of the 4 values are determined. The average frequency 'thickness' is then converted to: percent of the entire y-axis.



*Figure 6 :* Time-frequency localization determination for the Spectrogram of a 4-component frequency hopping signal (512 samples, SNR=10dB) with threshold value automatically set to 20%. From this threshold plot, the time-

frequency localization was measured manually from the top of the signal (top red arrow) to the bottom of the signal (bottom red arrow) in the y-direction (frequency). This frequency 'thickness' value was then converted to: % of entire y-axis.

7) *Lowest detectable SNR:* The lowest SNR level at which at least a portion of each of the signal components exceeded the set threshold listed in the percent detection section above.

plot algorithms so that the thresholds could be applied automatically during the plotting process. From the threshold plot, the signal was declared a detection if any portion of each of the signal components was visible. The lowest SNR level for which the signal was declared a detection is the lowest detectable SNR (see Figure 7).

For lowest detectable SNR determination, these threshold values were included in the time-frequency

60% Threshold of Max Intensity Spectrogram FSK 4-component bs=5KHz fc=1KHz, 1.75KHz, 1.75KHz, 1.25KHz modBW=iKHz modper=.025sec #samples=512 SNR=-2dB



*Figure 7 :* Lowest detectable SNR. Spectrogram of 4-component frequency hopping signal (512 samples, SNR=-2dB) with threshold value automatically set to 60%. From this threshold plot, the signal was declared a (visual) detection because at least a portion of each of the 4 frequency hopping signal components was visible. For this case, any lower SNR would have been a non-detect. Compare to Figure 2, which is the same plot, except that it has an SNR level equal to 10dB.

The data from all 50 runs for each test was used to produce the actual, error, and percent error for each of these metrics listed above.

The metrics from the Spectrogram were then compared to the metrics from the Scalogram. By and

large, the Scalogram outperformed the Spectrogram, as will be shown in the results section.

### III. Results

Table 1 presents the overall test metrics for the two classical time-frequency analysis techniques used in this testing (Spectrogram versus Scalogram).

*Table 1 :* Overall test metrics (average percent error: carrier frequency, modulation bandwidth, modulation period, time-frequency localization-y; average: percent detection, lowest detectable snr, plot time) for the two classical time-frequency analysis techniques (Spectrogram versus Scalogram).

parameters	Spectrogram	Scalogram
carrier frequency	0.67%	0.44%
modulation bandwidth	25.70%	21.62%
modulation period	11.37%	10.25%
time-frequency localization-y	9.77%	9.44%

percent detection	69.67%	80.84%
lowest detectable snr	-2.0db	-3.0db
Plot time	3.43s	5.62s

From Table 1, the Scalogram outperformed the Spectrogram in average percent error: carrier frequency (0.44% vs. 0.67%), modulation bandwidth (21.62% vs. 25.70%), modulation period (10.25% vs. 11.37%), and time-frequency localization (y-direction) (9.44% vs.

9.77%);and in average: percent detection (80.84% vs. 69.67%), and lowest detectable SNR (-3.0db vs. -2.0db), while the Spectrogram outperformed the Scalogram in average plot time (3.43s vs. 5.62s).

Figure 8 shows comparative plots of the Spectrogram vs. the Scalogram (4 component frequency hopping) at SNRs of 10dB (top), 0dB (middle), and -3dB (bottom).





*Figure 8 :* Comparative plots of the 4-component frequency hopping low probability of intercept radar signals (Spectrogram (left-hand side) vs. the Scalogram (right-hand side)). The SNR for the top row is 10dB, for the middle

row is 0dB, and for the bottom row is -3dB. In general, the Scalogram signals appear more localized ('thinner') than do the Spectrogram signals. In addition, the Scalogram signals appear more readable than the Spectrogram signals at every SNR level.

Figure 9 shows comparative plots of the Spectrogram vs. the Scalogram (8 component frequency hopping) at SNRs of 10dB (top), 0dB (middle), and -3dB (bottom).



*Figure 9 :* Comparative plots of the 8-component frequency hopping low probability of intercept radar signals (Spectrogram (left-hand side) vs. the Scalogram (right-hand side)). The SNR for the top row plots is 10dB, for the middle row plots is 0dB, and for the bottom row plots is -3dB (which is a non-detect for the Spectrogram). In general, the Scalogram signals appear more localized ('thinner') than do the Spectrogram signals. In addition, the Scalogram signals appear more readable than the Spectrogram signals at every SNR level.

### IV. DISCUSSION

This section will elaborate on the results from the previous section.

From Table 1, the performance of the Spectrogram and the Scalogram will be summarized, including strengths, weaknesses, and generic scenarios in which each particular signal analysis tool might be used.

Spectrogram: The Spectrogram outperformed the Scalogram in average plot time (3.43s vs 5.62s). However, the Spectrogram was outperformed by the Scalogram in every other category. The Spectrogram's extreme reduction of cross-term interference is grounds for its good plot time, but at the expense of signal localization (i.e. it produces a 'thicker' signal (as is seen in Figure 8 and Figure 9) – due to the trade-off between cross-term interference and signal localization). This poor signal localization ('thicker' signals) can account for the Spectrogram being outperformed in the areas of average percent error of modulation bandwidth, modulation period, and time-frequency localization (ydirection). The spectrogram might be used in a scenario where a short plot time is necessary, and where signal localization is not an issue. Such a scenario might be a 'quick and dirty' check to see if a signal is present, without precise extraction of its parameters.

The Scalogram: Scalogram outperformed the Spectrogram in every category but plot time. Because of the Spectrogram's extreme reduction of cross-terms at the expense of signal localization (i.e. it produces a 'thicker' signal), the Scalogram was more localized than the Spectrogram, accounting for its better performance in the areas of average percent error of modulation bandwidth, modulation period, and time-frequency localization (y-direction). In addition, since the compactly supported nature of thewavelet (basis of Scalogram) enables temporal localization of a signal's features, this may also have contributed to the the Scalogram's better average percent error of modulation period. Average percent detection and lowest detectable SNR are both based on visual detection in the Time-Frequency representation. Figures 8 and 9 clearly show that the signals in the Scalogram plots are more readable than those in the Spectrogram plots, which accounts for the Scalogram's better average percent detection and lowest detectable SNR. Since the irregularity of the wavelet basis (basis of Scalogram) lends itself to analysis of signals with discontinuities (such as the frequency hopping signals used in this testing), this may have been a contributing factor to the Scalogram's better overall performance versus the Spectrogram. Also, since the wavelet is irregular in shape and compactly supported, it makes it an ideal tool for analyzing signals of transient nature (such as the frequency hopping signals used in this testing), which may also have been a contributing factor to the

Scalogram's better overall perfromance. The scalogram might be used in a scenario where you need good signal localization in a fairly low SNR environment, without tight time constraints.

### V. Conclusions

Digital intercept receivers, whose main job is to detect and extract parameters from low probability of intercept radar signals, are currently moving away from Fourier-based analysis and towards classical timefrequency techniques, such analysis as the Spectrogram, and Scalogram, for the purpose of analyzing low probability of intercept radar signals. Based on the research performed for this paper (the novel direct comparison of the Spectrogram versus the Scalogram for the signal analysis of low probability of intercept frequency hopping radar signals) it was shown that the Scalogram by-and-large outperforms the Spectrogram for analyzing these low probability of intercept radar signals - for reasons brought out in the discussion section above. More accurate characterization metrics could well translate into saved equipment and lives.

Future plans include analysis of an additional low probability of intercept radar waveform (triangular modulated FMCW), again using the Spectrogram and the Scalogram as time-frequency analysis techniques.

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Based on potential and nature, the manuscript can be categorized under the following heads:

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The recommended size of original research paper is less than seven thousand words, review papers fewer than seven thousands words also. Preparation of research paper or how to write research paper, are major hurdle, while writing manuscript. The research articles and research letters should be fewer than three thousand words, the structure original research paper; sometime review paper should be as follows:

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(b) A brief Summary, "Abstract" (less than 150 words) containing the major results and conclusions.

(c) Up to ten keywords, that precisely identifies the paper's subject, purpose, and focus.

(d) An Introduction, giving necessary background excluding subheadings; objectives must be clearly declared.

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(g) Discussion should cover the implications and consequences, not just recapitulating the results; conclusions should be summarizing.

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(i) References in the proper form.

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- One should avoid outdated words.

Keywords are the key that opens a door to research work sources. Keyword searching is an art in which researcher's skills are bound to improve with experience and time.

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#### References

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**27. Refresh your mind after intervals:** Try to give rest to your mind by listening to soft music or by sleeping in intervals. This will also improve your memory.

**28. Make colleagues:** Always try to make colleagues. No matter how sharper or intelligent you are, if you make colleagues you can have several ideas, which will be helpful for your research.

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**33. Report concluded results:** Use concluded results. From raw data, filter the results and then conclude your studies based on measurements and observations taken. Significant figures and appropriate number of decimal places should be used. Parenthetical remarks are prohibitive. Proofread carefully at final stage. In the end give outline to your arguments. Spot out perspectives of further study of this subject. Justify your conclusion by at the bottom of them with sufficient justifications and examples.

**34.** After conclusion: Once you have concluded your research, the next most important step is to present your findings. Presentation is extremely important as it is the definite medium though which your research is going to be in print to the rest of the crowd. Care should be taken to categorize your thoughts well and present them in a logical and neat manner. A good quality research paper format is essential because it serves to highlight your research paper and bring to light all necessary aspects in your research.

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- Write your paper in the form, which is presented in the guidelines using the template.
- Please note the criterion for grading the final paper by peer-reviewers.

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The introduction will be compiled from reference matter and will reflect the design processes or outline of basis that direct you to make study. As you will carry out the process of study, the method and process section will be constructed as like that. The result segment will show related statistics in nearly sequential order and will direct the reviewers next to the similar intellectual paths throughout the data that you took to carry out your study. The discussion section will provide understanding of the data and projections as to the implication of the results. The use of good quality references all through the paper will give the effort trustworthiness by representing an alertness of prior workings.

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To make a paper clear

· Adhere to recommended page limits

#### Mistakes to evade

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- Separating a table/chart or figure impound each figure/table to a single page
- Submitting a manuscript with pages out of sequence

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- · Use paragraphs to split each significant point (excluding for the abstract)
- $\cdot$  Align the primary line of each section
- · Present your points in sound order
- $\cdot$  Use present tense to report well accepted
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The summary should be two hundred words or less. It should briefly and clearly explain the key findings reported in the manuscript-must have precise statistics. It should not have abnormal acronyms or abbreviations. It should be logical in itself. Shun citing references at this point.

An abstract is a brief distinct paragraph summary of finished work or work in development. In a minute or less a reviewer can be taught the foundation behind the study, common approach to the problem, relevant results, and significant conclusions or new questions.

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- Fundamental goal
- To the point depiction of the research
- Consequences, including <u>definite statistics</u> if the consequences are quantitative in nature, account quantitative data; results of any numerical analysis should be reported
- Significant conclusions or questions that track from the research(es)

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### Approach:

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Materials:

- Explain materials individually only if the study is so complex that it saves liberty this way.
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- Do not take in frequently found.
- If use of a definite type of tools.
- Materials may be reported in a part section or else they may be recognized along with your measures.

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- Report the method (not particulars of each process that engaged the same methodology)
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- To be succinct, present methods under headings dedicated to specific dealings or groups of measures
- Simplify details how procedures were completed not how they were exclusively performed on a particular day.
- If well known procedures were used, account the procedure by name, possibly with reference, and that's all.

### Approach:

- It is embarrassed or not possible to use vigorous voice when documenting methods with no using first person, which would focus the reviewer's interest on the researcher rather than the job. As a result when script up the methods most authors use third person passive voice.
- Use standard style in this and in every other part of the paper avoid familiar lists, and use full sentences.

### What to keep away from

- Resources and methods are not a set of information.
- Skip all descriptive information and surroundings save it for the argument.
- Leave out information that is immaterial to a third party.

#### **Results:**

The principle of a results segment is to present and demonstrate your conclusion. Create this part a entirely objective details of the outcome, and save all understanding for the discussion.

The page length of this segment is set by the sum and types of data to be reported. Carry on to be to the point, by means of statistics and tables, if suitable, to present consequences most efficiently. You must obviously differentiate material that would usually be incorporated in a study editorial from any unprocessed data or additional appendix matter that would not be available. In fact, such matter should not be submitted at all except requested by the instructor.



Content

- Sum up your conclusion in text and demonstrate them, if suitable, with figures and tables.
- In manuscript, explain each of your consequences, point the reader to remarks that are most appropriate.
- Present a background, such as by describing the question that was addressed by creation an exacting study.
- Explain results of control experiments and comprise remarks that are not accessible in a prescribed figure or table, if appropriate.

• Examine your data, then prepare the analyzed (transformed) data in the form of a figure (graph), table, or in manuscript form. What to stay away from

- Do not discuss or infer your outcome, report surroundings information, or try to explain anything.
- Not at all, take in raw data or intermediate calculations in a research manuscript.
- Do not present the similar data more than once.
- Manuscript should complement any figures or tables, not duplicate the identical information.
- Never confuse figures with tables there is a difference.

### Approach

- As forever, use past tense when you submit to your results, and put the whole thing in a reasonable order.
- Put figures and tables, appropriately numbered, in order at the end of the report
- If you desire, you may place your figures and tables properly within the text of your results part.

### Figures and tables

- If you put figures and tables at the end of the details, make certain that they are visibly distinguished from any attach appendix materials, such as raw facts
- Despite of position, each figure must be numbered one after the other and complete with subtitle
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### Discussion:

The Discussion is expected the trickiest segment to write and describe. A lot of papers submitted for journal are discarded based on problems with the Discussion. There is no head of state for how long a argument should be. Position your understanding of the outcome visibly to lead the reviewer through your conclusions, and then finish the paper with a summing up of the implication of the study. The purpose here is to offer an understanding of your results and hold up for all of your conclusions, using facts from your research and accepted information, if suitable. The implication of result should be visibly described. generally Infer your data in the conversation in suitable depth. This means that when you clarify an observable fact you must explain mechanisms that may account for the observation. If your results vary from your prospect, make clear why that may have happened. If your results agree, then explain the theory that the proof supported. It is never suitable to just state that the data approved with prospect, and let it drop at that.

- Make a decision if each premise is supported, discarded, or if you cannot make a conclusion with assurance. Do not just dismiss a study or part of a study as "uncertain."
- Research papers are not acknowledged if the work is imperfect. Draw what conclusions you can based upon the results that you have, and take care of the study as a finished work
- You may propose future guidelines, such as how the experiment might be personalized to accomplish a new idea.
- Give details all of your remarks as much as possible, focus on mechanisms.
- Make a decision if the tentative design sufficiently addressed the theory, and whether or not it was correctly restricted.
- Try to present substitute explanations if sensible alternatives be present.
- One research will not counter an overall question, so maintain the large picture in mind, where do you go next? The best studies unlock new avenues of study. What questions remain?
- Recommendations for detailed papers will offer supplementary suggestions.

### Approach:

- When you refer to information, differentiate data generated by your own studies from available information
- Submit to work done by specific persons (including you) in past tense.
- Submit to generally acknowledged facts and main beliefs in present tense.

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Discussion	Well organized, meaningful specification, sound conclusion, logical and concise explanation, highly structured paragraph reference cited	Wordy, unclear conclusion, spurious	Conclusion is not cited, unorganized, difficult to comprehend
References	Complete and correct format, well organized	Beside the point, Incomplete	Wrong format and structuring

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